

# AFE4400 集成模拟前端，用于心率监视器和低成本脉冲血氧计

## 1 特性

- 完全集成 模拟前端，用于脉冲血氧仪 应用范围：
  - 灵活的脉冲排序和定时控制
- 发射：
  - 集成发光二极管 (LED) 驱动器 (H 桥、推挽)
  - 动态范围：95dB
  - LED 电流：
    - 可编程至具有 8 位电流分辨率的 50mA 电流
  - 低功耗：
    - 100 $\mu$ A + 平均 LED 电流
  - 可编程 LED 接通时间
  - 独立的 LED2 和 LED1 电流基准
- 具有高动态范围的接收通道：
  - 13 个无噪声位
  - 低功耗：在使用 3.3V 电源供电时小于 670 $\mu$ A
  - 集成数字环境估算和删减
  - 灵活的接收采样时间
  - 具有可编程 LED 设置的灵活转阻放大器
- 集成式故障诊断：
  - 光电二极管和 LED 开路与短路检测
  - 线缆接通和断开检测
- 电源：
  - Rx = 2.0V 至 3.6V
  - Tx = 3.0V 至 5.25V
- 封装：紧凑型超薄四方扁平无引线 (VQFN)-40 (6mm x 6mm)
- 额定温度范围：0°C 至 70°C

## 2 应用

- 低成本 医疗脉冲血氧仪应用
- 光学心率监视器 (HRM)
- 工业光测量应用

## 3 说明

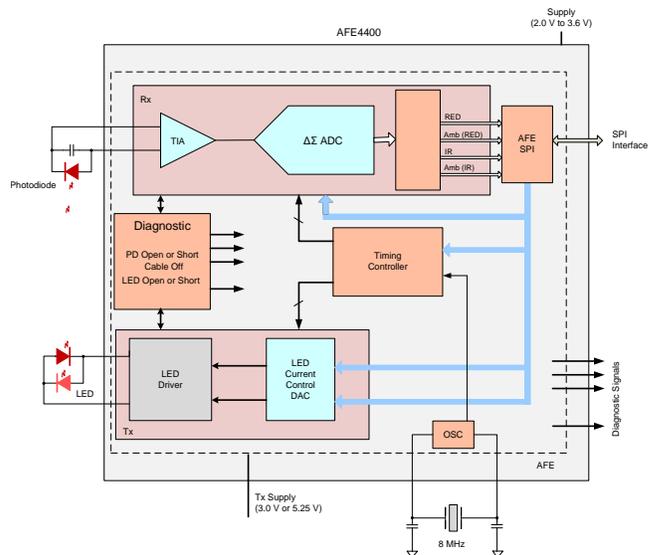
AFE4400 是一款非常适合于脉冲血氧仪应用的完全集成模拟前端 (AFE)。此器件包含一个具有集成模数转换器 (ADC) 的低噪声接收器通道、一个 LED 发射部件和针对传感器以及 LED 故障检测的诊断功能。此器件是一款可配置定时控制器。这个灵活性使得用户能够完全控制器件定时特性。为了简化计时要求并为 AFE4400 提供一个低抖动时钟，还集成了一个由外部晶振供频的振荡器。此器件使用一个 SPI™ 接口与外部微控制器或主机处理器通信。

此器件是一个单体、紧凑型 VQFN-40 封装 (6mm x 6mm) 内的完整 AFE 解决方案，其额定工作温度范围为 0°C 至 70°C。

### 器件信息(1)

部件号	封装	封装尺寸 (标称值)
AFE4400	VQFN (40)	6.00mm x 6.00mm

(1) 如需了解所有可用封装，请见数据表末尾的可订购产品附录。



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• Changed CDM value from $\pm 1500$ to $\pm 250$ in Handling Ratings table .....	7

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• 已更改 集成故障诊断 特性着重号的第二子着重号 .....	1
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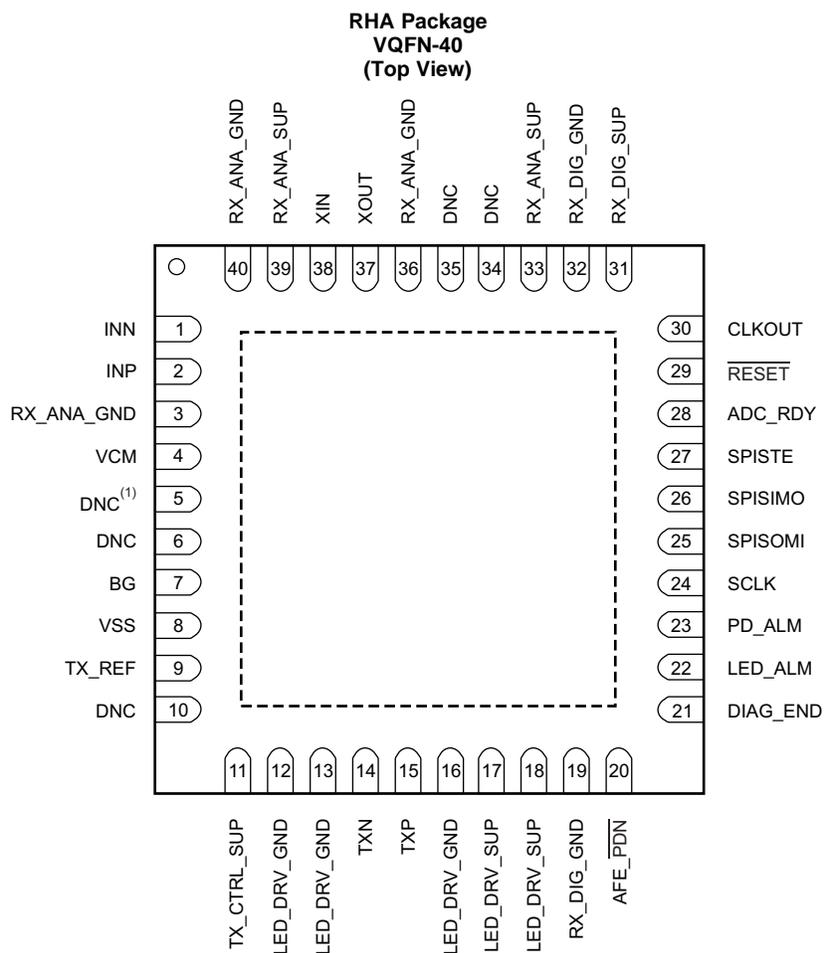
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## 5 Device Family Options

PRODUCT	PACKAGE-LEAD	LED DRIVE CONFIGURATION	LED DRIVE CURRENT (mA, max)	T <sub>x</sub> POWER SUPPLY (V)	OPERATING TEMPERATURE RANGE
AFE4400	VQFN-40	Bridge, push-pull	50	3 to 5.25	0°C to 70°C
AFE4490	VQFN-40	Bridge, push-pull	50, 75, 100, 150, and 200	3 to 5.25	-40°C to 85°C
AFE4403	DSBGA-36	Bridge, push-pull	25, 50, 75, and 100	3 to 5.25	-20°C to 70°C

## 6 Pin Configuration and Functions



(1) DNC = Do not connect.

### Pin Functions

PIN		FUNCTION	DESCRIPTION
NAME	NO.		
ADC_RDY	28	Digital	Output signal that indicates ADC conversion completion. Can be connected to the interrupt input pin of an external microcontroller.
AFE_PDN	20	Digital	AFE-only power-down input; active low. Can be connected to the port pin of an external microcontroller.
BG	7	Reference	Decoupling capacitor for internal band-gap voltage to ground. (2.2- $\mu$ F decoupling capacitor to ground)
CLKOUT	30	Digital	Buffered 4-MHz output clock output. Can be connected to the clock input pin of an external microcontroller.
DIAG_END	21	Digital	Output signal that indicates completion of diagnostics. Can be connected to the port pin of an external microcontroller.
DNC <sup>(1)</sup>	5, 6, 10, 34, 35	—	Do not connect these pins. Leave as open circuit.
INN	1	Analog	Receiver input pin. Connect to photodiode anode.
INP	2	Analog	Receiver input pin. Connect to photodiode cathode.
LED_DRV_GND	12, 13, 16	Supply	LED driver ground pin, H-bridge. Connect to common board ground.
LED_DRV_SUP	17, 18	Supply	LED driver supply pin, H-bridge. Connect to an external power supply capable of supplying the large LED current, which is drawn by this supply pin.
LED_ALM	22	Digital	Output signal that indicates an LED cable fault. Can be connected to the port pin of an external microcontroller.
PD_ALM	23	Digital	Output signal that indicates a PD sensor or cable fault. Can be connected to the port pin of an external microcontroller.
RESET	29	Digital	AFE-only reset input, active low. Can be connected to the port pin of an external microcontroller.
RX_ANA_GND	3, 36, 40	Supply	Rx analog ground pin. Connect to common board ground.
RX_ANA_SUP	33, 39	Supply	Rx analog supply pin; 0.1- $\mu$ F decoupling capacitor to ground
RX_DIG_GND	19, 32	Supply	Rx digital ground pin. Connect to common board ground.
RX_DIG_SUP	31	Supply	Rx digital supply pin; 0.1- $\mu$ F decoupling capacitor to ground
SCLK	24	SPI	SPI clock pin
SPISIMO	26	SPI	SPI serial in master out
SPISOMI	25	SPI	SPI serial out master in
SPISTE	27	SPI	SPI serial interface enable
TX_CTRL_SUP	11	Supply	Transmit control supply pin (0.1- $\mu$ F decoupling capacitor to ground)
TX_REF	9	Reference	Transmitter reference voltage, 0.75 V default after reset. Connect a 2.2- $\mu$ F decoupling capacitor to ground.
TXN	14	Analog	LED driver out B, H-bridge output. Connect to LED.
TXP	15	Analog	LED driver out B, H-bridge output. Connect to LED.
VCM	4	Reference	Input common-mode voltage output. Connect a series resistor (1 k $\Omega$ ) and a decoupling capacitor (10 nF) to ground. The voltage across the capacitor can be used to shield (guard) the INP, INN traces.
VSS	8	Supply	Substrate ground. Connect to common board ground.
XOUT	37	Digital	Crystal oscillator pins. Connect an external 8-MHz crystal between these pins with the correct load capacitor (as specified by vendor) to ground.
XIN	38	Digital	Crystal oscillator pins. Connect an external 8-MHz crystal between these pins with the correct load capacitor (as specified by vendor) to ground.

(1) Leave pins as open circuit. Do not connect.

## 7 Specifications

### 7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

		MIN	MAX	UNIT
RX_ANA_SUP, RX_DIG_SUP to RX_ANA_GND, RX_DIG_GND		-0.3	4	V
TX_CTRL_SUP, LED_DRV_SUP to LED_DRV_GND		-0.3	6	V
RX_ANA_GND, RX_DIG_GND to LED_DRV_GND		-0.3	0.3	V
Analog inputs		RX_ANA_GND – 0.3	RX_ANA_SUP + 0.3	V
Digital inputs		RX_DIG_GND – 0.3	RX_DIG_SUP + 0.3	V
TXP, TXN pins		-0.3	Minimum [6, (LED_DRV_SUP + 0.3)]	V
Input current to any pin except supply pins <sup>(2)</sup>			±7	mA
Input current	Momentary		±50	mA
	Continuous		±7	mA
Operating temperature range		0	70	°C
Maximum junction temperature, T <sub>J</sub>			125	°C

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) Input terminals are diode-clamped to the power-supply rails. Input signals that can swing beyond the supply rails must be current-limited to 10 mA or less.

### 7.2 Handling Ratings

			MIN	MAX	UNIT
T <sub>stg</sub>	Storage temperature range		-60	150	°C
V <sub>(ESD)</sub>	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins <sup>(1)</sup>	-1000	1000	V
		Charged device model (CDM), per JEDEC specification JESD22-C101, all pins <sup>(2)</sup>	-250	250	

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

### 7.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

PARAMETER		MIN	MAX	UNIT		
<b>SUPPLIES</b>						
RX_ANA_SUP	AFE analog supply	2.0	3.6	V		
RX_DIG_SUP	AFE digital supply	2.0	3.6	V		
TX_CTRL_SUP	Transmit controller supply	3.0	5.25	V		
LED_DRV_SUP	Transmit LED driver supply	H-bridge or common anode configuration		[3.0 or (1.0 + V <sub>LED</sub> + V <sub>CABLE</sub> ) <sup>(1)(2)</sup> , whichever is greater]	5.25	V
	Difference between LED_DRV_SUP and TX_CTRL_SUP	-0.3	0.3	V		
<b>TEMPERATURE</b>						
	Specified temperature range	0	70	°C		
	Storage temperature range	-60	150	°C		

- (1) V<sub>LED</sub> refers to the maximum voltage drop across the external LED (at maximum LED current) connected between the TXP and TXN pins (in H-bridge mode) and from the TXP and TXN pins to LED\_DRV\_SUP (in the common anode configuration).
- (2) V<sub>CABLE</sub> refers to voltage drop across any cable, connector, or any other component in series with the LED.

### 7.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>		AFE4400	UNIT
		RHA (VQFN)	
		40 PINS	
R <sub>θJA</sub>	Junction-to-ambient thermal resistance	35	°C/W
R <sub>θJC(top)</sub>	Junction-to-case (top) thermal resistance	31	
R <sub>θJB</sub>	Junction-to-board thermal resistance	26	
ψ <sub>JT</sub>	Junction-to-top characterization parameter	0.1	
ψ <sub>JB</sub>	Junction-to-board characterization parameter	n/a	
R <sub>θJC(bot)</sub>	Junction-to-case (bottom) thermal resistance	n/a	

- (1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).

## 7.5 Electrical Characteristics

Minimum and maximum specifications are at  $T_A = 0^\circ\text{C}$  to  $70^\circ\text{C}$ , typical specifications are at  $T_A = 25^\circ\text{C}$ . All specifications are at  $RX\_ANA\_SUP = RX\_DIG\_SUP = 3\text{ V}$ ,  $TX\_CTRL\_SUP = LED\_DRV\_SUP = 3.3\text{ V}$ , stage 2 amplifier disabled, and  $f_{CLK} = 8\text{ MHz}$ , unless otherwise noted.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>PERFORMANCE (Full-Signal Chain)</b>						
$I_{IN\_FS}$	Full-scale input current	$R_F = 10\text{ k}\Omega$		50		$\mu\text{A}$
		$R_F = 25\text{ k}\Omega$		20		$\mu\text{A}$
		$R_F = 50\text{ k}\Omega$		10		$\mu\text{A}$
		$R_F = 100\text{ k}\Omega$		5		$\mu\text{A}$
		$R_F = 250\text{ k}\Omega$		2		$\mu\text{A}$
		$R_F = 500\text{ k}\Omega$		1		$\mu\text{A}$
		$R_F = 1\text{ M}\Omega$			0.5	
PRF	Pulse repetition frequency		62.5		5000	SPS
DC <sub>PRF</sub>	PRF duty cycle				25%	
CMRR	Common-mode rejection ratio	$f_{CM} = 50\text{ Hz}$ and $60\text{ Hz}$ , LED1 and LED2 with $R_{SERIES} = 500\text{ k}\Omega$ , $R_F = 500\text{ k}\Omega$		75		dB
		$f_{CM} = 50\text{ Hz}$ and $60\text{ Hz}$ , LED1-AMB and LED2-AMB with $R_{SERIES} = 500\text{ k}\Omega$ , $R_F = 500\text{ k}\Omega$		95		dB
PSRR	Power-supply rejection ratio	$f_{PS} = 50\text{ Hz}$ , $60\text{ Hz}$ at PRF = $200\text{ Hz}$		100		dB
		$f_{PS} = 50\text{ Hz}$ , $60\text{ Hz}$ at PRF = $600\text{ Hz}$		106		dB
PSRR <sub>LED</sub>	PSRR, transmit LED driver	With respect to ripple on LED_DRV_SUP		75		dB
PSRR <sub>TX</sub>	PSRR, transmit control	With respect to ripple on TX_CTRL_SUP		60		dB
PSRR <sub>Rx</sub>	PSRR, receiver	With respect to ripple on RX_ANA_SUP and RX_DIG_SUP		60		dB
	Total integrated noise current, input-referred (receiver with transmitter loop back, 0.1-Hz to 5-Hz bandwidth)	$R_F = 100\text{ k}\Omega$ , PRF = $600\text{ Hz}$ , duty cycle = 5%		36		$\text{pA}_{RMS}$
		$R_F = 500\text{ k}\Omega$ , PRF = $600\text{ Hz}$ , duty cycle = 5%		13		$\text{pA}_{RMS}$
$N_{FB}$	Noise-free bits (receiver with transmitter loop back, 0.1-Hz to 5-Hz bandwidth)	$R_F = 100\text{ k}\Omega$ , PRF = $600\text{ Hz}$ , duty cycle = 5%		14.3		Bits
		$R_F = 500\text{ k}\Omega$ , PRF = $600\text{ Hz}$ , duty cycle = 5%		13.5		Bits
<b>RECEIVER FUNCTIONAL BLOCK LEVEL SPECIFICATION</b>						
	Total integrated noise current, input referred (receiver alone) over 0.1-Hz to 5-Hz bandwidth	$R_F = 500\text{ k}\Omega$ , ambient cancellation enabled, stage 2 gain = 4, PRF = $1200\text{ Hz}$ , LED duty cycle = 25%		1.4		$\text{pA}_{RMS}$
		$R_F = 500\text{ k}\Omega$ , ambient cancellation enabled, stage 2 gain = 4, PRF = $1200\text{ Hz}$ , LED duty cycle = 5%		5		$\text{pA}_{RMS}$
<b>I-V TRANSIMPEDANCE AMPLIFIER</b>						
G	Gain	$R_F = 10\text{ k}\Omega$ to $1\text{ M}\Omega$	See the <a href="#">Receiver Channel</a> section for details			$\text{V}/\mu\text{A}$
	Gain accuracy					$\pm 7\%$
	Feedback resistance	$R_F$	10k, 25k, 50k, 100k, 250k, 500k, and 1M			$\Omega$
	Feedback resistor tolerance	$R_F$				$\pm 20\%$
	Feedback capacitance	$C_F$	5, 10, 25, 50, 100, and 250			pF
	Feedback capacitor tolerance	$C_F$				$\pm 20\%$
	Full-scale differential output voltage					1
	Common-mode voltage on input pins	Set internally				0.9
	External differential input capacitance	Includes equivalent capacitance of photodiode, cables, EMI filter, and so forth	10		1000	pF
	Shield output voltage, $V_{CM}$	With a 1-k $\Omega$ series resistor and a 10-nF decoupling capacitor to ground				0.9

## Electrical Characteristics (continued)

Minimum and maximum specifications are at  $T_A = 0^\circ\text{C}$  to  $70^\circ\text{C}$ , typical specifications are at  $T_A = 25^\circ\text{C}$ . All specifications are at  $RX\_ANA\_SUP = RX\_DIG\_SUP = 3\text{ V}$ ,  $TX\_CTRL\_SUP = LED\_DRV\_SUP = 3.3\text{ V}$ , stage 2 amplifier disabled, and  $f_{CLK} = 8\text{ MHz}$ , unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
<b>AMBIENT CANCELLATION STAGE</b>						
Gain		0, 3.5, 6, 9.5, and 12			dB	
Current DAC range		0			10	$\mu\text{A}$
Current DAC step size		1				$\mu\text{A}$
<b>LOW-PASS FILTER</b>						
Low-pass corner frequency	3-dB attenuation	500				Hz
Pass-band attenuation, 2 Hz to 10 Hz	Duty cycle = 25%	0.004				dB
	Duty cycle = 10%	0.041				dB
Filter settling time	After diagnostics mode	28				ms
<b>ANALOG-TO-DIGITAL CONVERTER</b>						
Resolution		22				Bits
Sample rate	See the <a href="#">ADC Operation and Averaging Module</a> section	4 × PRF				SPS
ADC full-scale voltage		±1.2				V
ADC conversion time	See the <a href="#">ADC Operation and Averaging Module</a> section	50	PRF / 4			$\mu\text{s}$
ADC reset time		2				$t_{CLK}$
<b>TRANSMITTER</b>						
Output current range		Selectable, 0 to 50 (see the <a href="#">LEDCNTRL: LED Control Register</a> for details)				mA
LED current DAC error		±10%				
Output current resolution		8				Bits
Transmitter noise dynamic range, over 0.1-Hz to 5-Hz bandwidth	At 5-mA output current	95				dB
	At 25-mA output current	95				dB
	At 50-mA output current	95				dB
Voltage on TXP (or TXN) pin when low-side switch connected to TXP (or TXN) turns on	At 50-mA output current	1.0 + (voltage drop across LED, cable, and so forth) to 5.25				V
Minimum sample time of LED1 and LED2 pulses		50				$\mu\text{s}$
LED current DAC leakage current	LED_ON = 0	1				$\mu\text{A}$
	LED_ON = 1	50				$\mu\text{A}$
LED current DAC linearity	Percent of full-scale current	0.5%				
Output current settling time (with resistive load)	From zero current to 50 mA	7				$\mu\text{s}$
	From 50 mA to zero current	7				$\mu\text{s}$
<b>DIAGNOSTICS</b>						
Duration of diagnostics state machine	Start of diagnostics after the DIAG_EN register bit is set. End of diagnostic is indicated by DIAG_END going high.	16				ms
Open fault resistance		> 100				k $\Omega$
Short fault resistance		< 10				k $\Omega$
<b>INTERNAL OSCILLATOR</b>						
$f_{CLKOUT}$ CLKOUT frequency	With an 8-MHz crystal connected to the XIN, XOUT pins	4				MHz
CLKOUT duty cycle		50%				
Crystal oscillator start-up time	With an 8-MHz crystal connected to the XIN, XOUT pins	200				$\mu\text{s}$

## Electrical Characteristics (continued)

Minimum and maximum specifications are at  $T_A = 0^\circ\text{C}$  to  $70^\circ\text{C}$ , typical specifications are at  $T_A = 25^\circ\text{C}$ . All specifications are at  $RX\_ANA\_SUP = RX\_DIG\_SUP = 3\text{ V}$ ,  $TX\_CTRL\_SUP = LED\_DRV\_SUP = 3.3\text{ V}$ , stage 2 amplifier disabled, and  $f_{CLK} = 8\text{ MHz}$ , unless otherwise noted.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>EXTERNAL CLOCK</b>					
Maximum allowable external clock jitter	For SPO2 applications		50		ps
	For optical heart rate only			1000	ps
External clock input frequency	$\pm 10\%$		8		MHz
External clock input voltage	Voltage input high ( $V_{IH}$ )	$0.75 \times RX\_DIG\_SUP$			V
	Voltage input low ( $V_{IL}$ )	$0.25 \times RX\_DIG\_SUP$			V
<b>TIMING</b>					
Wake-up time from complete power-down			1000		ms
Wake-up time from Rx power-down			100		$\mu\text{s}$
Wake-up time from Tx power-down			1000		ms
$t_{RESET}$	Active low RESET pulse duration		1		ms
$t_{DIAGEND}$	DIAG_END pulse duration at the completion of diagnostics		4		CLKOUT cycles
$t_{ADCRDY}$	ADC_RDY pulse duration		1		CLKOUT cycle
<b>DIGITAL SIGNAL CHARACTERISTICS</b>					
$V_{IH}$	Logic high input voltage	AFE_PDN, SCLK, SPISIMO, SPISTE, RESET	$0.8 DVDD$	$DVDD + 0.1$	V
$V_{IL}$	Logic low input voltage	AFE_PDN, SCLK, SPISIMO, SPISTE, RESET	-0.1	$0.2 DVDD$	V
$I_{IN}$	Logic input current	$0\text{ V} < V_{DigitalInput} < DVDD$	-10	10	$\mu\text{A}$
$V_{OH}$	Logic high output voltage	DIAG_END, LED_ALM, PD_ALM, SPISOMI, ADC_RDY, CLKOUT	$0.9 DVDD$	$> (RX\_DIG\_SUP - 0.2\text{ V})$	V
$V_{OL}$	Logic low output voltage	DIAG_END, LED_ALM, PD_ALM, SPISOMI, ADC_RDY, CLKOUT	$< 0.4$	$0.1 DVDD$	V
<b>SUPPLY CURRENT</b>					
Receiver analog supply current		RX_ANA_SUP = 3.0 V, with 8-MHz clock running, Rx stage 2 disabled		0.6	mA
		RX_ANA_SUP = 3.0 V, with 8-MHz clock running, Rx stage 2 enabled		0.7	mA
Receiver digital supply current		RX_DIG_SUP = 3.0 V		0.27	mA
LED_DRV_SUP	LED driver supply current	With zero LED current setting		55	$\mu\text{A}$
TX_CTRL_SUP	Transmitter control supply current			15	$\mu\text{A}$
Complete power-down (using AFE_PDN pin)		Receiver current only (RX_ANA_SUP)		3	$\mu\text{A}$
		Receiver current only (RX_DIG_SUP)		3	$\mu\text{A}$
		Transmitter current only (LED_DRV_SUP)		1	$\mu\text{A}$
		Transmitter current only (TX_CTRL_SUP)		1	$\mu\text{A}$
Power-down Rx alone		Receiver current only (RX_ANA_SUP)		220	$\mu\text{A}$
		Receiver current only (RX_DIG_SUP)		220	$\mu\text{A}$
Power-down Tx alone		Transmitter current only (LED_DRV_SUP)		2	$\mu\text{A}$
		Transmitter current only (TX_CTRL_SUP)		2	$\mu\text{A}$

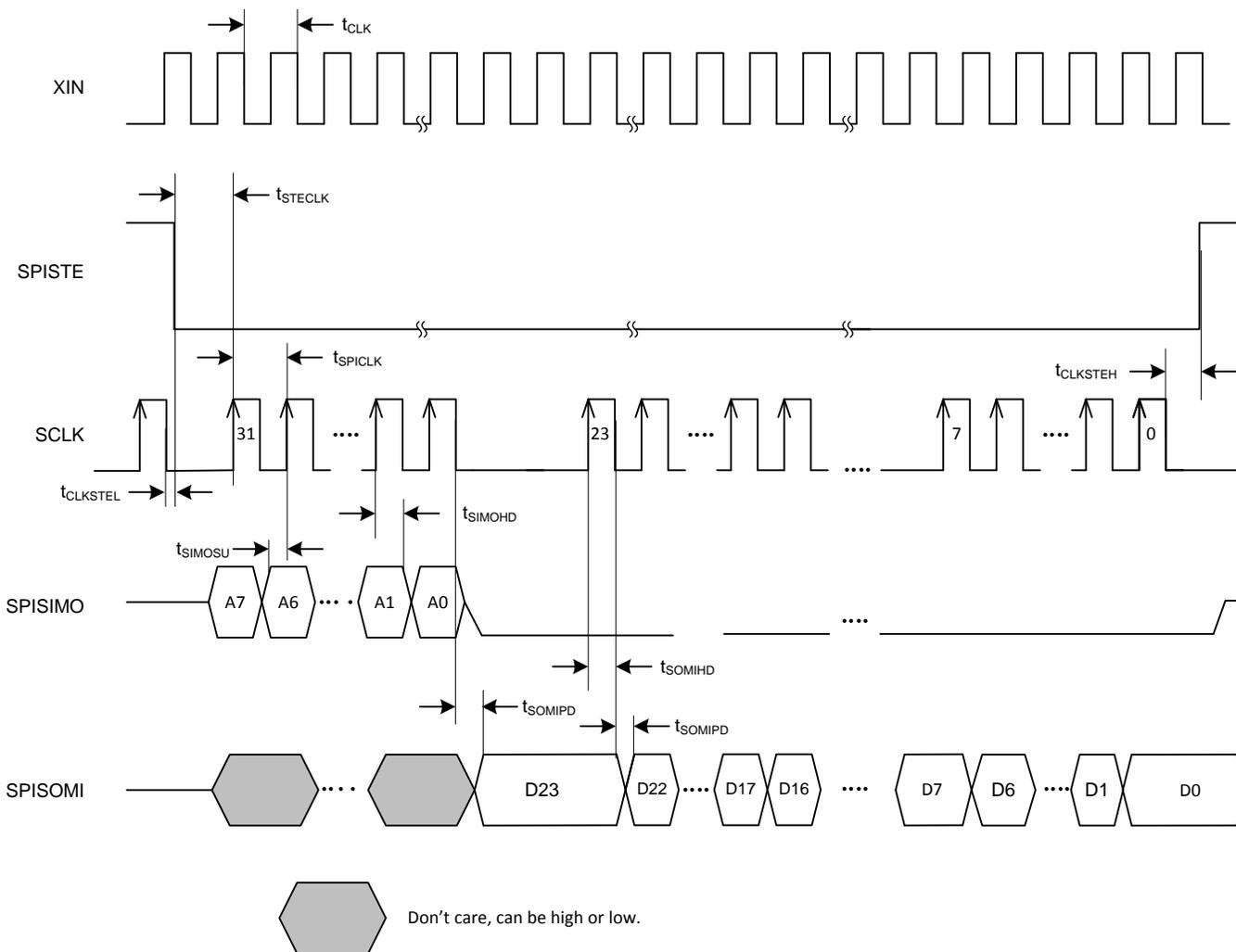
**Electrical Characteristics (continued)**

Minimum and maximum specifications are at  $T_A = 0^\circ\text{C}$  to  $70^\circ\text{C}$ , typical specifications are at  $T_A = 25^\circ\text{C}$ . All specifications are at  $RX\_ANA\_SUP = RX\_DIG\_SUP = 3\text{ V}$ ,  $TX\_CTRL\_SUP = LED\_DRV\_SUP = 3.3\text{ V}$ , stage 2 amplifier disabled, and  $f_{CLK} = 8\text{ MHz}$ , unless otherwise noted.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>POWER DISSIPATION</b>						
Quiescent power dissipation		Normal operation (excluding LEDs)		2.84		mW
		Power-down		0.1		mW
Power-down with the AFE_PDN pin	LED_DRV_SUP	LED_DRV_SUP current value. Does not include LED current.		1		$\mu\text{A}$
	TX_CTRL_SUP			1		$\mu\text{A}$
	RX_ANA_SUP			5		$\mu\text{A}$
	RX_DIG_SUP			0.1		$\mu\text{A}$
Power-down with the PDNAFE register bit	LED_DRV_SUP	LED_DRV_SUP current value. Does not include LED current.		1		$\mu\text{A}$
	TX_CTRL_SUP			1		$\mu\text{A}$
	RX_ANA_SUP			15		$\mu\text{A}$
	RX_DIG_SUP			20		$\mu\text{A}$
Power-down Rx	LED_DRV_SUP	LED_DRV_SUP current value. Does not include LED current.		50		$\mu\text{A}$
	TX_CTRL_SUP			15		$\mu\text{A}$
	RX_ANA_SUP			220		$\mu\text{A}$
	RX_DIG_SUP			220		$\mu\text{A}$
Power-down Tx	LED_DRV_SUP	LED_DRV_SUP current value. Does not include LED current.		2		$\mu\text{A}$
	TX_CTRL_SUP			2		$\mu\text{A}$
	RX_ANA_SUP			600		$\mu\text{A}$
	RX_DIG_SUP			230		$\mu\text{A}$
After reset, with 8-MHz clock running	LED_DRV_SUP	LED_DRV_SUP current value. Does not include LED current.		55		$\mu\text{A}$
	TX_CTRL_SUP			15		$\mu\text{A}$
	RX_ANA_SUP			600		$\mu\text{A}$
	RX_DIG_SUP			230		$\mu\text{A}$
With stage 2 mode enabled and 8-MHz clock running	LED_DRV_SUP	LED_DRV_SUP current value. Does not include LED current.		55		$\mu\text{A}$
	TX_CTRL_SUP			15		$\mu\text{A}$
	RX_ANA_SUP			700		$\mu\text{A}$
	RX_DIG_SUP			270		$\mu\text{A}$

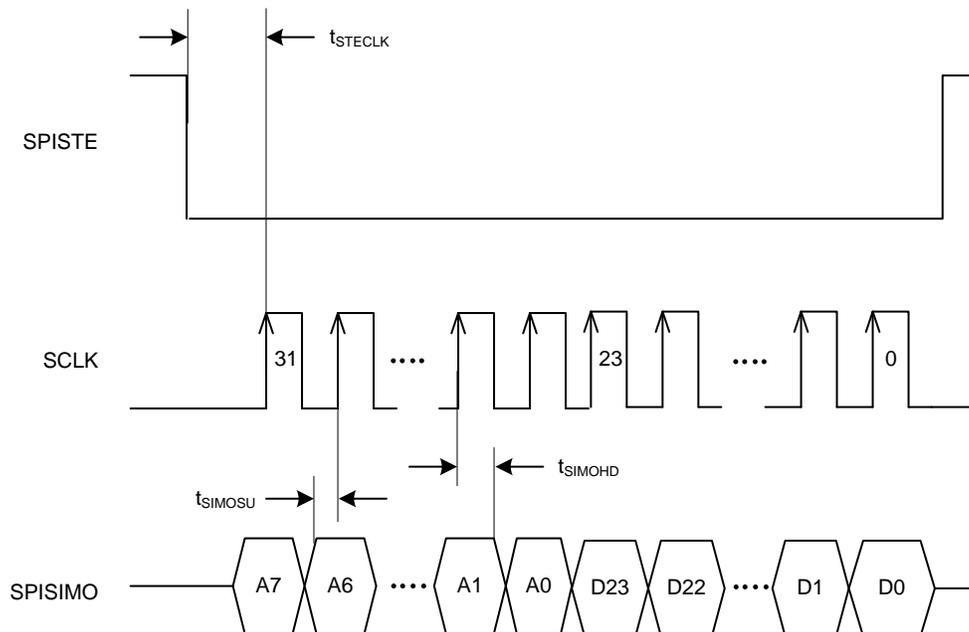
### 7.6 Timing Requirements

PARAMETER		MIN	TYP	MAX	UNIT
t <sub>CLK</sub>	Clock frequency on the XIN pin		8		MHz
t <sub>SCLK</sub>	Serial shift clock period	62.5			ns
t <sub>STECLK</sub>	STE low to SCLK rising edge, setup time	10			ns
t <sub>CLKSTEH,L</sub>	SCLK transition to SPI STE high or low	10			ns
t <sub>SIMOSU</sub>	SIMO data to SCLK rising edge, setup time	10			ns
t <sub>SIMOHD</sub>	Valid SIMO data after SCLK rising edge, hold time	10			ns
t <sub>SOMIPD</sub>	SCLK falling edge to valid SOMI, setup time	17			ns
t <sub>SOMIHD</sub>	SCLK rising edge to invalid data, hold time	0.5			t <sub>SCLK</sub>



- (1) The SPI\_READ register bit must be enabled before attempting a register read.
- (2) Specify the register address whose contents must be read back on A[7:0].
- (3) The AFE outputs the contents of the specified register on the SPISOMI pin.

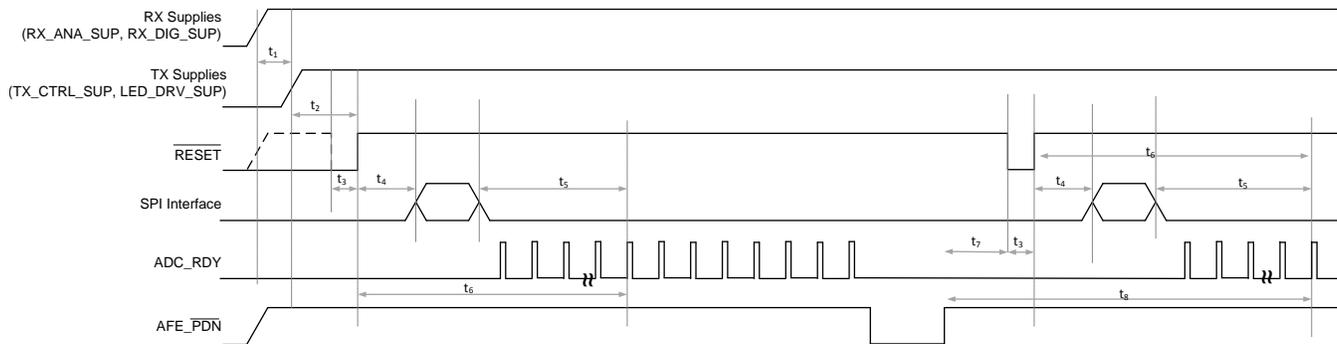
**Figure 1. Serial Interface Timing Diagram, Read Operation<sup>(1)(2)(3)</sup>**


**Figure 2. Serial Interface Timing Diagram, Write Operation**

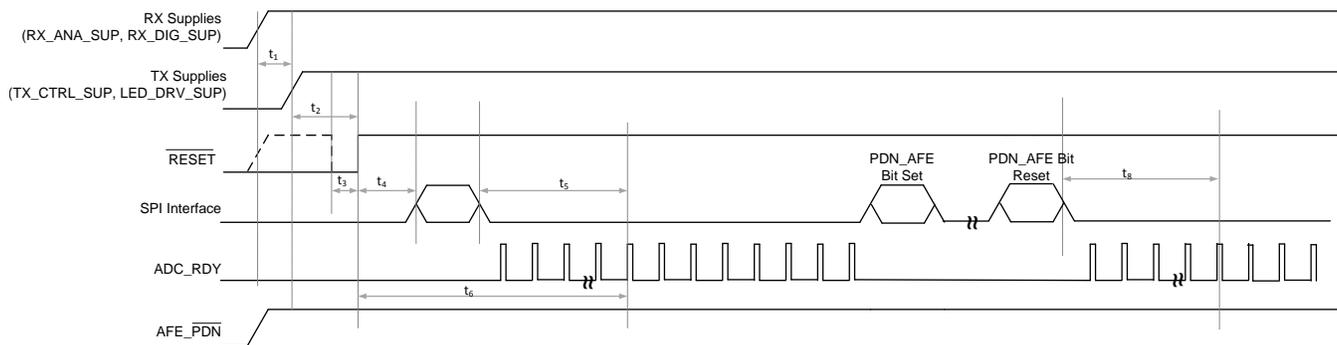
## 7.7 Timing Requirements: Supply Ramp and Power-Down

PARAMETER	VALUE
$t_1$ Time between Rx and Tx supplies ramping up	Keep as small as possible (for example, $\pm 10$ ms)
$t_2$ Time between both supplies stabilizing and high-going $\overline{\text{RESET}}$ edge	> 100 ms
$t_3$ $\overline{\text{RESET}}$ pulse duration	> 0.5 ms
$t_4$ Time between $\overline{\text{RESET}}$ and SPI commands	> 1 $\mu\text{s}$
$t_5$ Time between SPI commands and the $\overline{\text{ADC\_RESET}}$ which corresponds to valid data	> 3 ms of cumulative sampling time in each phase <sup>(1)(2)(3)</sup>
$t_6$ Time between $\overline{\text{RESET}}$ pulse and high-accuracy data coming out of the signal chain	> 1 s <sup>(3)</sup>
$t_7$ Time from $\overline{\text{AFE\_PDN}}$ high-going edge and $\overline{\text{RESET}}$ pulse <sup>(4)</sup>	> 100 ms
$t_8$ Time from $\overline{\text{AFE\_PDN}}$ high-going edge (or $\overline{\text{PDN\_AFE}}$ bit reset) to high-accuracy data coming out of the signal chain	> 1 s <sup>(3)</sup>

- (1) This time is required for each of the four switched RC filters to fully settle to the new settings. The same time is applicable whenever there is a change to any of the signal chain controls (for example, LED current setting, TIA gain, and so forth).
- (2) If the SPI commands involve a change in the TX\_REF value from its default, then there is additional wait time of approximately 1 s (for a 2.2- $\mu\text{F}$  decoupling capacitor on the TX\_REF pin).
- (3) Dependent on the value of the capacitors on the BG and TX\_REF pins. The 1-s wait time is necessary when the capacitors are 2.2  $\mu\text{F}$  and scale down proportionate to the capacitor value. A very low capacitor (for example, 0.1  $\mu\text{F}$ ) on these pins causes the transmitter dynamic range to reduce to approximately 100 dB.
- (4) After an active power-down from  $\overline{\text{AFE\_PDN}}$ , the device should be reset using a low-going  $\overline{\text{RESET}}$  pulse.



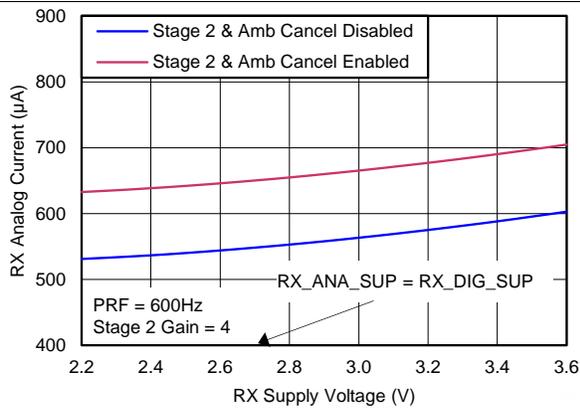
**Figure 3. Supply Ramp and Hardware Power-Down Timing**



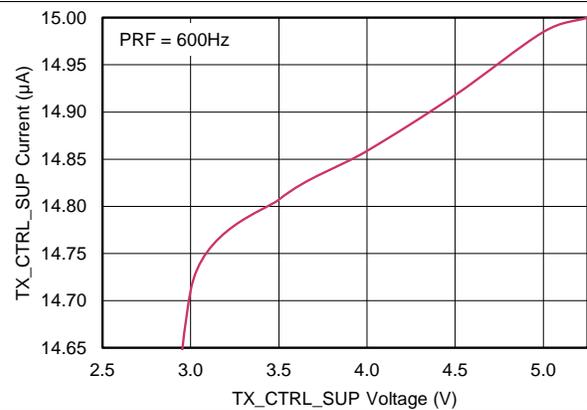
**Figure 4. Supply Ramp and Software Power-Down Timing**

## 7.8 Typical Characteristics

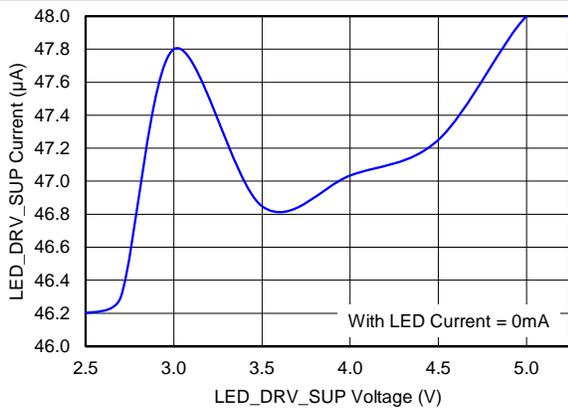
Minimum and maximum specifications are at  $T_A = 0^\circ\text{C}$  to  $70^\circ\text{C}$ . Typical specifications are at  $T_A = 25^\circ\text{C}$ ,  $RX\_ANA\_SUP = RX\_DIG\_SUP = 3.0\text{ V}$ ,  $TX\_CTRL\_SUP = LED\_DRV\_SUP = 3.3\text{ V}$ , and  $f_{CLK} = 8\text{ MHz}$ , unless otherwise noted.



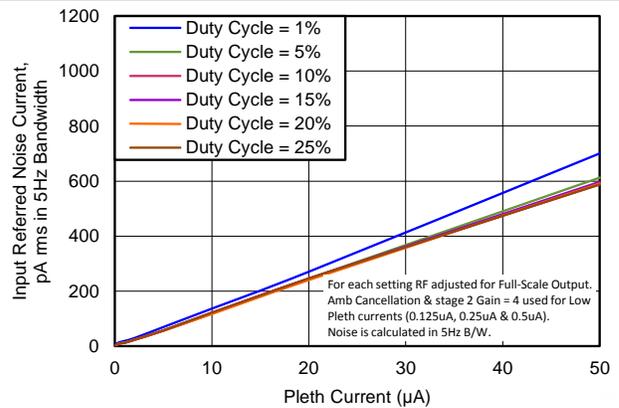
**Figure 5. Total Rx Current vs Rx Supply Voltage**



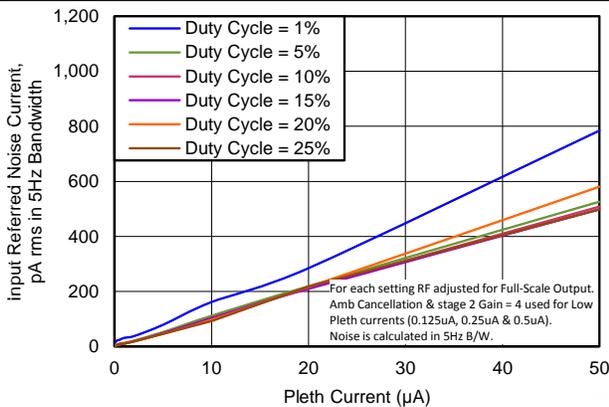
**Figure 6. TX\_CTRL\_SUP Current vs TX\_CTRL\_SUP Voltage**



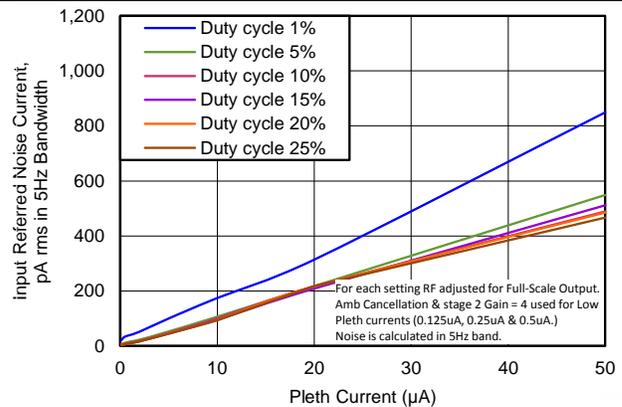
**Figure 7. LED\_DRV\_SUP Current vs LED\_DRV\_SUP Voltage**



**Figure 8. Input-Referred Noise Current vs Pleth Current (PRF = 100 Hz)**



**Figure 9. Input-Referred Noise Current vs Pleth Current (PRF = 300 Hz)**



**Figure 10. Input-Referred Noise Current vs Pleth Current (PRF = 600 Hz)**

Typical Characteristics (continued)

Minimum and maximum specifications are at  $T_A = 0^\circ\text{C}$  to  $70^\circ\text{C}$ . Typical specifications are at  $T_A = 25^\circ\text{C}$ ,  $RX\_ANA\_SUP = RX\_DIG\_SUP = 3.0\text{ V}$ ,  $TX\_CTRL\_SUP = LED\_DRV\_SUP = 3.3\text{ V}$ , and  $f_{CLK} = 8\text{ MHz}$ , unless otherwise noted.

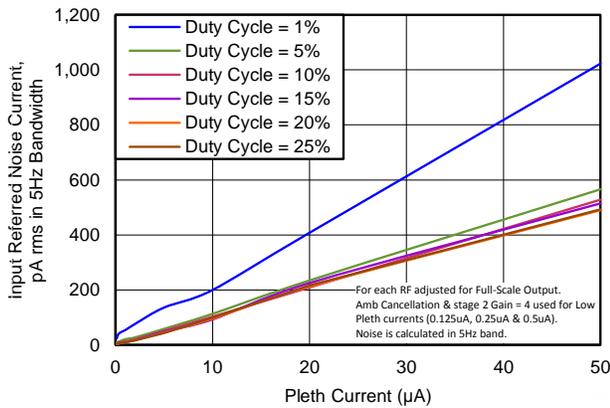


Figure 11. Input-Referred Noise Current vs Pleth Current (PRF = 1200 Hz)

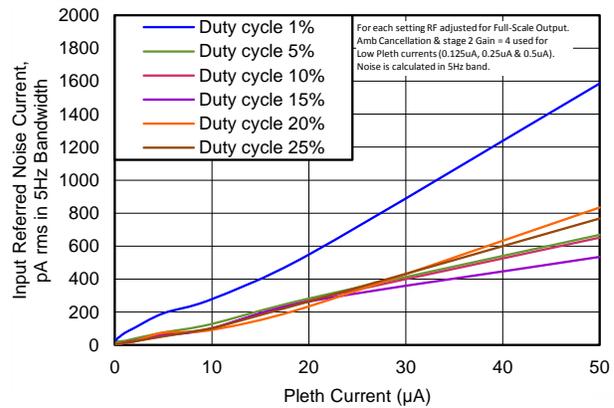


Figure 12. Input-Referred Noise Current vs Pleth Current (PRF = 2500 Hz)

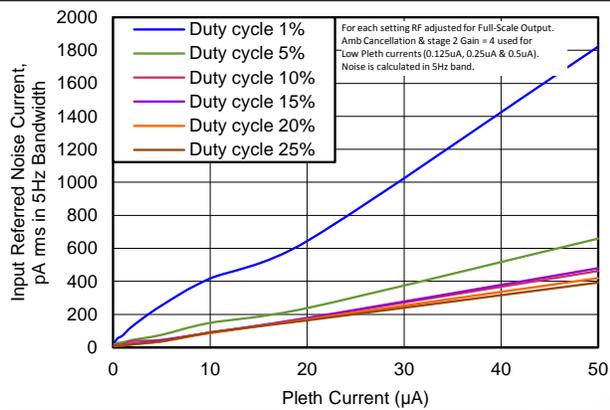


Figure 13. Input-Referred Noise Current vs Pleth Current (PRF = 5000 Hz)

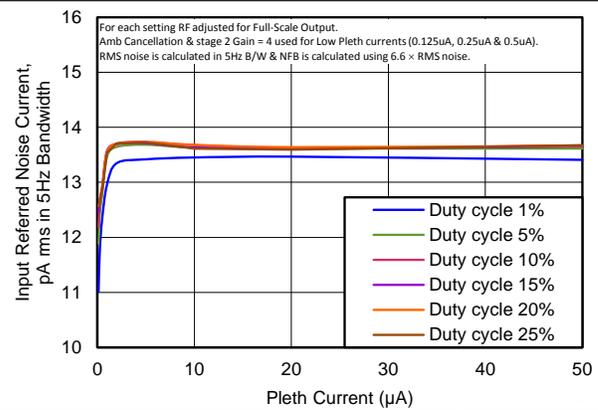


Figure 14. Noise-Free Bits vs Pleth Current (PRF = 100 Hz)

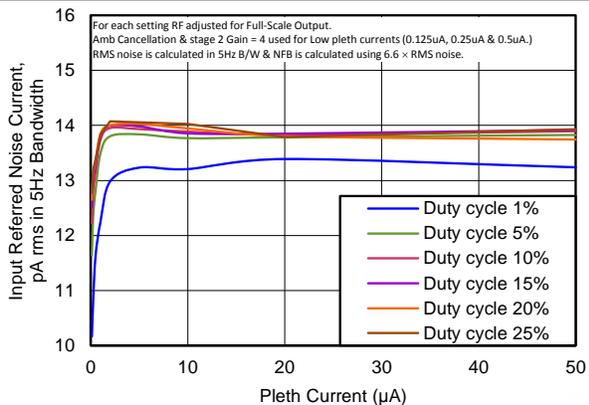


Figure 15. Noise-Free Bits vs Pleth Current (PRF = 300 Hz)

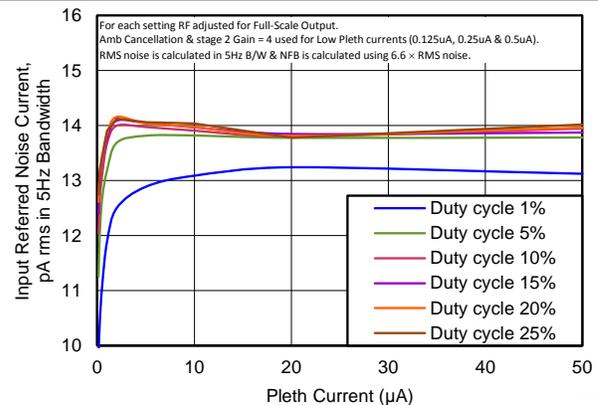


Figure 16. Noise-Free Bits vs Pleth Current (PRF = 600 Hz)

Typical Characteristics (continued)

Minimum and maximum specifications are at  $T_A = 0^\circ\text{C}$  to  $70^\circ\text{C}$ . Typical specifications are at  $T_A = 25^\circ\text{C}$ ,  $RX\_ANA\_SUP = RX\_DIG\_SUP = 3.0\text{ V}$ ,  $TX\_CTRL\_SUP = LED\_DRV\_SUP = 3.3\text{ V}$ , and  $f_{CLK} = 8\text{ MHz}$ , unless otherwise noted.

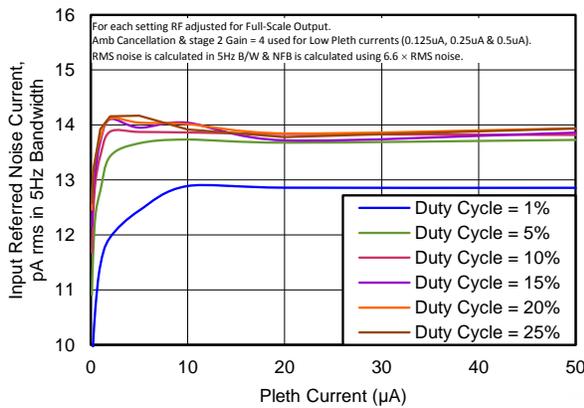


Figure 17. Noise-Free Bits vs Pleth Current (PRF = 1200 Hz)

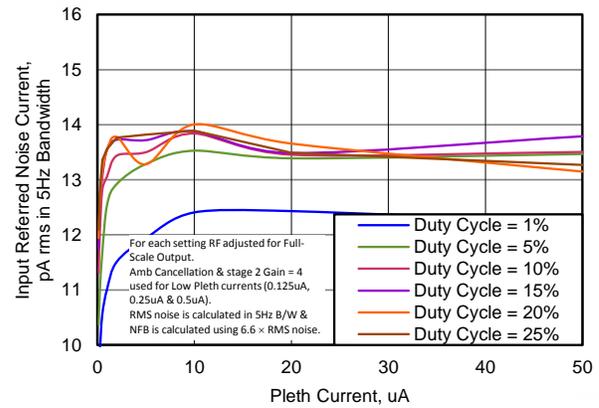


Figure 18. Noise-Free Bits vs Pleth Current (PRF = 2500 Hz)

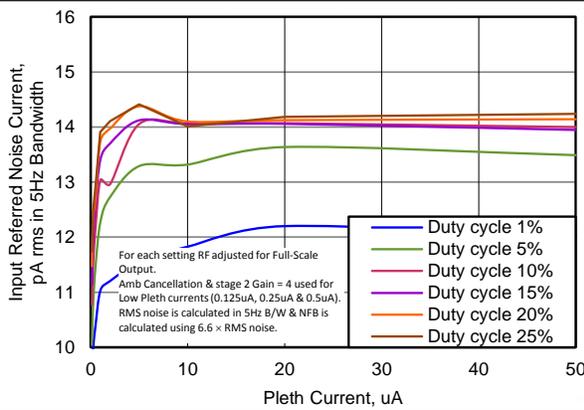


Figure 19. Noise-Free Bits vs Pleth Current (PRF = 5000 Hz)

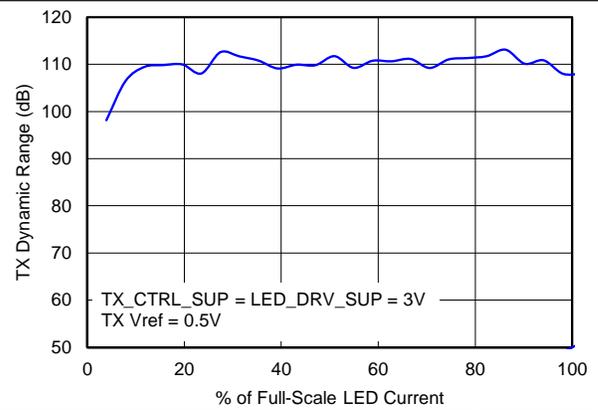


Figure 20. Transmitter Dynamic Range (5-Hz BW)

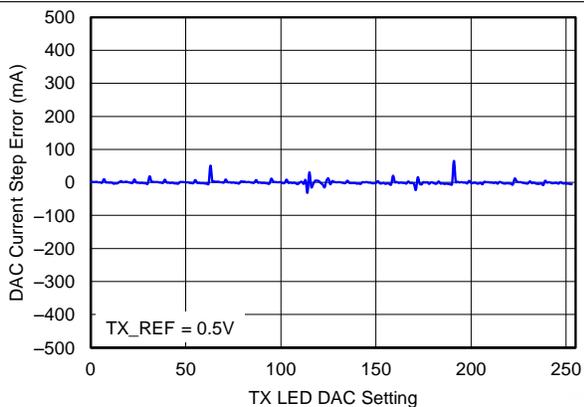


Figure 21. Transmitter DAC Current Step Error (50 mA, Max)

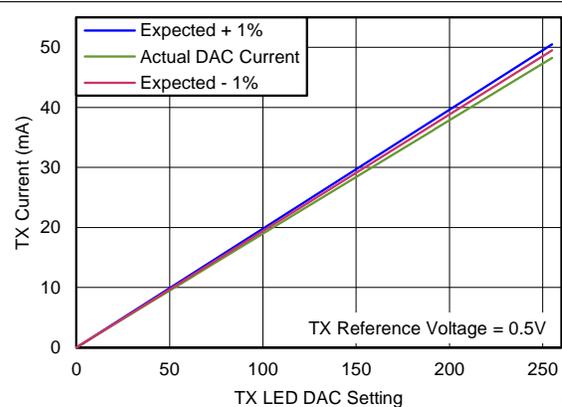


Figure 22. Transmitter Current Linearity (50-mA Range)

Typical Characteristics (continued)

Minimum and maximum specifications are at  $T_A = 0^\circ\text{C}$  to  $70^\circ\text{C}$ . Typical specifications are at  $T_A = 25^\circ\text{C}$ ,  $RX\_ANA\_SUP = RX\_DIG\_SUP = 3.0\text{ V}$ ,  $TX\_CTRL\_SUP = LED\_DRV\_SUP = 3.3\text{ V}$ , and  $f_{CLK} = 8\text{ MHz}$ , unless otherwise noted.

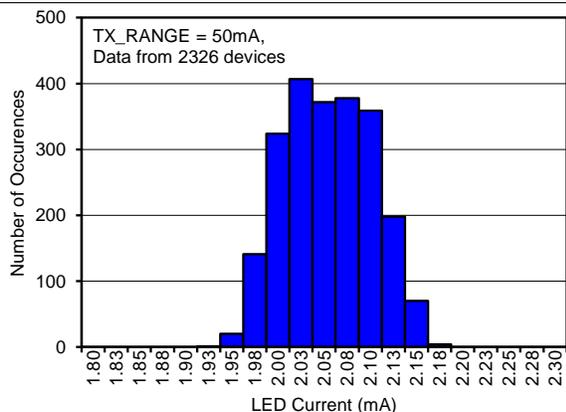


Figure 23. LED Current with Tx DAC Setting = 10 (2 mA)

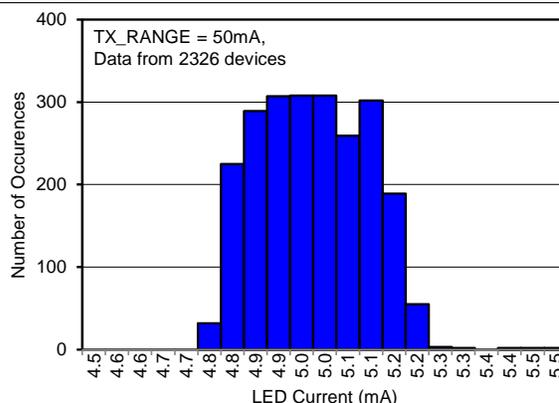


Figure 24. LED Current with Tx DAC Setting = 25 (5 mA)

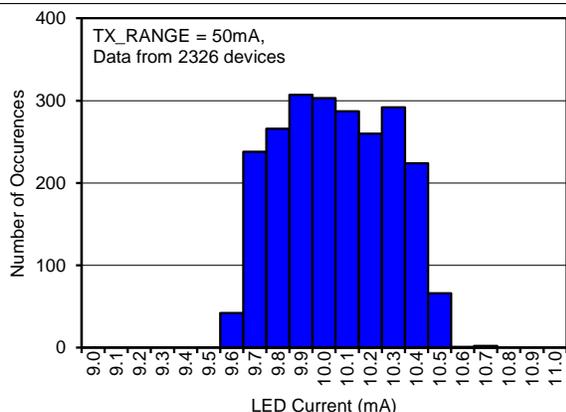


Figure 25. LED Current with Tx DAC Setting = 51 (10 mA)

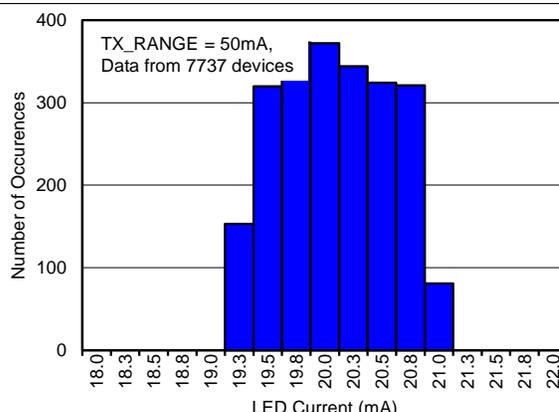


Figure 26. LED Current with Tx DAC Setting = 102 (20 mA)

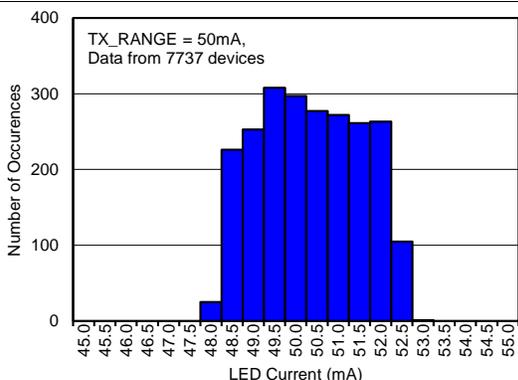


Figure 27. LED Current with Tx DAC Setting = 255 (50 mA)

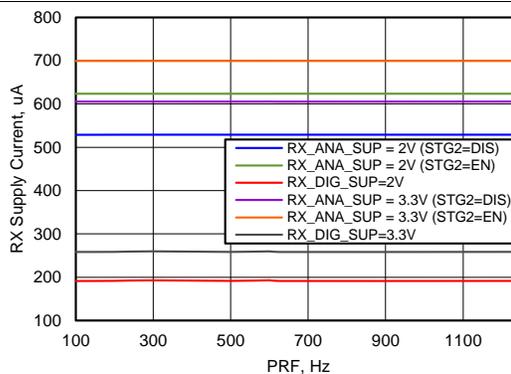


Figure 28. Receiver Supplies vs PRF

Typical Characteristics (continued)

Minimum and maximum specifications are at  $T_A = 0^\circ\text{C}$  to  $70^\circ\text{C}$ . Typical specifications are at  $T_A = 25^\circ\text{C}$ ,  $RX\_ANA\_SUP = RX\_DIG\_SUP = 3.0\text{ V}$ ,  $TX\_CTRL\_SUP = LED\_DRV\_SUP = 3.3\text{ V}$ , and  $f_{CLK} = 8\text{ MHz}$ , unless otherwise noted.

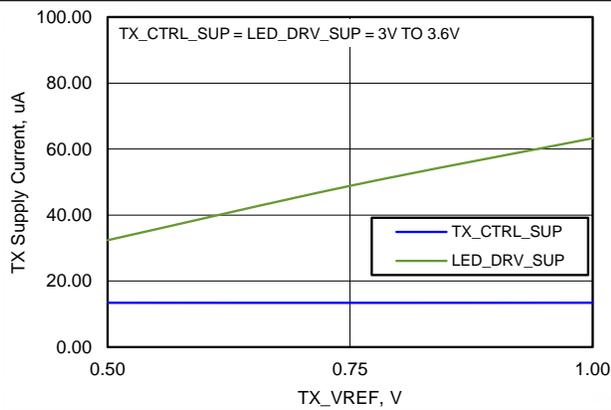


Figure 29. Transmitter Supplies vs TX\_REF

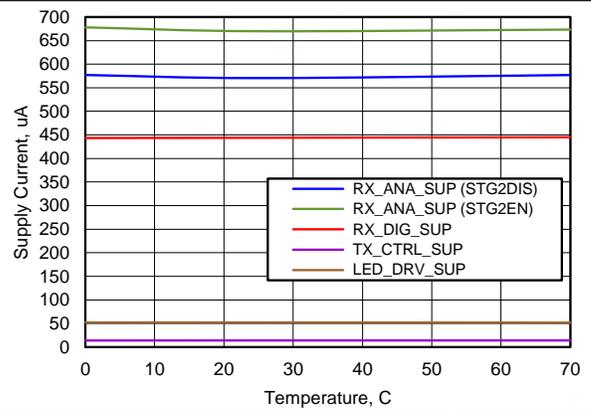


Figure 30. Power Supplies vs Temperature

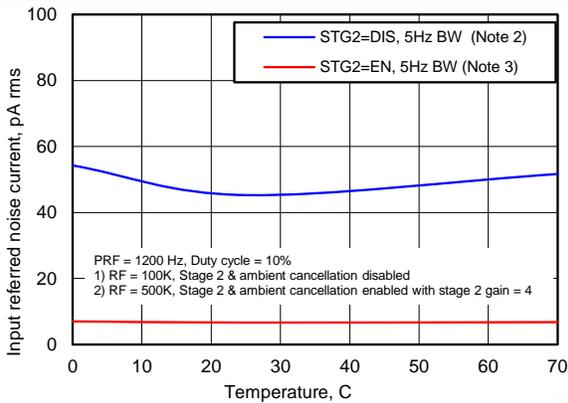


Figure 31. Input-Referred Noise vs Temperature

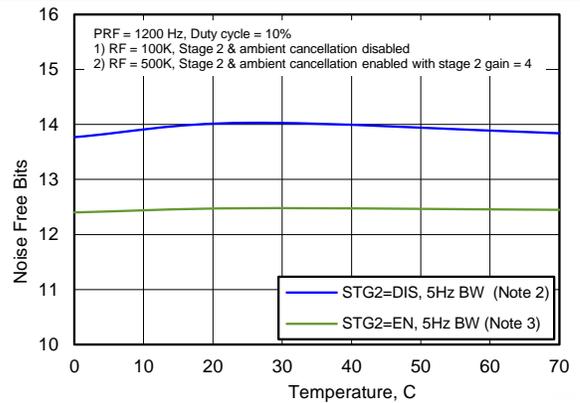


Figure 32. Noise-Free Bits vs Temperature

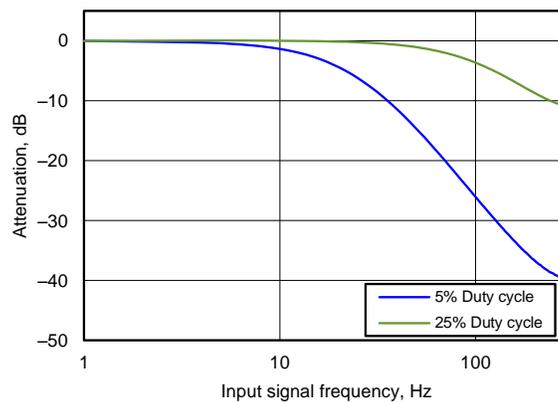


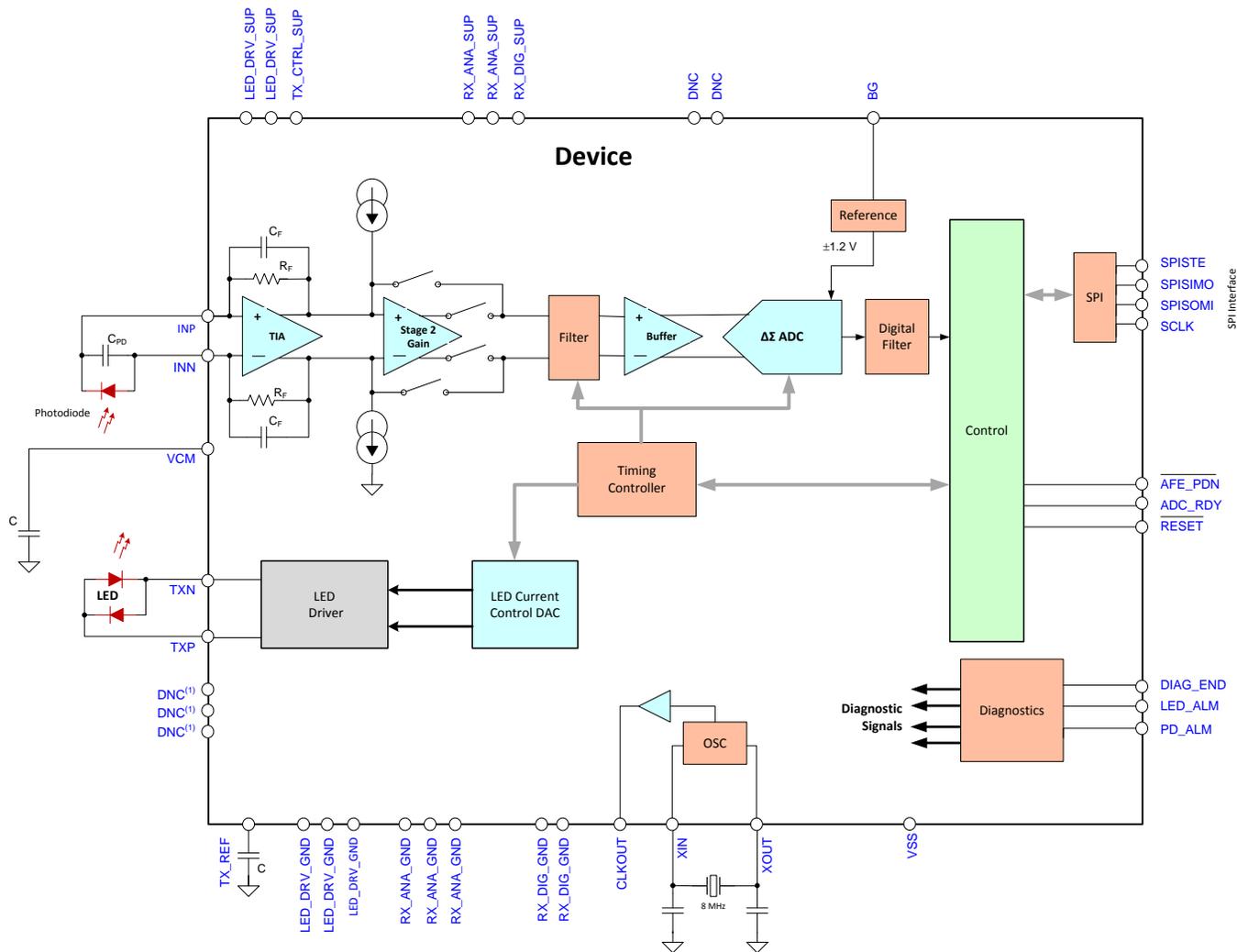
Figure 33. Filter Response vs Duty Cycle

## 8 Detailed Description

### 8.1 Overview

The AFE4400 is a complete analog front-end (AFE) solution targeted for pulse oximeter applications. The device consists of a low-noise receiver channel, an LED transmit section, and diagnostics for sensor and LED fault detection. To ease clocking requirements and provide the low-jitter clock to the AFE, an oscillator is also integrated that functions from an external crystal. The device communicates to an external microcontroller or host processor using an SPI interface. The *Functional Block Diagram* section provides a detailed block diagram for the AFE4400. The blocks are described in more detail in the following sections.

### 8.2 Functional Block Diagram



### 8.3 Feature Description

#### 8.3.1 Receiver Channel

This section describes the functionality of the receiver channel.

##### 8.3.1.1 Receiver Front-End

The receiver consists of a differential current-to-voltage (I-V) transimpedance amplifier (TIA) that converts the input photodiode current into an appropriate voltage, as shown in Figure 34. The feedback resistor of the amplifier ( $R_F$ ) is programmable to support a wide range of photodiode currents. Available  $R_F$  values include: 1 M $\Omega$ , 500 k $\Omega$ , 250 k $\Omega$ , 100 k $\Omega$ , 50 k $\Omega$ , 25 k $\Omega$ , and 10 k $\Omega$ .

The device is ideally suited as a front-end for a PPG (photoplethysmography) application. In such an application, the light from the LED is reflected (or transmitted) from (or through) the various components inside the body (such as blood, tissue, and so forth) and are received by the photodiode. The signal received by the photodiode has three distinct components:

1. A pulsatile or ac component that arises as a result of the changes in blood volume through the arteries.
2. A constant dc signal that is reflected or transmitted from the time invariant components in the path of light. This constant dc component is referred to as the pleth signal.
3. Ambient light entering the photodiode.

The ac component is usually a small fraction of the pleth component, with the ratio referred to as the perfusion index (PI). Thus, the allowed signal chain gain is usually determined by the amplitude of the dc component.

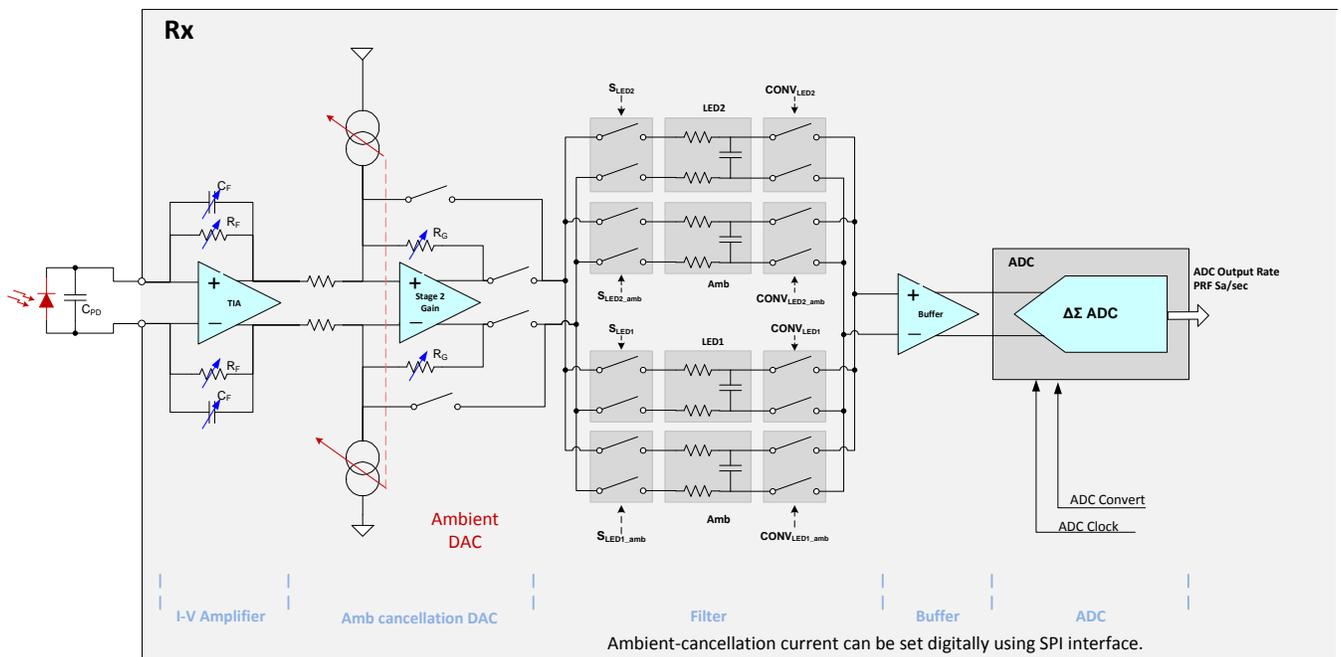


Figure 34. Receiver Front-End

The  $R_F$  amplifier and the feedback capacitor ( $C_F$ ) form a low-pass filter for the input signal current. Always ensure that the low-pass filter RC time constant has sufficiently high bandwidth (as shown by Equation 1) because the input current consists of pulses. For this reason, the feedback capacitor is also programmable. Available  $C_F$  values include: 5 pF, 10 pF, 25 pF, 50 pF, 100 pF, and 250 pF. Any combination of these capacitors can also be used.

$$R_F \times C_F \leq \frac{\text{Rx Sample Time}}{10} \tag{1}$$

## Feature Description (continued)

The output voltage of the I-V amplifier includes the pleth component (the desired signal) and a component resulting from the ambient light leakage. The I-V amplifier is followed by the second stage, which consists of a current digital-to-analog converter (DAC) that sources the cancellation current and an amplifier that gains up the pleth component alone. The amplifier has five programmable gain settings: 0 dB, 3.5 dB, 6 dB, 9.5 dB, and 12 dB. The gained-up pleth signal is then low-pass filtered (500-Hz bandwidth) and buffered before driving a 22-bit ADC. The current DAC has a cancellation current range of 10  $\mu$ A with 10 steps (1  $\mu$ A each). The DAC value can be digitally specified with the SPI interface. Using ambient compensation with the ambient DAC allows the dc-biased signal to be centered to near mid-point of the amplifier ( $\pm 0.9$  V). Using the gain of the second stage allows for more of the available ADC dynamic range to be used.

The output of the ambient cancellation amplifier is separated into LED2 and LED1 channels. When LED2 is on, the amplifier output is filtered and sampled on capacitor  $C_{LED2}$ . Similarly, the LED1 signal is sampled on the  $C_{LED1}$  capacitor when LED1 is on. In between the LED2 and LED1 pulses, the idle amplifier output is sampled to estimate the ambient signal on capacitors  $C_{LED2\_amb}$  and  $C_{LED1\_amb}$ .

The sampling duration is termed the *Rx sample time* and is programmable for each signal, independently. The sampling can start after the I-V amplifier output is stable (to account for LED and cable settling times). The Rx sample time is used for all dynamic range calculations; the minimum time recommended is 50  $\mu$ s. While the AFE4400 can support pulse widths lower than 50  $\mu$ s, having too low a pulse width could result in a degraded signal and noise from the photodiode.

A single, 22-bit ADC converts the sampled LED2, LED1, and ambient signals sequentially. Each conversion provides a single digital code at the ADC output. As discussed in the [Receiver Timing](#) section, the conversions are meant to be staggered so that the LED2 conversion starts after the end of the LED2 sample phase, and so on.

Note that four data streams are available at the ADC output (LED2, LED1, ambient LED2, and ambient LED1) at the same rate as the pulse repetition frequency. The ADC is followed by a digital ambient subtraction block that additionally outputs the (LED2 – ambient LED2) and (LED1 – ambient LED1) data values.

Feature Description (continued)

8.3.1.2 Ambient Cancellation Scheme and Second Stage Gain Block

The receiver provides digital samples corresponding to ambient duration. The host processor (external to the AFE) can use these ambient values to estimate the amount of ambient light leakage. The processor must then set the value of the ambient cancellation DAC using the SPI, as shown in Figure 35.

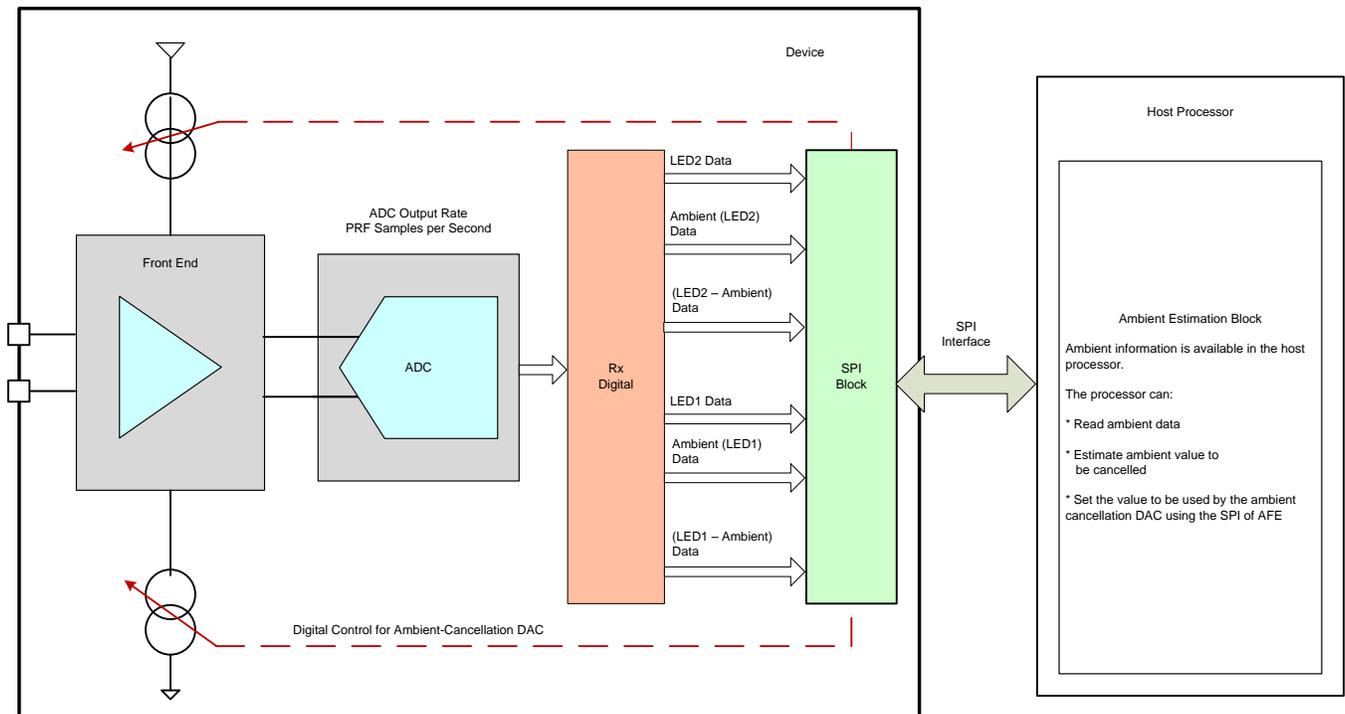
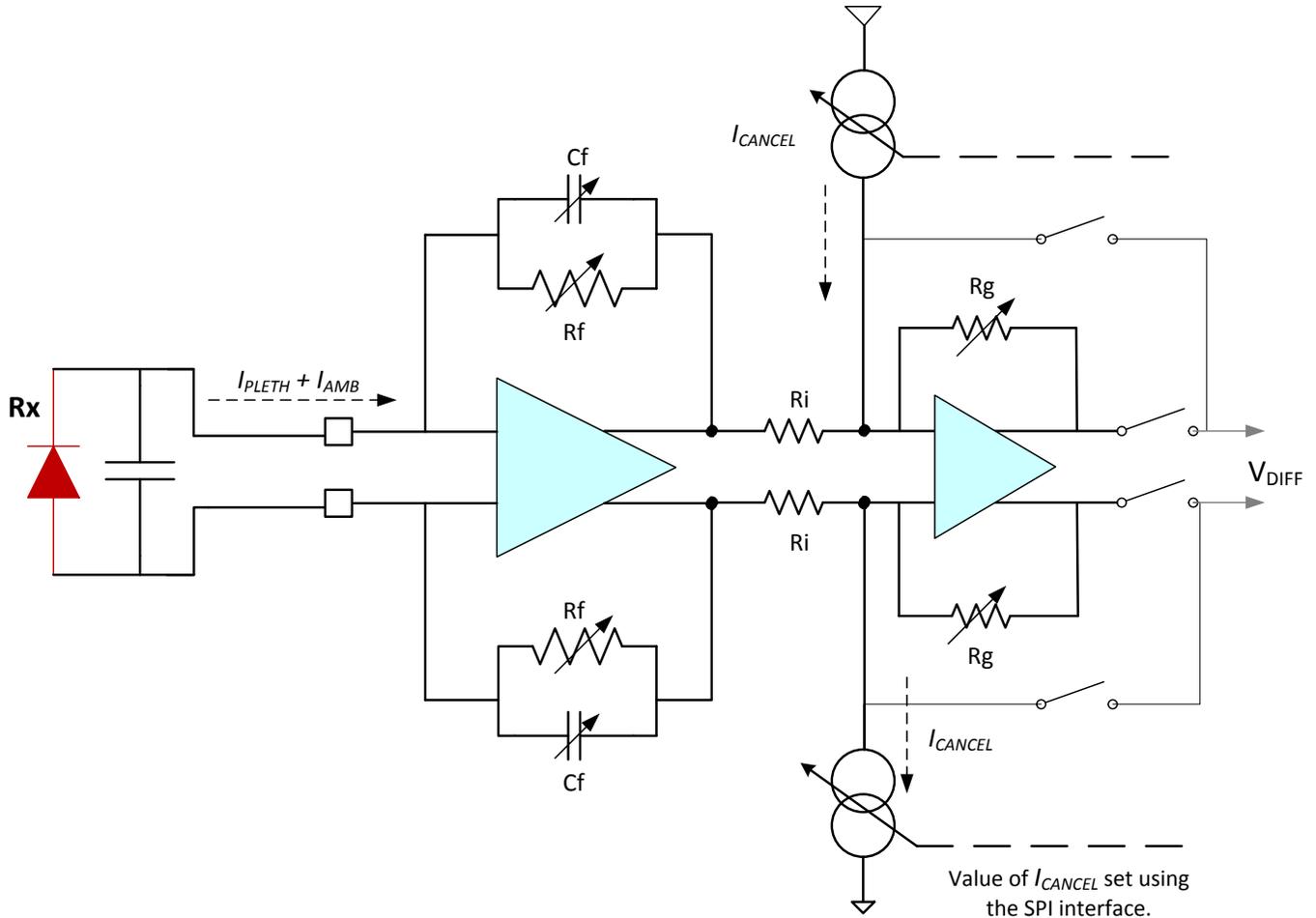


Figure 35. Ambient Cancellation Loop (Closed by the Host Processor)

**Feature Description (continued)**

Using the set value, the ambient cancellation stage subtracts the ambient component and gains up only the pleth component of the received signal; see Figure 36. The amplifier gain is programmable to 0 dB, 3.5 dB, 6 dB, 9.5 dB, and 12 dB.



**Figure 36. Front-End (I-V Amplifier and Cancellation Stage)**

## Feature Description (continued)

The differential output of the second stage is  $V_{DIFF}$ , as given by [Equation 2](#):

$$V_{DIFF} = 2 \times \left[ I_{PLETH} \times \frac{R_F}{R_I} + I_{AMB} \times \frac{R_F}{R_I} - I_{CANCEL} \right] \times R_G$$

where:

- $R_I = 100 \text{ k}\Omega$ ,
- $I_{PLETH}$  = photodiode current pleth component,
- $I_{AMB}$  = photodiode current ambient component, and
- $I_{CANCEL}$  = the cancellation current DAC value (as estimated by the host processor). (2)

$R_G$  values with various gain settings are listed in [Table 1](#).

**Table 1.  $R_G$  Values**

GAIN	$R_G(\text{k}\Omega)$
0 (x1)	100
3.5 (x1.5)	150
6 (x2)	200
9.5 (x3)	300
12 (x4)	400

### 8.3.1.3 Receiver Control Signals

**LED2 sample phase ( $S_{LED2}$  or  $S_R$ ):** When this signal is high, the amplifier output corresponds to the LED2 on-time. The amplifier output is filtered and sampled into capacitor  $C_{LED2}$ . To avoid settling effects resulting from the LED or cable, program  $S_{LED2}$  to start after the LED turns on. This settling delay is programmable.

**Ambient sample phase ( $S_{LED2\_amb}$  or  $S_{R\_amb}$ ):** When this signal is high, the amplifier output corresponds to the LED2 off-time and can be used to estimate the ambient signal (for the LED2 phase). The amplifier output is filtered and sampled into capacitor  $C_{LED2\_amb}$ .

**LED1 sample phase ( $S_{LED1}$  or  $S_{IR}$ ):** When this signal is high, the amplifier output corresponds to the LED1 on-time. The amplifier output is filtered and sampled into capacitor  $C_{LED1}$ . To avoid settling effects resulting from the LED or cable, program  $S_{LED1}$  to start after the LED turns on. This settling delay is programmable.

**Ambient sample phase ( $S_{LED1\_amb}$  or  $S_{IR\_amb}$ ):** When this signal is high, the amplifier output corresponds to the LED1 off-time and can be used to estimate the ambient signal (for the LED1 phase). The amplifier output is filtered and sampled into capacitor  $C_{LED1\_amb}$ .

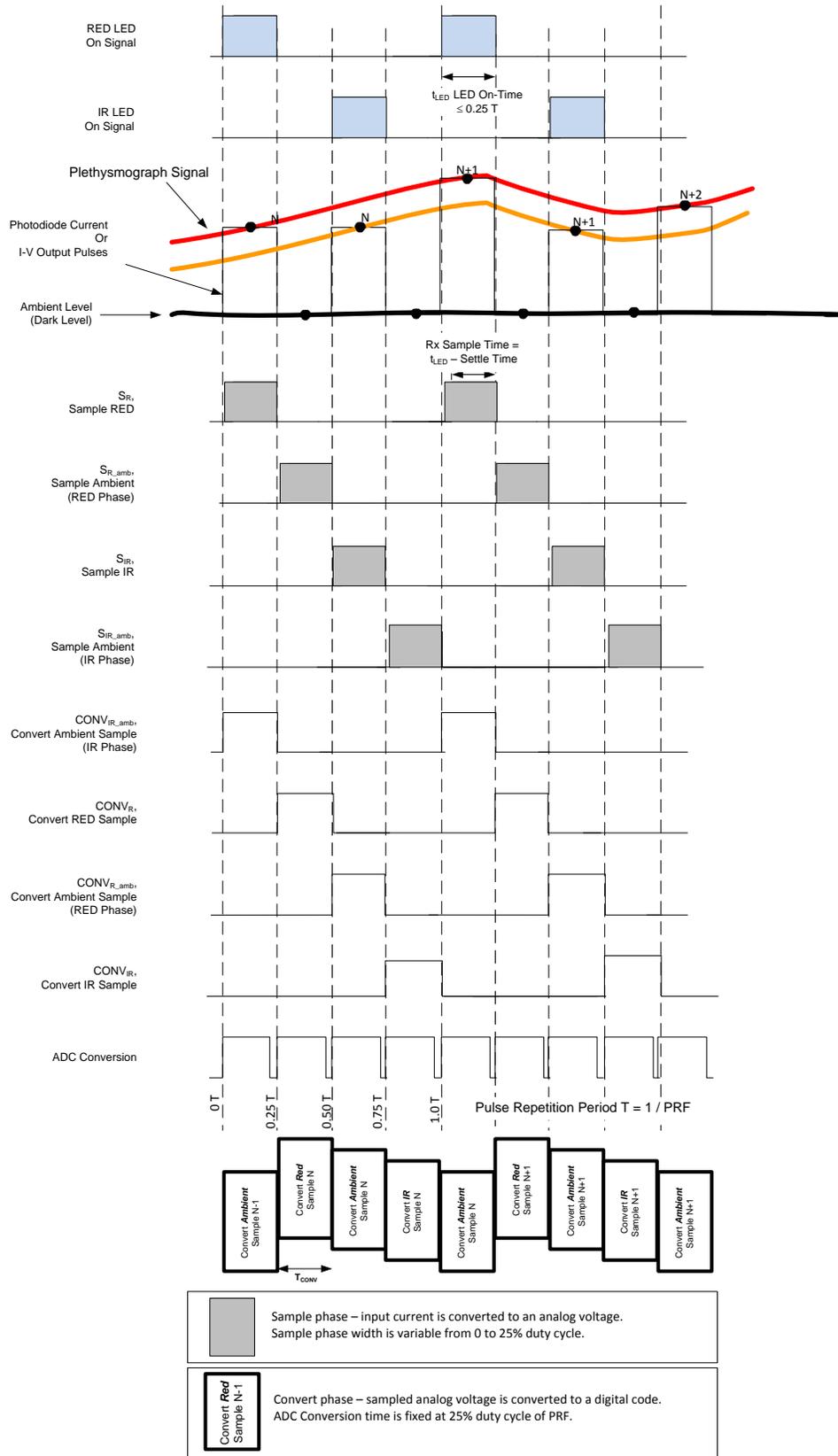
**LED2 convert phase ( $CONV_{LED2}$  or  $CONV_R$ ):** When this signal is high, the voltage sampled on  $C_{LED2}$  is buffered and applied to the ADC for conversion. At the end of the conversion, the ADC provides a single digital code corresponding to the LED2 sample.

**Ambient convert phases ( $CONV_{LED2\_amb}$  or  $CONV_{R\_amb}$ ,  $CONV_{LED1\_amb}$  or  $CONV_{IR\_amb}$ ):** When this signal is high, the voltage sampled on  $C_{LED2\_amb}$  (or  $C_{LED1\_amb}$ ) is buffered and applied to the ADC for conversion. At the end of the conversion, the ADC provides a single digital code corresponding to the ambient sample.

**LED1 convert phase ( $CONV_{LED1}$  or  $CONV_{IR}$ ):** When this signal is high, the voltage sampled on  $C_{LED1}$  is buffered and applied to the ADC for conversion. At the end of the conversion, the ADC provides a single digital code corresponding to the LED1 sample.

### 8.3.1.4 Receiver Timing

See [Figure 37](#) for a timing diagram detailing the control signals related to the LED on-time, Rx sample time, and the ADC conversion times for each channel. [Figure 37](#) shows the timing for a case where each phase occupies 25% of the pulse repetition period. However, this percentage is not a requirement. In cases where the device is operated with low pulse repetition frequency (PRF) or low LED pulse durations, the active portion of the pulse repetition period can be reduced. Using the dynamic power-down feature, the overall power consumption can be significantly reduced.

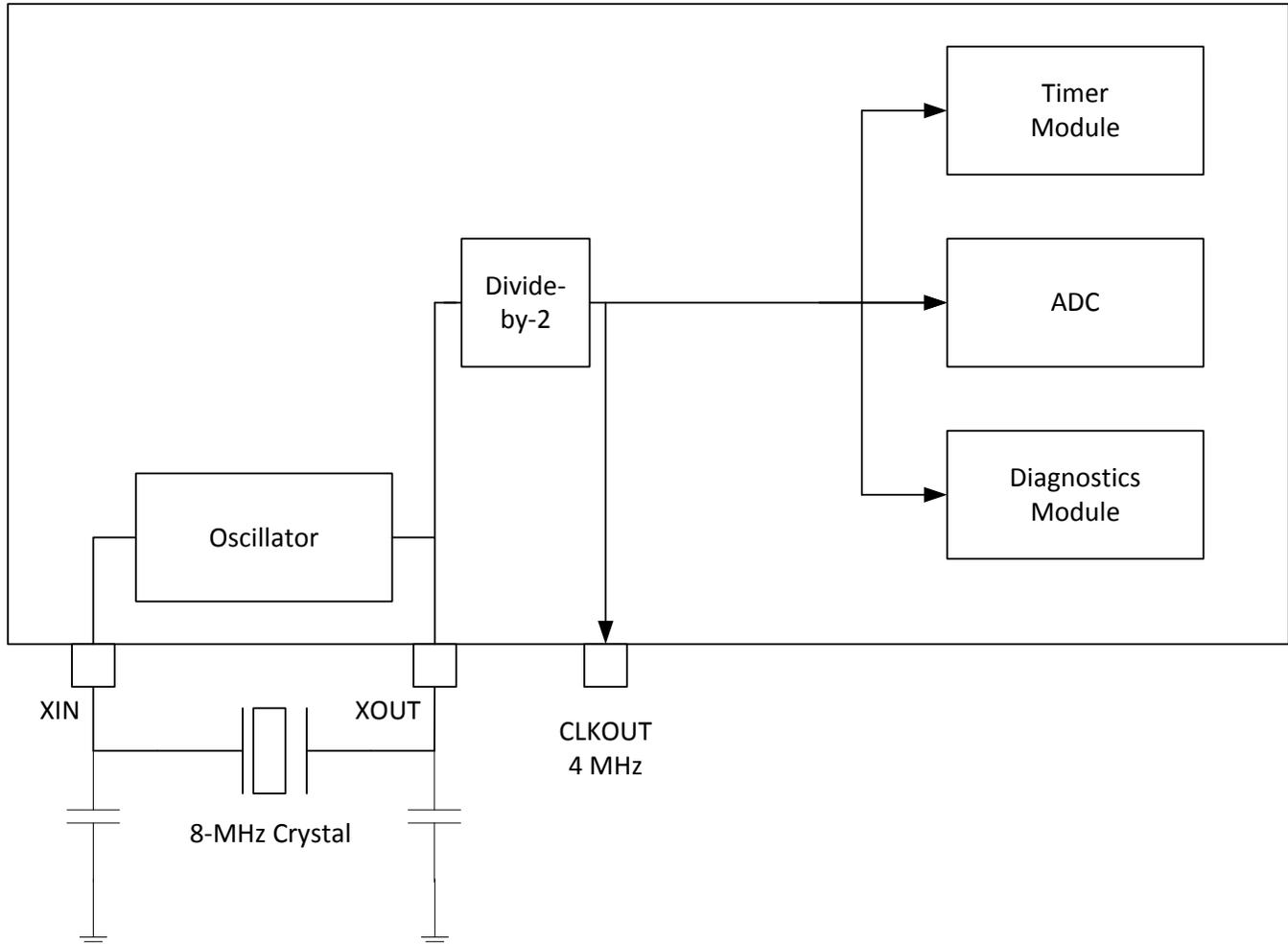


NOTE: Relationship to the AFE4400 EVM is: LED1 = IR and LED2 = RED.

Figure 37. Rx Timing Diagram

### 8.3.2 Clocking and Timing Signal Generation

The crystal oscillator generates a master clock signal using an external crystal. In the default mode, a divide-by-2 block converts the 8-MHz clock to 4 MHz, which is used by the AFE to operate the timer modules, ADC, and diagnostics. The 4-MHz clock is buffered and output from the AFE in order to clock an external microcontroller. The clocking functionality is shown in [Figure 38](#).

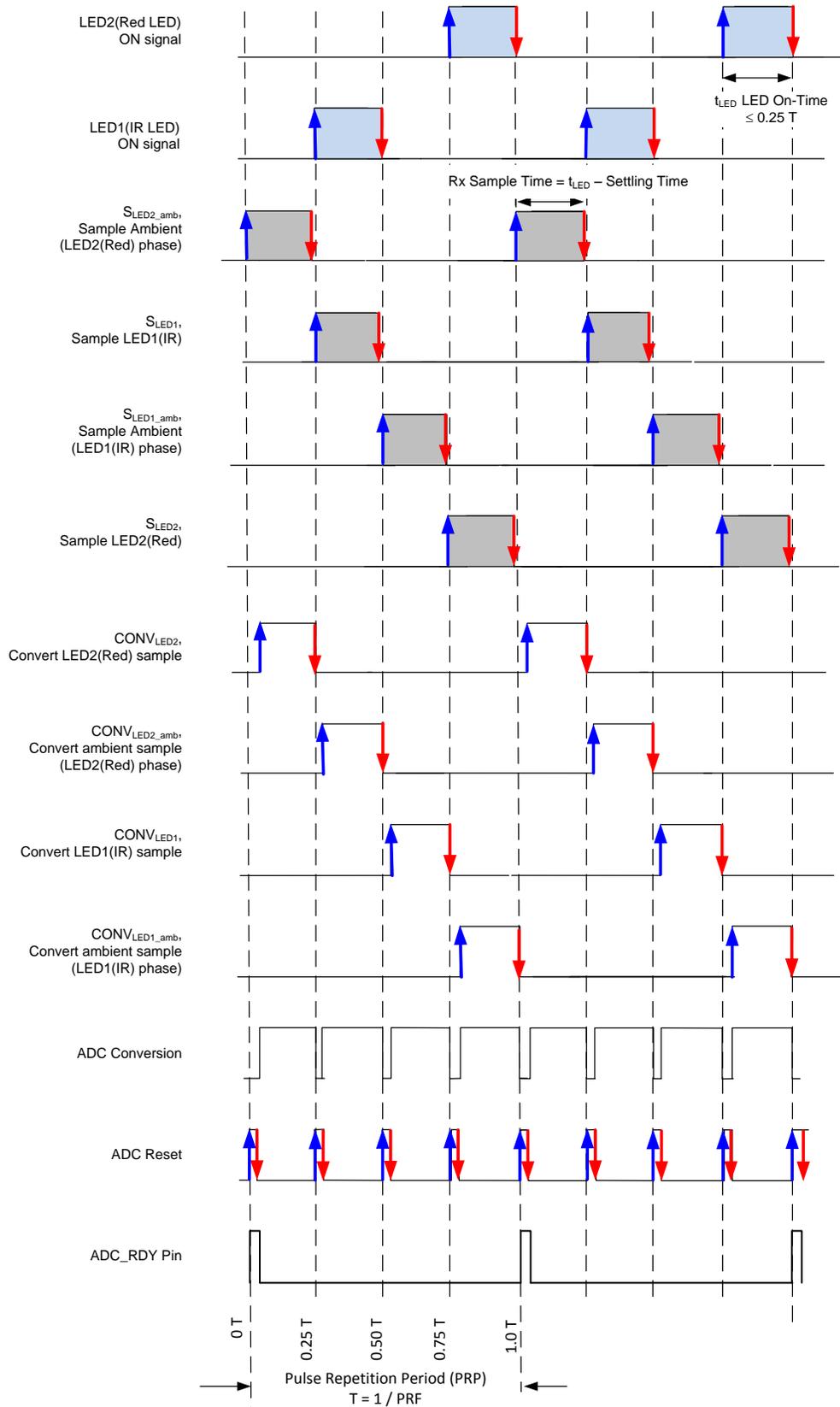


**Figure 38. AFE Clocking**

### 8.3.3 Timer Module

See [Figure 39](#) for a timing diagram detailing the various timing edges that are programmable using the timer module. The rising and falling edge positions of 11 signals can be controlled. The module uses a single 16-bit counter (running off of the 4-MHz clock) to set the time-base.

All timing signals are set with reference to the pulse repetition period (PRP). Therefore, a dedicated compare register compares the 16-bit counter value with the reference value specified in the PRF register. Every time that the 16-bit counter value is equal to the reference value in the PRF register, the counter is reset to 0.



NOTE: Programmable edges are shown in blue and red.

Figure 39. AFE Control Signals

For the timing signals in Figure 37, the start and stop edge positions are programmable with respect to the PRF period. Each signal uses a separate timer compare module that compares the counter value with preprogrammed reference values for the start and stop edges. All reference values can be set using the SPI interface.

After the counter value has exceeded the stop reference value, the output signal is set. When the counter value equals the stop reference value, the output signal is reset. Figure 40 shows a diagram of the timer compare register. With a 4-MHz clock, the edge placement resolution is 0.25  $\mu$ s.

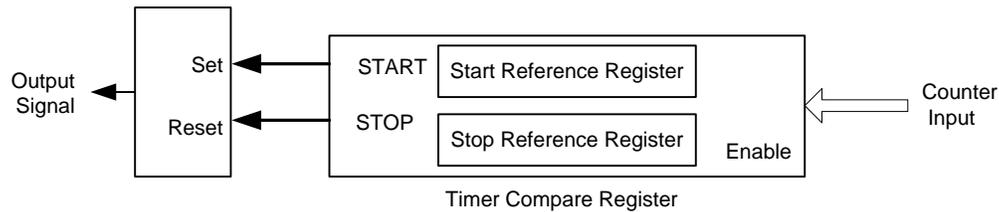


Figure 40. Compare Register

The ADC conversion signal requires four pulses in each PRF clock period. Timer compare register 11 uses four sets of start and stop registers to control the ADC conversion signal, as shown in Figure 41.

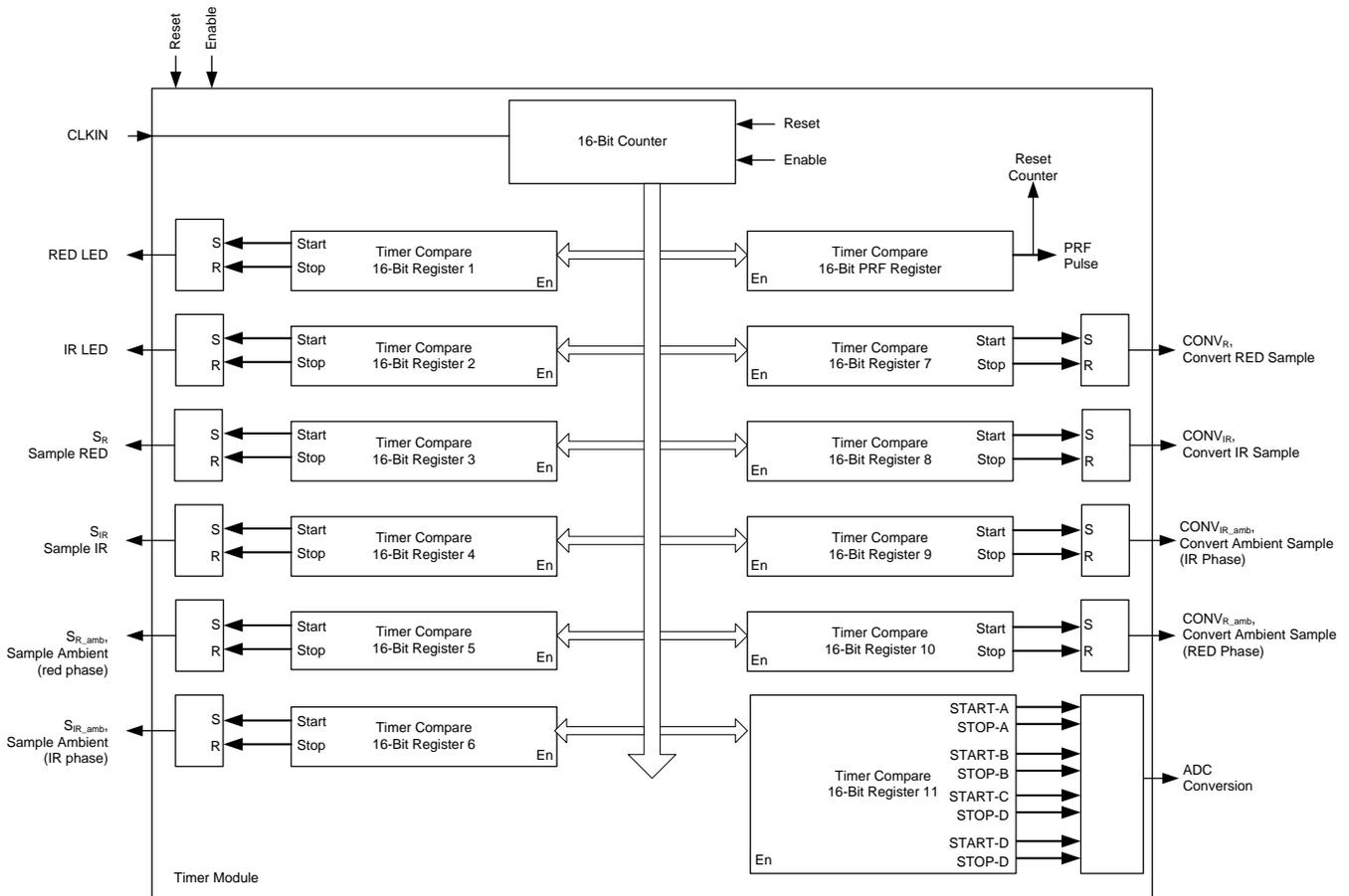


Figure 41. Timer Module

### 8.3.3.1 Using the Timer Module

The timer module registers can be used to program the start and end instants in units of 4-MHz clock cycles. These timing instants and the corresponding registers are listed in [Table 2](#).

Note that the device does not restrict the values in these registers; thus, the start and end edges can be positioned anywhere within the pulse repetition period. Care must be taken by the user to program suitable values in these registers to avoid overlapping the signals and to make sure none of the edges exceed the value programmed in the PRP register. Writing the same value in the start and end registers results in a pulse duration of one clock cycle. The following steps describe the timer sequencing configuration:

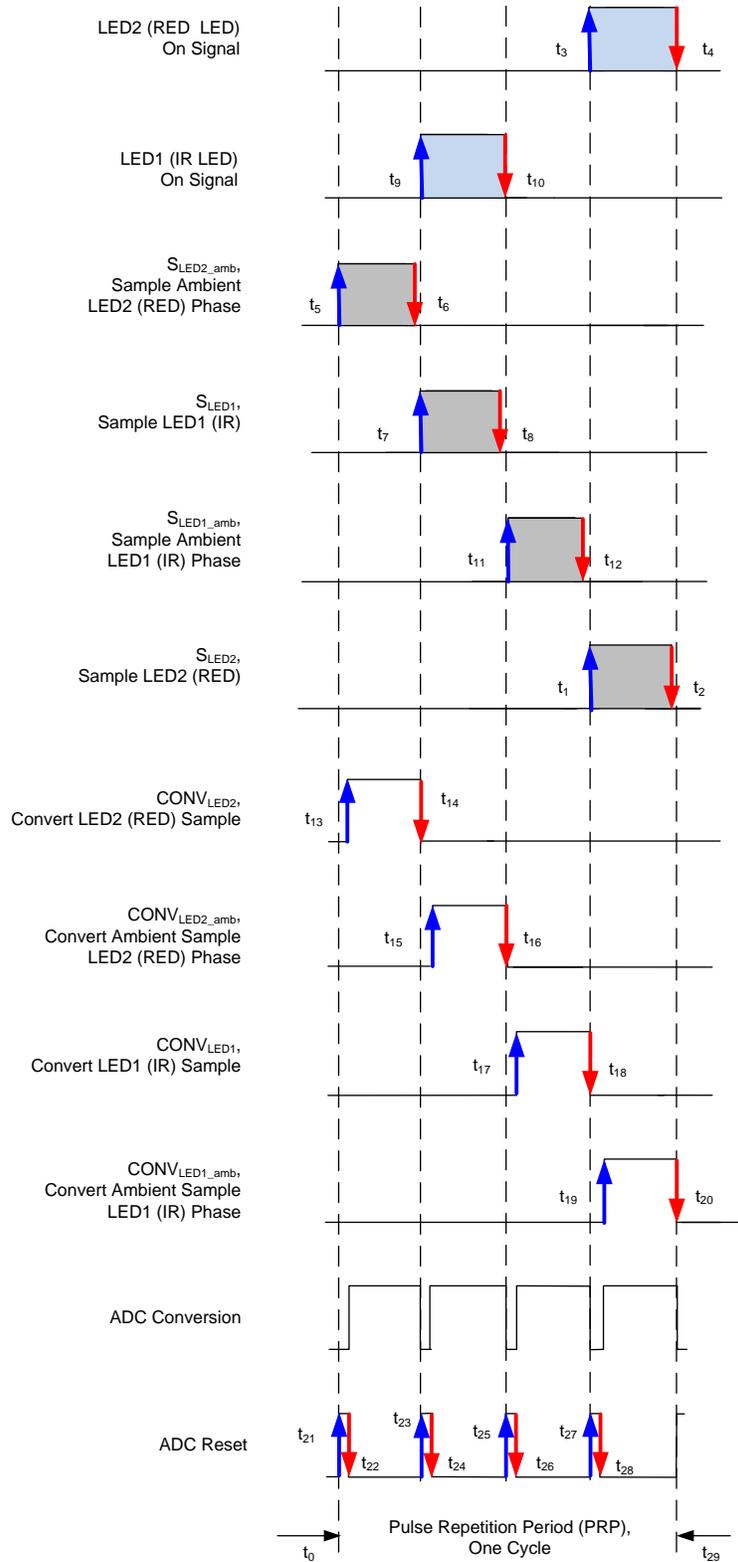
1. With respect to the start of the PRP period (indicated by timing instant  $t_0$  in [Figure 42](#)), the following sequence of conversions must be followed in order: convert LED2 → LED2 ambient → LED1 → LED1 ambient.
2. Also, starting from  $t_0$ , the sequence of sampling instants must be staggered with respect to the respective conversions as follows: sample LED2 ambient → LED1 → LED1 ambient → LED2.
3. Finally, align the edges for the two LED pulses with the respective sampling instants.

**Table 2. Clock Edge Mapping to SPI Registers**

TIME INSTANT (See <a href="#">Figure 42</a> and <a href="#">Figure 43</a> )	DESCRIPTION	CORRESPONDING REGISTER ADDRESS AND REGISTER BITS	EXAMPLE <sup>(1)</sup> (Decimal)
$t_0$	Start of pulse repetition period	No register control	—
$t_1$	Start of sample LED2 pulse	LED2STC[15:0], register 01h	6050
$t_2$	End of sample LED2 pulse	LED2ENDC[15:0], register 02h	7998
$t_3$	Start of LED2 pulse	LED2LEDSTC[15:0], register 03h	6000
$t_4$	End of LED2 pulse	LED2LEDENDC[15:0], register 04h	7999
$t_5$	Start of sample LED2 ambient pulse	ALED2STC[15:0], register 05h	50
$t_6$	End of sample LED2 ambient pulse	ALED2ENDC[15:0], register 06h	1998
$t_7$	Start of sample LED1 pulse	LED1STC[15:0], register 07h	2050
$t_8$	End of sample LED1 pulse	LED1ENDC[15:0], register 08h	3998
$t_9$	Start of LED1 pulse	LED1LEDSTC[15:0], register 09h	2000
$t_{10}$	End of LED1 pulse	LED1LEDENDC[15:0], register 0Ah	3999
$t_{11}$	Start of sample LED1 ambient pulse	ALED1STC[15:0], register 0Bh	4050
$t_{12}$	End of sample LED1 ambient pulse	ALED1ENDC[15:0], register 0Ch	5998
$t_{13}$	Start of convert LED2 pulse	LED2CONVST[15:0], register 0Dh Must start one AFE clock cycle after the ADC reset pulse ends.	4
$t_{14}$	End of convert LED2 pulse	LED2CONVEND[15:0], register 0Eh	1999
$t_{15}$	Start of convert LED2 ambient pulse	ALED2CONVST[15:0], register 0Fh Must start one AFE clock cycle after the ADC reset pulse ends.	2004
$t_{16}$	End of convert LED2 ambient pulse	ALED2CONVEND[15:0], register 10h	3999
$t_{17}$	Start of convert LED1 pulse	LED1CONVST[15:0], register 11h Must start one AFE clock cycle after the ADC reset pulse ends.	4004
$t_{18}$	End of convert LED1 pulse	LED1CONVEND[15:0], register 12h	5999
$t_{19}$	Start of convert LED1 ambient pulse	ALED1CONVST[15:0], register 13h Must start one AFE clock cycle after the ADC reset pulse ends.	6004
$t_{20}$	End of convert LED1 ambient pulse	ALED1CONVEND[15:0], register 14h	7999
$t_{21}$	Start of first ADC conversion reset pulse	ADCRSTSTCT0[15:0], register 15h	0
$t_{22}$	End of first ADC conversion reset pulse <sup>(2)</sup>	ADCRSTENDCT0[15:0], register 16h	3
$t_{23}$	Start of second ADC conversion reset pulse	ADCRSTSTCT1[15:0], register 17h	2000
$t_{24}$	End of second ADC conversion reset pulse <sup>(2)</sup>	ADCRSTENDCT1[15:0], register 18h	2003
$t_{25}$	Start of third ADC conversion reset pulse	ADCRSTSTCT2[15:0], register 19h	4000
$t_{26}$	End of third ADC conversion reset pulse <sup>(2)</sup>	ADCRSTENDCT2[15:0], register 1Ah	4003
$t_{27}$	Start of fourth ADC conversion reset pulse	ADCRSTSTCT3[15:0], register 1Bh	6000
$t_{28}$	End of fourth ADC conversion reset pulse <sup>(2)</sup>	ADCRSTENDCT3[15:0], register 1Ch	6003
$t_{29}$	End of pulse repetition period	PRPCOUNT[15:0], register 1Dh	7999

(1) Values are based off of a pulse repetition frequency (PRF) = 500 Hz and duty cycle = 25%.

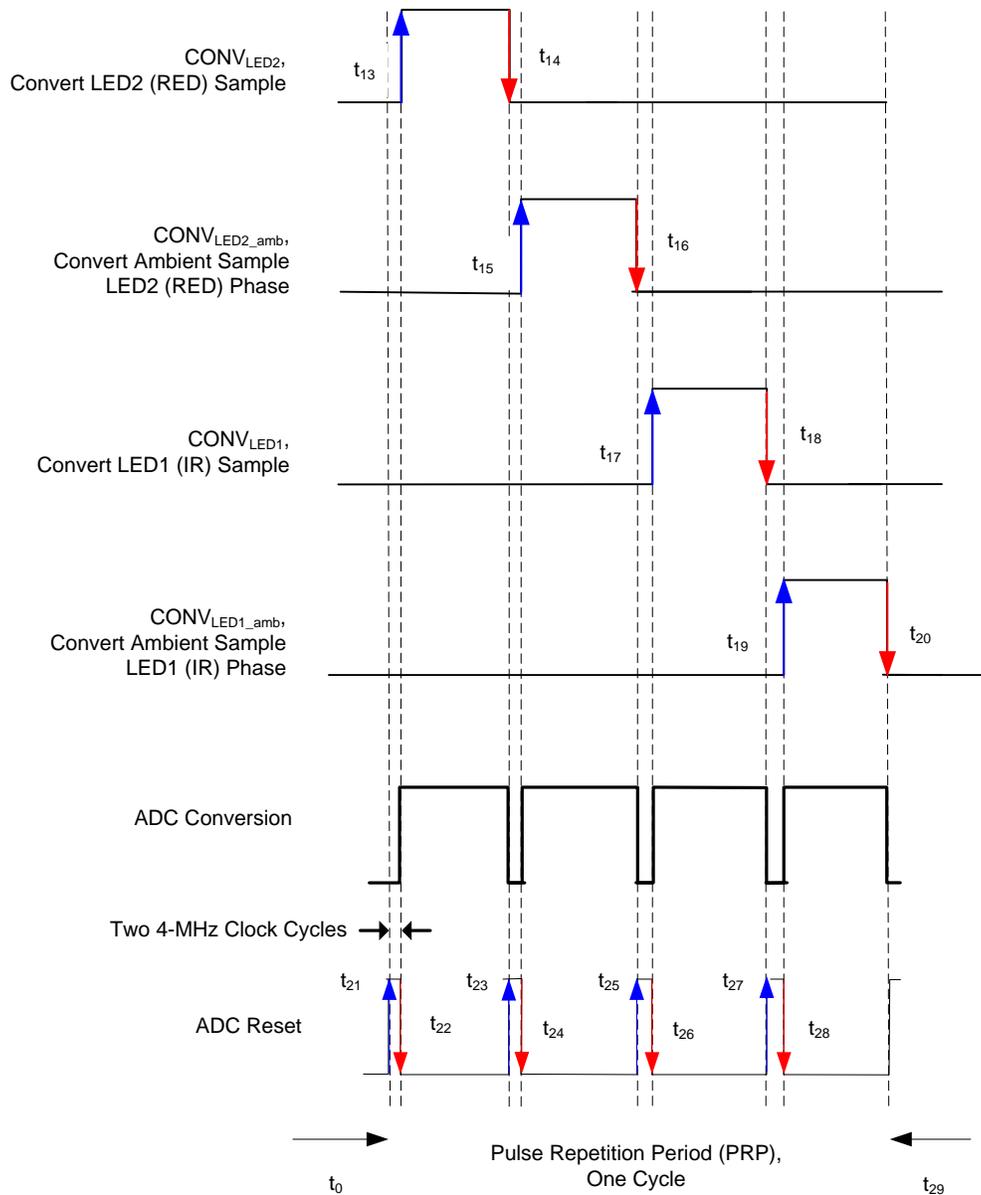
(2) See [Figure 43](#), note 2 for the effect of the ADC reset time crosstalk.



(1) RED = LED2, IR = LED1.

(2) A low ADC reset time can result in a small component of the LED signal leaking into the ambient phase. With an ADC reset of two clock cycles, a -60-dB leakage is expected. In many cases, this leakage does not affect system performance. However, if this crosstalk must be completely eliminated, a longer ADC reset time of approximately six clock cycles is recommended for  $t_{22}$ ,  $t_{24}$ ,  $t_{26}$ , and  $t_{28}$ .

Figure 42. Programmable Clock Edges<sup>(1)(2)</sup>



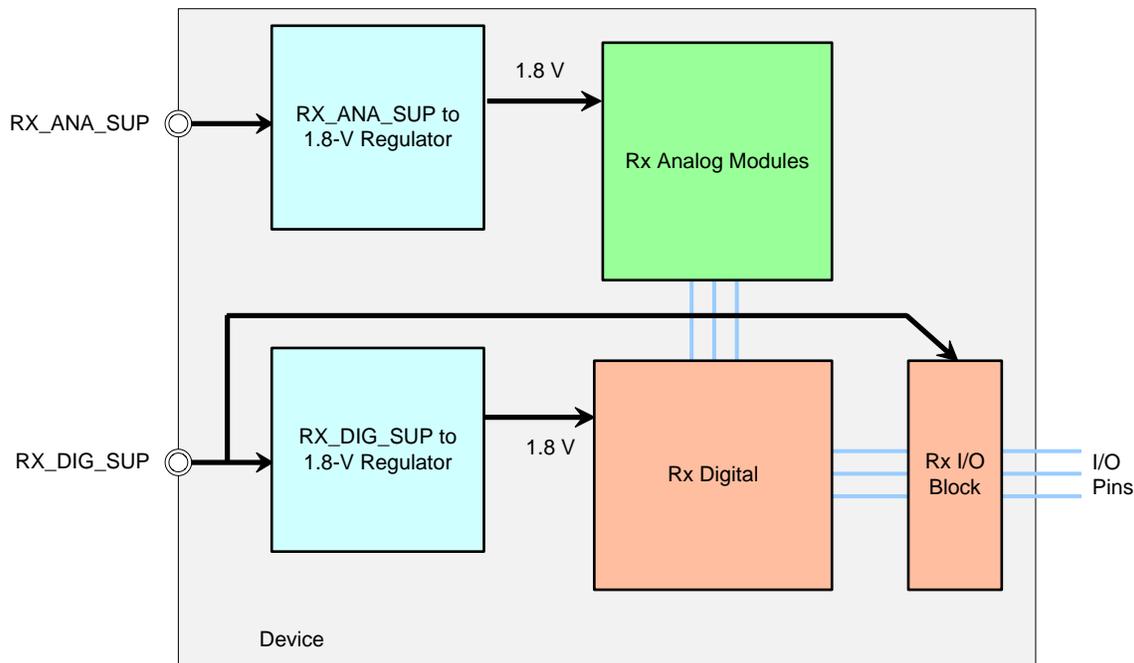
(1) RED = LED2, IR = LED1.

(2) A low ADC reset time can result in a small component of the LED signal leaking into the ambient phase. With an ADC reset of two clock cycles, a  $-60$ -dB leakage is expected. In many cases, this leakage does not affect system performance. However, if this crosstalk must be completely eliminated, a longer ADC reset time of approximately six clock cycles is recommended for  $t_{22}$ ,  $t_{24}$ ,  $t_{26}$ , and  $t_{28}$ .

**Figure 43. Relationship Between the ADC Reset and ADC Conversion Signals<sup>(1)(2)</sup>**

### 8.3.4 Receiver Subsystem Power Path

The block diagram in [Figure 44](#) shows the AFE4400 Rx subsystem power routing. Internal LDOs running off RX\_ANA\_SUP and RX\_DIG\_SUP generate the 1.8-V supplies required to drive the internal blocks. The two receive supplies could be shorted to a single supply on the board.



**Figure 44. Receive Subsystem Power Routing**

### 8.3.5 Transmit Section

The transmit section integrates the LED driver and the LED current control section with 8-bit resolution. This integration is designed to meet a dynamic range of better than 105 dB (based on a 1-sigma LED current noise).

The RED and IR LED reference currents can be independently set. The current source ( $I_{LED}$ ) locally regulates and ensures that the actual LED current tracks the specified reference. The transmitter section uses an internal 0.5-V reference voltage for operation. This reference voltage is available on the REF\_TX pin and must be decoupled to ground with a 2.2- $\mu$ F capacitor. The TX\_REF voltage is derived from the TX\_CTRL\_SUP. The maximum LED current setting supports up to 50-mA LED current.

Note that reducing the value of the band-gap reference capacitor on pin 7 reduces the time required for the device to wake-up and settle. However, this reduction in time is a trade-off between wake-up time and noise performance.

The minimum LED\_DRV\_SUP voltage required for operation depends on:

- Voltage drop across the LED ( $V_{LED}$ ),
- Voltage drop across the external cable, connector, and any other component in series with the LED ( $V_{CABLE}$ ), and
- Transmitter reference voltage.

Using the internal 0.5-V reference voltage, the minimum LED\_DRV\_SUP voltage can be as low as 3.0 V, provided that  $[3.0 \text{ V} - (V_{LED} + V_{CABLE}) > 1.4 \text{ V}]$  is met.

See the [Recommended Operating Conditions](#) table for further details.

Two LED driver schemes are supported:

- An H-bridge drive for a two-terminal back-to-back LED package; see [Figure 45](#).
- A push-pull drive for a three-terminal LED package; see [Figure 46](#).

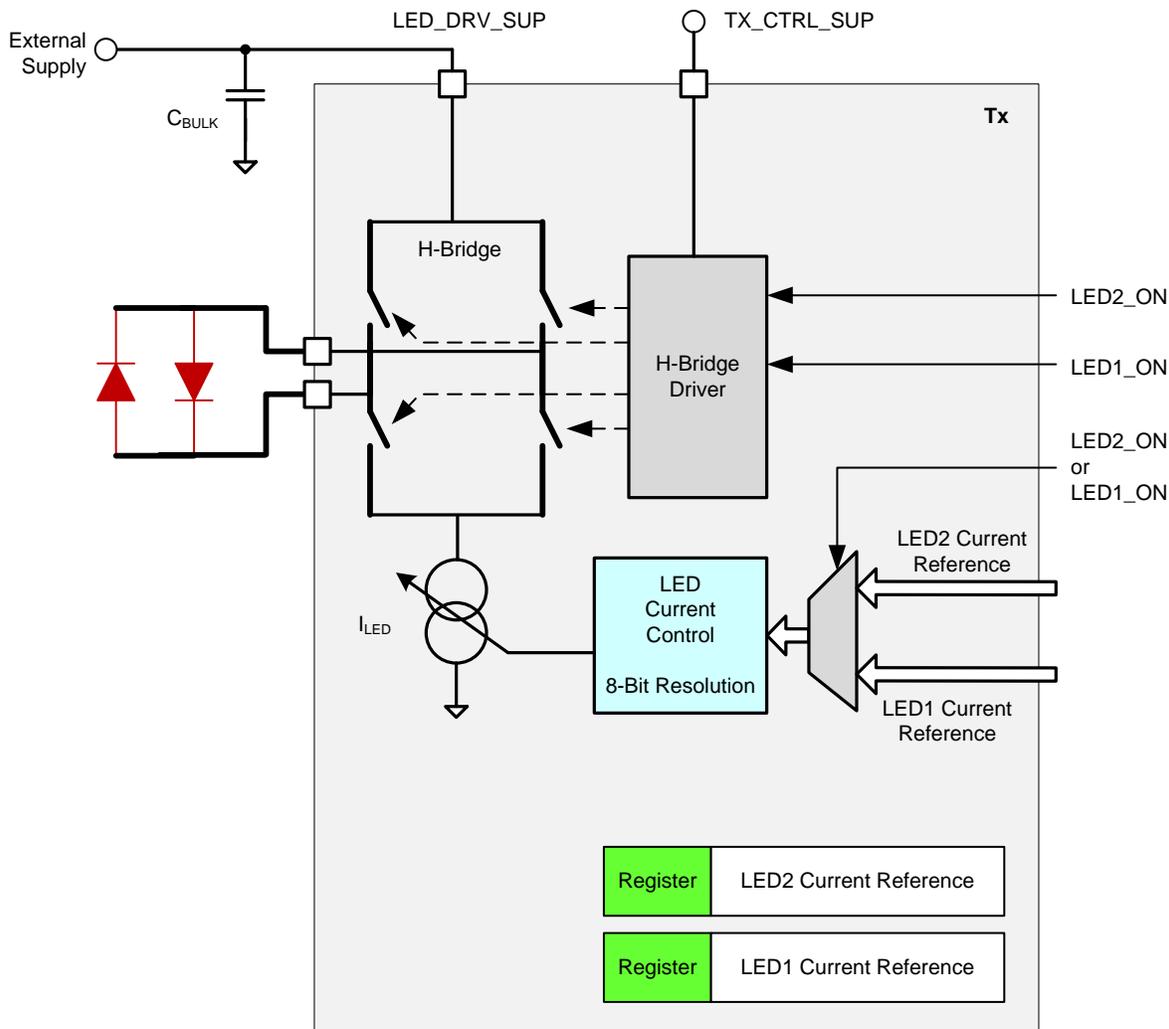


Figure 45. Transmit: H-Bridge Drive

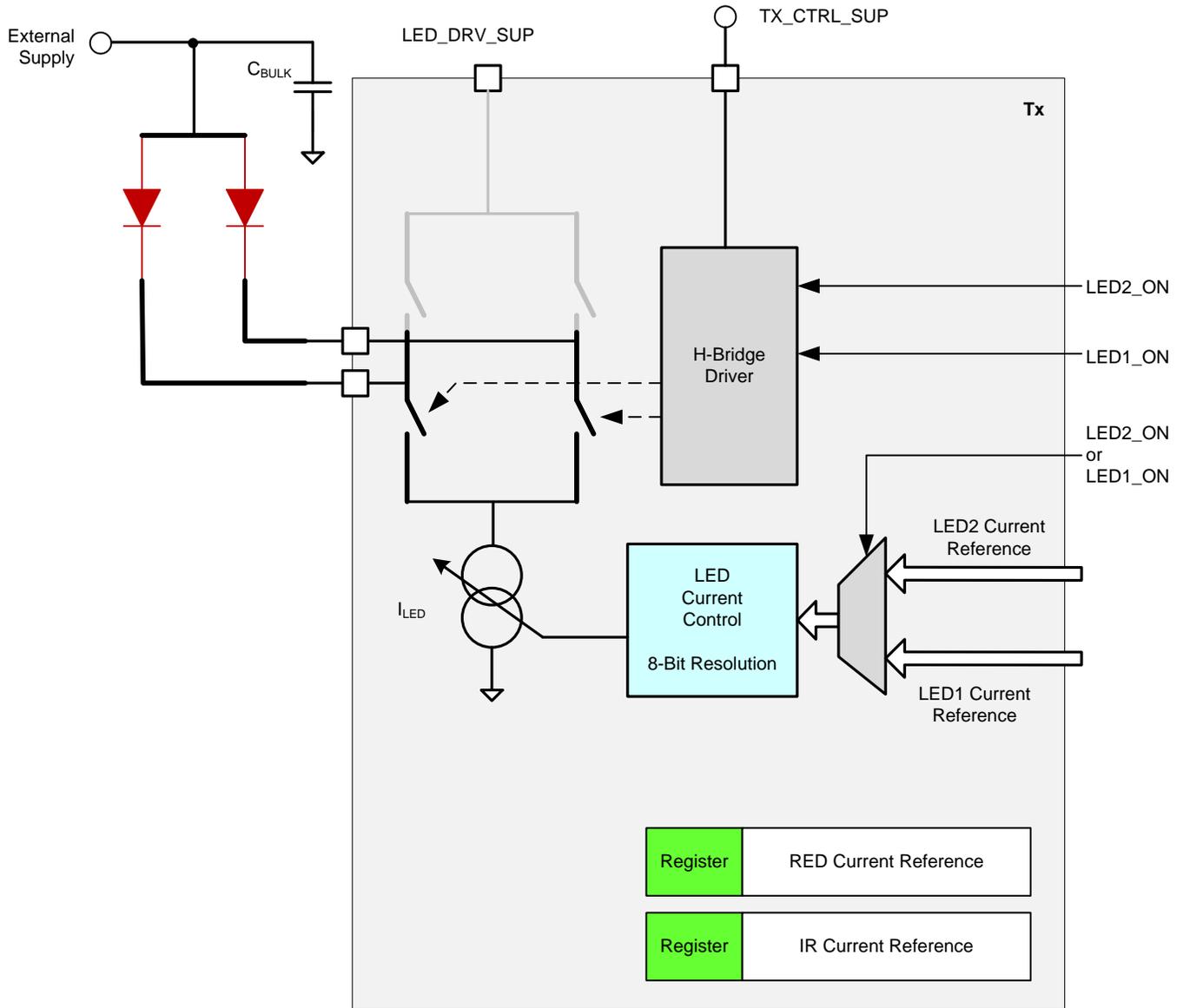


Figure 46. Transmit: Push-Pull LED Drive for Common Anode LED Configuration

### 8.3.5.1 Transmitter Power Path

The block diagram in Figure 47 shows the AFE4400 Tx subsystem power routing.

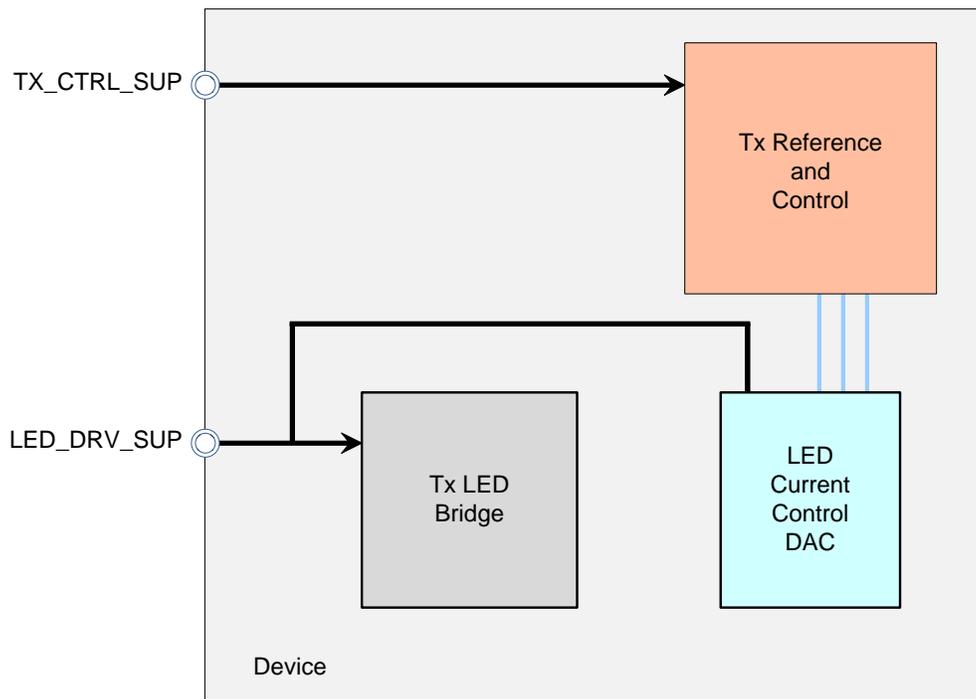


Figure 47. Transmit Subsystem Power Routing

### 8.3.5.2 LED Power Reduction During Periods of Inactivity

The diagram in Figure 48 shows how LED bias current passes 50  $\mu\text{A}$  whenever LED\_ON occurs. In order to minimize power consumption in periods of inactivity, the LED\_ON control must be turned off. Furthermore, the TIMEREN bit in the CONTROL1 register should be disabled by setting the value to 0.

Note that depending on the LEDs used, the LED may sometimes appear dimly lit even when the LED current is set to 0 mA. This appearance is because of the switching leakage currents (as shown in Figure 48) inherent to the timer function. The dimmed appearance does not effect the ambient light level measurement because during the ambient cycle, LED\_ON is turned off for the duration of the ambient measurement.

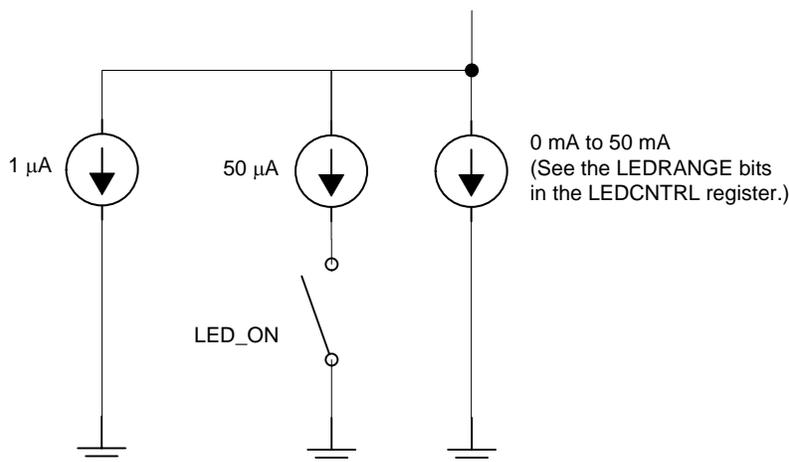


Figure 48. LED Bias Current

## 8.4 Device Functional Modes

### 8.4.1 ADC Operation and Averaging Module

After the falling edge of the ADC reset signal, the ADC conversion phase starts (refer to Figure 43). Each ADC conversion takes 50  $\mu$ s.

The ADC operates with averaging. The averaging module averages multiple ADC samples and reduces noise to improve dynamic range. Figure 49 shows a diagram of the averaging module. The ADC output format is in 22-bit twos complement, as shown in Figure 50. The two MSB bits of the 24-bit data can be ignored.

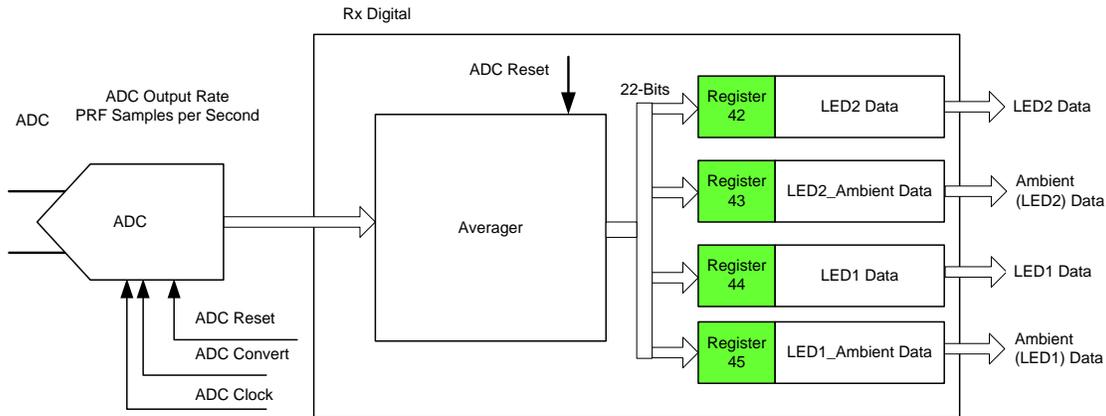


Figure 49. Averaging Module

Figure 50. 22-Bit Word

23	22	21	20	19	18	17	16	15	14	13	12
Ignore											
22-Bit ADC Code, MSB to LSB											
11	10	9	8	7	6	5	4	3	2	1	0
22-Bit ADC Code, MSB to LSB											

Table 3 shows the mapping of the input voltage to the ADC to its output code.

Table 3. ADC Input Voltage Mapping

DIFFERENTIAL INPUT VOLTAGE AT ADC INPUT	22-BIT ADC OUTPUT CODE
-1.2 V	10000000000000000000
$(-1.2 / 2^{21})$ V	11111111111111111111
0	00000000000000000000
$(1.2 / 2^{21})$ V	00000000000000000001
1.2 V	01111111111111111111

The data format is binary twos complement format, MSB-first. Because the TIA has a full-scale range of  $\pm 1$  V, TI recommends that the input to the ADC does not exceed  $\pm 1$  V, which is approximately 80% of its full-scale.

In cases where having the processor read the data as a 24-bit word instead of a 22-bit word is more convenient, the entire register can be mapped to the input level as shown in Figure 51.

Figure 51. 24-Bit Word

23	22	21	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
24-Bit ADC Code, MSB to LSB																							

Table 4 shows the mapping of the input voltage to the ADC to its output code when the entire 24-bit word is considered.

**Table 4. Input Voltage Mapping**

DIFFERENTIAL INPUT VOLTAGE AT ADC INPUT	24-BIT ADC OUTPUT CODE
-1.2 V	111000000000000000000000
$(-1.2 / 2^{21})$ V	111111111111111111111111
0	000000000000000000000000
$(1.2 / 2^{21})$ V	000000000000000000000001
1.2 V	000111111111111111111111

Now the data can be considered as a 24-bit data in binary twos complement format, MSB-first. The advantage of using the entire 24-bit word is that the ADC output is correct, even when the input is over the normal operating range.

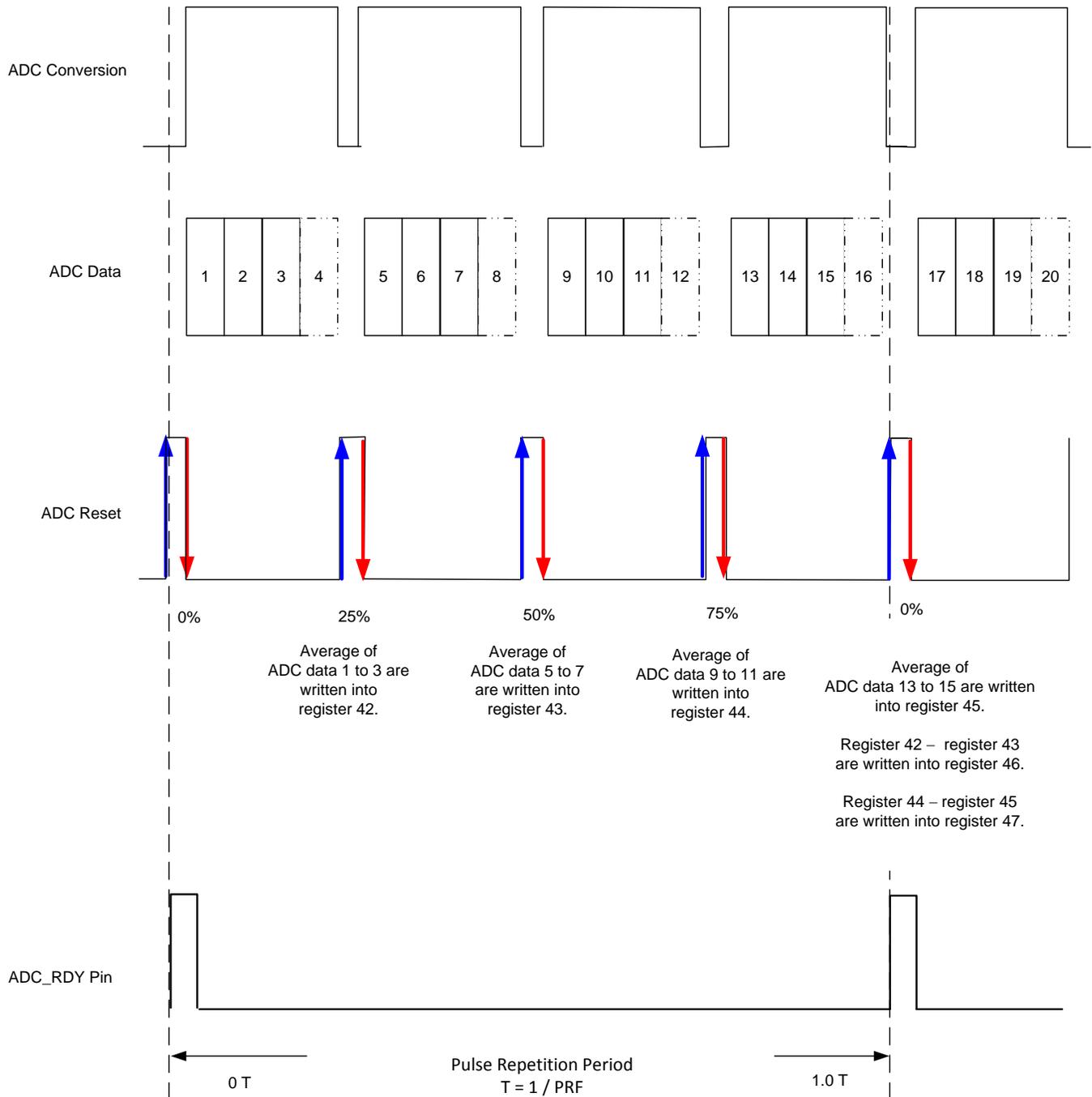
#### 8.4.1.1 Operation

The ADC digital samples are accumulated and averaged after every 50  $\mu$ s. Then, at the next rising edge of the ADC reset signal, the average value (22-bit) is written into the output registers sequentially as follows (see Figure 52):

- At the 25% reset signal, the averaged 22-bit word is written to register 2Ah.
- At the 50% reset signal, the averaged 22-bit word is written to register 2Bh.
- At the 75% reset signal, the averaged 22-bit word is written to register 2Ch.
- At the next 0% reset signal, the averaged 22-bit word is written to register 2Dh. The contents of registers 2Ah and 2Bh are written to register 2Eh and the contents of registers 2Ch and 2Dh are written to register 2Fh.

At the rising edge of the ADC\_RDY signal, the contents of all six result registers can be read out.

The number of samples to be used per conversion phase is preset to 2.



NOTE: This example shows three data averages.

**Figure 52. ADC Data with Averaging**

### 8.4.2 Diagnostics

The device includes diagnostics to detect open or short conditions of the LED and photosensor, LED current profile feedback, and cable on or off detection.

#### 8.4.2.1 Photodiode-Side Fault Detection

Figure 53 shows the diagnostic for the photodiode-side fault detection.

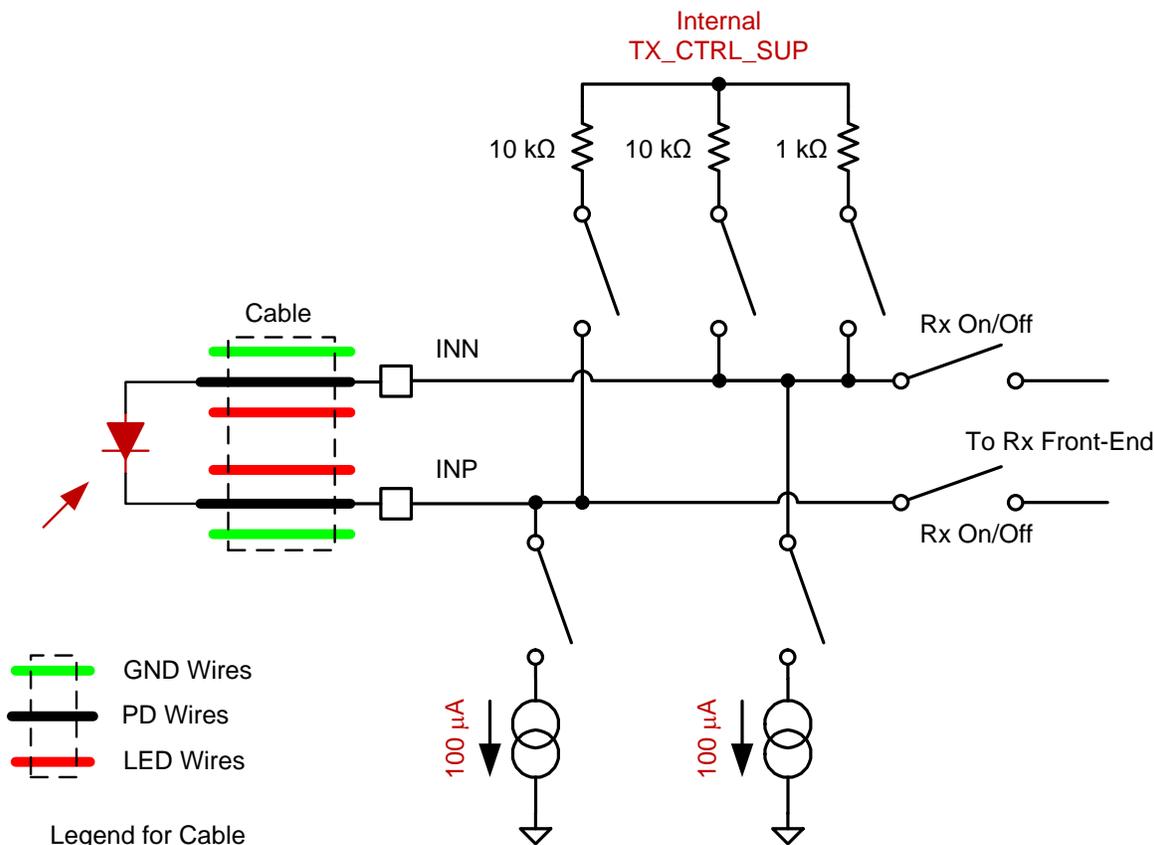


Figure 53. Photodiode Diagnostic



**8.4.2.3 Diagnostics Module**

The diagnostics module, when enabled, checks for nine types of faults sequentially. The results of all faults are latched in 11 separate flags.

At the end of the sequence, the state of the 11 flags are combined to generate two interrupt signals: PD\_ALM for photodiode-related faults and LED\_ALM for transmit-related faults.

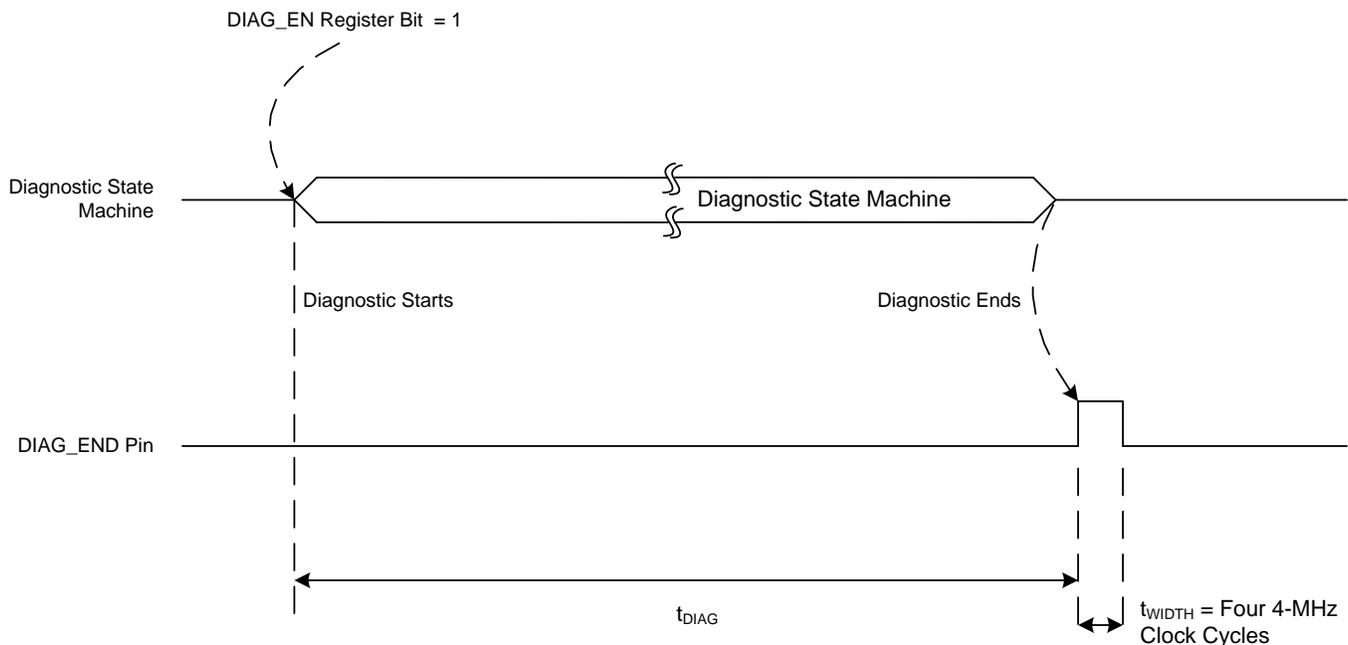
The status of all flags can also be read using the SPI interface. [Table 5](#) details each fault and flag used. Note that the diagnostics module requires all AFE blocks to be enabled in order to function reliably.

**Table 5. Fault and Flag Diagnostics<sup>(1)</sup>**

MODULE	SEQ.	FAULT	FLAG1	FLAG2	FLAG3	FLAG4	FLAG5	FLAG6	FLAG7	FLAG8	FLAG9	FLAG10	FLAG11
—	—	No fault	0	0	0	0	0	0	0	0	0	0	0
PD	1	Rx INP cable shorted to LED cable	1										
	2	Rx INN cable shorted to LED cable		1									
	3	Rx INP cable shorted to GND cable			1								
	4	Rx INN cable shorted to GND cable				1							
	5	PD open or shorted					1	1					
LED	6	Tx OUTM line shorted to GND cable							1				
	7	Tx OUTP line shorted to GND cable								1			
	8	LED open or shorted									1	1	
	9	LED open or shorted											1

(1) Resistances below 10 kΩ are considered to be shorted.

Figure 55 shows the timing for the diagnostic function.



**Figure 55. Diagnostic Timing Diagram**

By default, the diagnostic function takes  $t_{DIAG} = 16$  ms to complete. After the diagnostics function completes, the AFE4400 filter must be allowed time to settle. See the [Electrical Characteristics](#) for the filter settling time.

## 8.5 Programming

### 8.5.1 Serial Programming Interface

The SPI-compatible serial interface consists of four signals: SCLK (serial clock), SPISOMI (serial interface data output), SPISIMO (serial interface data input), and SPISTE (serial interface enable).

The serial clock (SCLK) is the serial peripheral interface (SPI) serial clock. SCLK shifts in commands and shifts out data from the device. SCLK features a Schmitt-triggered input and clocks data out on the SPISOMI. Data are clocked in on the SPISIMO pin. Even though the input has hysteresis, TI recommends keeping SCLK as clean as possible to prevent glitches from accidentally shifting the data. When the serial interface is idle, hold SCLK low.

The SPI serial out master in (SPISOMI) pin is used with SCLK to clock out the AFE4400 data. The SPI serial in master out (SPISIMO) pin is used with SCLK to clock in data to the AFE4400. The SPI serial interface enable (SPISTE) pin enables the serial interface to clock data on the SPISIMO pin in to the device.

### 8.5.2 Reading and Writing Data

The device has a set of internal registers that can be accessed by the serial programming interface formed by the SPISTE, SCLK, SPISIMO, and SPISOMI pins.

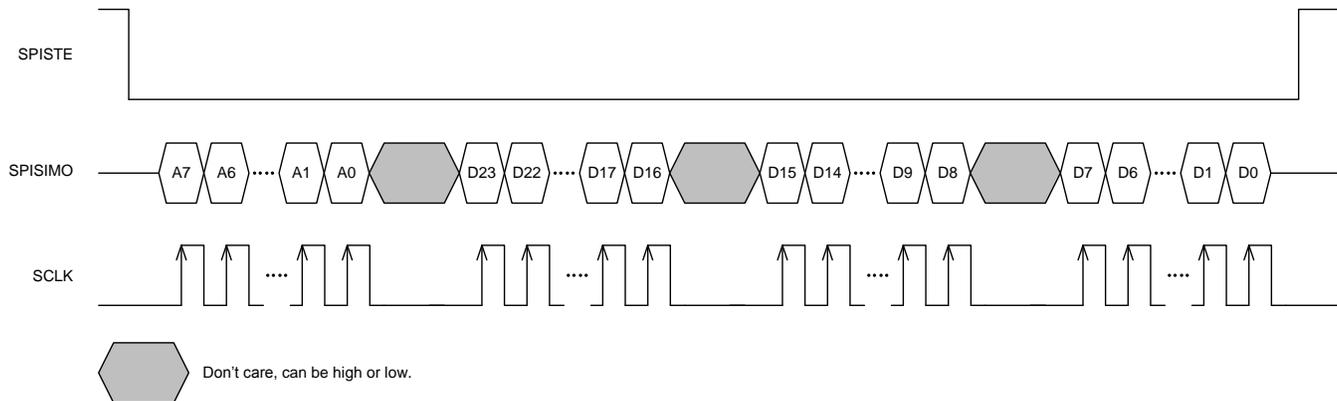
#### 8.5.2.1 Writing Data

The SPI\_READ register bit must be first set to 0 before writing to a register. When SPISTE is low:

- Serially shifting bits into the device is enabled.
- Serial data (on the SPISIMO pin) are latched at every SCLK rising edge.
- The serial data are loaded into the register at every 32nd SCLK rising edge.

**Programming (continued)**

In case the word length exceeds a multiple of 32 bits, the excess bits are ignored. Data can be loaded in multiples of 32-bit words within a single active SPISTE pulse. The first eight bits form the register address and the remaining 24 bits form the register data. [Figure 56](#) shows an SPI timing diagram for a single write operation. For multiple read and write cycles, refer to the [Multiple Data Reads and Writes](#) section.

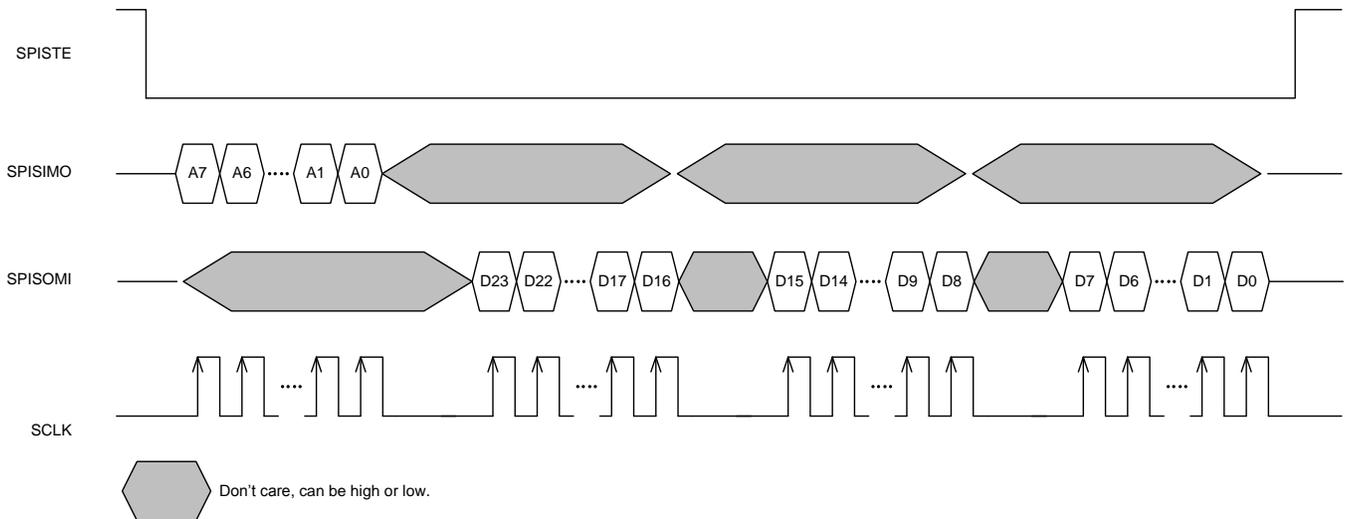


**Figure 56. AFE SPI Write Timing Diagram**

## Programming (continued)

### 8.5.2.2 Reading Data

The SPI\_READ register bit must be first set to 1 before reading from a register. The AFE4400 includes a mode where the contents of the internal registers can be read back on the SPISOMI pin. This mode may be useful as a diagnostic check to verify the serial interface communication between the external controller and the AFE. To enable this mode, first set the SPI\_READ register bit using the SPI write command, as described in the [Writing Data](#) section. In the next command, specify the SPI register address with the desired content to be read. Within the same SPI command sequence, the AFE outputs the contents of the specified register on the SPISOMI pin. [Figure 57](#) shows an SPI timing diagram for a single read operation. For multiple read and write cycles, refer to the [Multiple Data Reads and Writes](#) section.

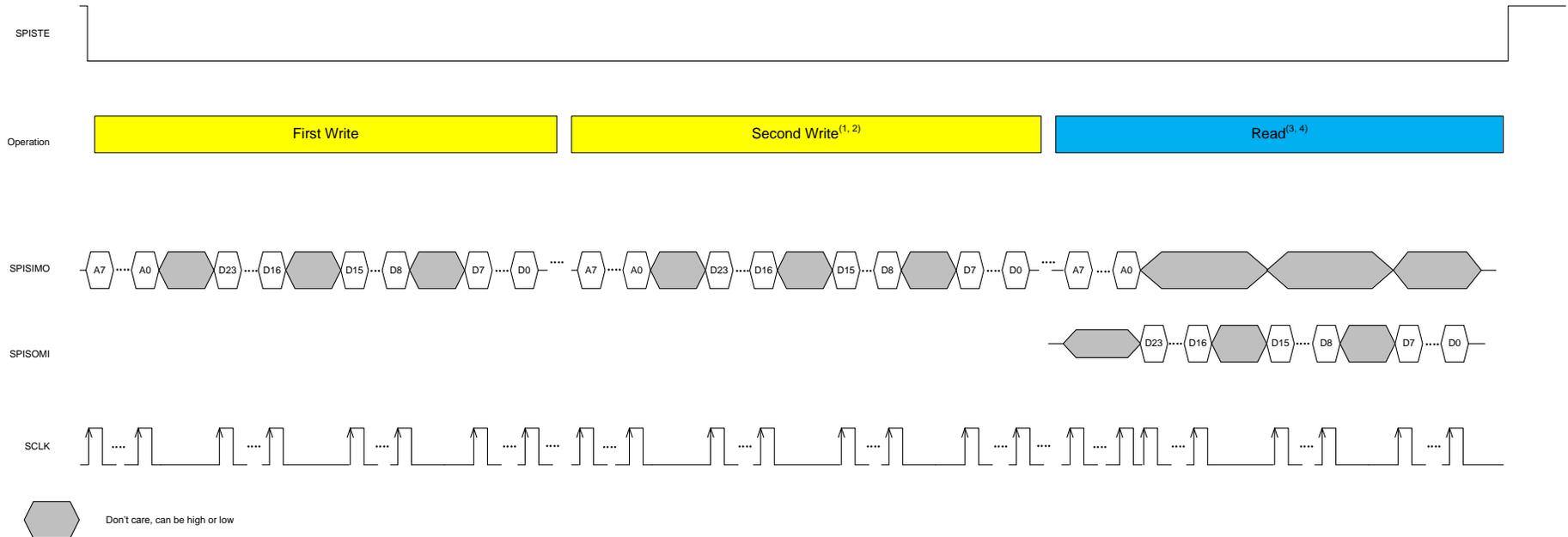


- (1) The SPI\_READ register bit must be enabled before attempting a serial readout from the AFE.
- (2) Specify the register address of the content that must be readback on bits A[7:0].
- (3) The AFE outputs the contents of the specified register on the SPISOMI pin.

**Figure 57. AFE SPI Read Timing Diagram**

### 8.5.2.3 Multiple Data Reads and Writes

The device includes functionality where multiple read and write operations can be performed during a single SPISTE event. To enable this functionality, the first eight bits determine the register address to be written and the remaining 24 bits determine the register data. Perform two writes with the SPI read bit enabled during the second write operation in order to prepare for the read operation, as described in the *Writing Data* section. In the next command, specify the SPI register address with the desired content to be read. Within the same SPI command sequence, the AFE outputs the contents of the specified register on the SPISOMI pin. This functionality is described in the *Writing Data* and *Reading Data* sections. Figure 58 shows a timing diagram for the SPI multiple read and write operations.



- (1) The SPI read register bit must be enabled before attempting a serial readout from the AFE.
- (2) The second write operation must be configured for register 0 with data 000001h.
- (3) Specify the register address whose contents must be read back on A[7:0].
- (4) The AFE outputs the contents of the specified register on the SPISOMI pin.

**Figure 58. Serial Multiple Read and Write Operations**

#### 8.5.2.4 Register Initialization

After power-up, the internal registers **must** be initialized to the default values. This initialization can be done in one of two ways:

- Through a hardware reset by applying a low-going pulse on the  $\overline{\text{RESET}}$  pin, or
- By applying a software reset. Using the serial interface, set SW\_RESET (bit D3 in register 00h) high. This setting initializes the internal registers to the default values and then self-resets to 0. In this case, the  $\overline{\text{RESET}}$  pin is kept high (inactive).

#### 8.5.2.5 AFE SPI Interface Design Considerations

Note that when the AFE4400 is deselected, the SPISOMI, CLKOUT, ADC\_RDY, PD\_ALM, LED\_ALM, and DIAG\_END digital output pins do not enter a 3-state mode. This condition, therefore, must be taken into account when connecting multiple devices to the SPI port and for power-management considerations. In order to avoid loading the SPI bus when multiple devices are connected, the DIGOUT\_TRISTATE register bit must be to 1 whenever the AFE SPI is inactive.



**Table 6. AFE Register Map (continued)**

NAME	REGISTER CONTROL <sup>(1)</sup>	ADDRESS		REGISTER DATA																								
		Hex	Dec	23	22	21	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	
ADCRSTSTCT3	R/W	1B	27	0	0	0	0	0	0	0	0	ADCRSTCT3[15:0]																
ADCRSTENDCT3	R/W	1C	28	0	0	0	0	0	0	0	0	ADCRENDCT3[15:0]																
PRPCOUNT	R/W	1D	29	0	0	0	0	0	0	0	0	PRPCT[15:0]																
CONTROL1	R/W	1E	30	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	1	0
SPARE1	N/A	1F	31	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
TIAGAIN	R/W	20	32	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
TIA_AMP_GAIN	R/W	21	33	0	0	0	0	AMBDAC[3:0]			0	STAGE2EN	0	0	0	STG2GAIN[2:0]			CF_LED[4:0]				RF_LED[2:0]					
LEDCNTRL	R/W	22	34	0	0	0	0	0	0	LEDCUROFF	1	LED1[7:0]							LED2[7:0]									
CONTROL2	R/W	23	35	0	0	0	0	0	0	0	1	0	0	0	0	0	TXBRGMOD	DIGOUT_TRISTATE	XTALDIS	1	0	0	0	0	0	0	0	0
SPARE2	N/A	24	36	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
SPARE3	N/A	25	37	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
SPARE4	N/A	26	38	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
RESERVED1	N/A	27	39	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
RESERVED2	N/A	28	40	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
ALARM	R/W	29	41	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	ALMPINCKEN	0	0	0	0	0	0	0
LED2VAL	R	2A	42	LED2VAL[23:0]																								
ALED2VAL	R	2B	43	ALED2VAL[23:0]																								
LED1VAL	R	2C	44	LED1VAL[23:0]																								
ALED1VAL	R	2D	45	ALED1VAL[23:0]																								
LED2-ALED2VAL	R	2E	46	LED2-ALED2VAL[23:0]																								
LED1-ALED1VAL	R	2F	47	LED1-ALED1VAL[23:0]																								



## 8.6.2 AFE Register Description

**Figure 59. CONTROL0: Control Register 0 (Address = 00h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	0	0	0	0
11	10	9	8	7	6	5	4	3	2	1	0
0	0	0	0	0	0	0	0	SW_RST	DIAG_EN	TIM_COUNT_RST	SPI_READ

This register is write-only. CONTROL0 is used for AFE software and count timer reset, diagnostics enable, and SPI read functions.

**Bits 23:4**      **Must be 0**

**Bit 3**            **SW\_RST: Software reset**

0 = No action (default after reset)

1 = Software reset applied; resets all internal registers to the default values and self-clears to 0

**Bit 2**            **DIAG\_EN: Diagnostic enable**

0 = No action (default after reset)

1 = Diagnostic mode is enabled and the diagnostics sequence starts when this bit is set.

At the end of the sequence, all fault status are stored in the [DIAG: Diagnostics Flag Register](#). Afterwards, the DIAG\_EN register bit self-clears to 0.

Note that the diagnostics enable bit is automatically reset after the diagnostics completes (16 ms). During the diagnostics mode, ADC data are invalid because of the toggling diagnostics switches.

**Bit 1**            **TIM\_CNT\_RST: Timer counter reset**

0 = Disables timer counter reset, required for normal timer operation (default after reset)

1 = Timer counters are in reset state

**Bit 0**            **SPI\_READ: SPI read**

0 = SPI read is disabled (default after reset)

1 = SPI read is enabled

**Figure 60. LED2STC: Sample LED2 Start Count Register (Address = 01h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	LED2STC[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
LED2STC[15:0]											

This register sets the start timing value for the LED2 signal sample.

**Bits 23:16**      **Must be 0**

**Bits 15:0**        **LED2STC[15:0]: Sample LED2 start count**

The contents of this register can be used to position the start of the sample LED2 signal with respect to the pulse repetition period (PRP), as specified in the PRPCOUNT register. The count is specified as the number of

4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 61. LED2ENDC: Sample LED2 End Count Register (Address = 02h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	LED2ENDC[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
LED2ENDC[15:0]											

This register sets the end timing value for the LED2 signal sample.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **LED2ENDC[15:0]: Sample LED2 end count**

The contents of this register can be used to position the end of the sample LED2 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 62. LED2LEDSTC: LED2 LED Start Count Register (Address = 03h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	LED2LEDSTC[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
LED2LEDSTC[15:0]											

This register sets the start timing value for when the LED2 signal turns on.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **LED2LEDSTC[15:0]: LED2 start count**

The contents of this register can be used to position the start of the LED2 with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 63. LED2LEDENDC: LED2 LED End Count Register (Address = 04h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	LED2LEDENDC[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
LED2LEDENDC[15:0]											

This register sets the end timing value for when the LED2 signal turns off.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **LED2LEDENDC[15:0]: LED2 end count**

The contents of this register can be used to position the end of the LED2 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 64. ALED2STC: Sample Ambient LED2 Start Count Register (Address = 05h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	ALED2STC[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
ALED2STC[15:0]											

This register sets the start timing value for the ambient LED2 signal sample.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **ALED2STC[15:0]: Sample ambient LED2 start count**

The contents of this register can be used to position the start of the sample ambient LED2 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 65. ALED2ENDC: Sample Ambient LED2 End Count Register (Address = 06h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	ALED2ENDC[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
ALED2ENDC[15:0]											

This register sets the end timing value for the ambient LED2 signal sample.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **ALED2ENDC[15:0]: Sample ambient LED2 end count**

The contents of this register can be used to position the end of the sample ambient LED2 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 66. LED1STC: Sample LED1 Start Count Register (Address = 07h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	LED1STC[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
LED1STC[15:0]											

This register sets the start timing value for the LED1 signal sample.

**Bits 23:17**      **Must be 0**

**Bits 16:0**      **LED1STC[15:0]: Sample LED1 start count**

The contents of this register can be used to position the start of the sample LED1 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 67. LED1ENDC: Sample LED1 End Count (Address = 08h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	LED1ENDC[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
LED1ENDC[15:0]											

This register sets the end timing value for the LED1 signal sample.

**Bits 23:17**      **Must be 0**

**Bits 16:0**      **LED1ENDC[15:0]: Sample LED1 end count**

The contents of this register can be used to position the end of the sample LED1 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 68. LED1LEDSTC: LED1 LED Start Count Register (Address = 09h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	LED1LEDSTC[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
LED1LEDSTC[15:0]											

This register sets the start timing value for when the LED1 signal turns on.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **LED1LEDSTC[15:0]: LED1 start count**

The contents of this register can be used to position the start of the LED1 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 69. LED1LEDENDC: LED1 LED End Count Register (Address = 0Ah, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	LED1LEDENDC[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
LED1LEDENDC[15:0]											

This register sets the end timing value for when the LED1 signal turns off.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **LED1LEDENDC[15:0]: LED1 end count**

The contents of this register can be used to position the end of the LED1 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 70. ALED1STC: Sample Ambient LED1 Start Count Register (Address = 0Bh, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	ALED1STC[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
ALED1STC[15:0]											

This register sets the start timing value for the ambient LED1 signal sample.

**Bits 23:16**
**Must be 0**
**Bits 15:0**
**ALED1STC[15:0]: Sample ambient LED1 start count**

The contents of this register can be used to position the start of the sample ambient LED1 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 71. ALED1ENDC: Sample Ambient LED1 End Count Register (Address = 0Ch, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	ALED1ENDC[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
ALED1ENDC[15:0]											

This register sets the end timing value for the ambient LED1 signal sample.

**Bits 23:16**
**Must be 0**
**Bits 15:0**
**ALED1ENDC[15:0]: Sample ambient LED1 end count**

The contents of this register can be used to position the end of the sample ambient LED1 signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 72. LED2CONVST: LED2 Convert Start Count Register (Address = 0Dh, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	LED2CONVST[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
LED2CONVST[15:0]											

This register sets the start timing value for the LED2 conversion.

**Bits 23:16**
**Must be 0**
**Bits 15:0**
**LED2CONVST[15:0]: LED2 convert start count**

The contents of this register can be used to position the start of the LED2 conversion signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 73. LED2CONVEND: LED2 Convert End Count Register (Address = 0Eh, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	LED2CONVEND[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
LED2CONVEND[15:0]											

This register sets the end timing value for the LED2 conversion.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **LED2CONVEND[15:0]: LED2 convert end count**

The contents of this register can be used to position the end of the LED2 conversion signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 74. ALED2CONVST: LED2 Ambient Convert Start Count Register (Address = 0Fh, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	ALED2CONVST[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
ALED2CONVST[15:0]											

This register sets the start timing value for the ambient LED2 conversion.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **ALED2CONVST[15:0]: LED2 ambient convert start count**

The contents of this register can be used to position the start of the LED2 ambient conversion signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 75. ALED2CONVEND: LED2 Ambient Convert End Count Register (Address = 10h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	ALED2CONVEND[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
ALED2CONVEND[15:0]											

This register sets the end timing value for the ambient LED2 conversion.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **ALED2CONVEND[15:0]: LED2 ambient convert end count**

The contents of this register can be used to position the end of the LED2 ambient conversion signal with respect to the PRP. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 76. LED1CONVST: LED1 Convert Start Count Register (Address = 11h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	LED1CONVST[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
LED1CONVST[15:0]											

This register sets the start timing value for the LED1 conversion.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **LED1CONVST[15:0]: LED1 convert start count**

The contents of this register can be used to position the start of the LED1 conversion signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 77. LED1CONVEND: LED1 Convert End Count Register (Address = 12h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	LED1CONVEND[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
LED1CONVEND[15:0]											

This register sets the end timing value for the LED1 conversion.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **LED1CONVEND[15:0]: LED1 convert end count**

The contents of this register can be used to position the end of the LED1 conversion signal with respect to the PRP. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 78. ALED1CONVST: LED1 Ambient Convert Start Count Register (Address = 13h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	ALED1CONVST[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
ALED1CONVST[15:0]											

This register sets the start timing value for the ambient LED1 conversion.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **ALED1CONVST[15:0]: LED1 ambient convert start count**

The contents of this register can be used to position the start of the LED1 ambient conversion signal with respect to the PRP, as specified in the PRPCOUNT register. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 79. ALED1CONVEND: LED1 Ambient Convert End Count Register (Address = 14h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	ALED1CONVEND[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
ALED1CONVEND[15:0]											

This register sets the end timing value for the ambient LED1 conversion.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **ALED1CONVEND[15:0]: LED1 ambient convert end count**

The contents of this register can be used to position the end of the LED1 ambient conversion signal with respect to the PRP. The count is specified as the number of 4-MHz clock cycles. Refer to the [Using the Timer Module](#) section for details.

**Figure 80. ADCRSTSTCT0: ADC Reset 0 Start Count Register (Address = 15h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	ADCRSTSTCT0[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
ADCRSTSTCT0[15:0]											

This register sets the start position of the ADC0 reset conversion signal.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **ADCRSTSTCT0[15:0]: ADC RESET 0 start count**

The contents of this register can be used to position the start of the ADC reset conversion signal (default value after reset is 0000h). Refer to the [Using the Timer Module](#) section for details.

**Figure 81. ADCRSTENDCT0: ADC Reset 0 End Count Register (Address = 16h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	ADCRSTENDCT0[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
ADCRSTENDCT0[15:0]											

This register sets the end position of the ADC0 reset conversion signal.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **ADCRSTENDCT0[15:0]: ADC RESET 0 end count**

The contents of this register can be used to position the end of the ADC reset conversion signal (default value after reset is 0000h). Refer to the [Using the Timer Module](#) section for details.

**Figure 82. ADCRSTSTCT1: ADC Reset 1 Start Count Register (Address = 17h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	ADCRSTSTCT1[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
ADCRSTSTCT1[15:0]											

This register sets the start position of the ADC1 reset conversion signal.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **ADCRSTSTCT1[15:0]: ADC RESET 1 start count**

The contents of this register can be used to position the start of the ADC reset conversion. Refer to the [Using the Timer Module](#) section for details.

**Figure 83. ADCRSTENDCT1: ADC Reset 1 End Count Register (Address = 18h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	ADCRSTENDCT1[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
ADCRSTENDCT1[15:0]											

This register sets the end position of the ADC1 reset conversion signal.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **ADCRSTENDCT1[15:0]: ADC RESET 1 end count**

The contents of this register can be used to position the end of the ADC reset conversion. Refer to the [Using the Timer Module](#) section for details.

**Figure 84. ADCRSTSTCT2: ADC Reset 2 Start Count Register (Address = 19h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	ADCRSTSTCT2[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
ADCRSTSTCT2[15:0]											

This register sets the start position of the ADC2 reset conversion signal.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **ADCRSTSTCT2[15:0]: ADC RESET 2 start count**

The contents of this register can be used to position the start of the ADC reset conversion. Refer to the [Using the Timer Module](#) section for details.

**Figure 85. ADCRSTENDCT2: ADC Reset 2 End Count Register (Address = 1Ah, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	ADCRSTENDCT2[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
ADCRSTENDCT2[15:0]											

This register sets the end position of the ADC2 reset conversion signal.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **ADCRSTENDCT2[15:0]: ADC RESET 2 end count**

The contents of this register can be used to position the end of the ADC reset conversion. Refer to the [Using the Timer Module](#) section for details.

**Figure 86. ADCRSTSTCT3: ADC Reset 3 Start Count Register (Address = 1Bh, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	ADCRSTSTCT3[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
ADCRSTSTCT3[15:0]											

This register sets the start position of the ADC3 reset conversion signal.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **ADCRSTSTCT3[15:0]: ADC RESET 3 start count**

The contents of this register can be used to position the start of the ADC reset conversion. Refer to the [Using the Timer Module](#) section for details.

**Figure 87. ADCRSTENDCT3: ADC Reset 3 End Count Register (Address = 1Ch, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	ADCRSTENDCT3[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
ADCRSTENDCT3[15:0]											

This register sets the end position of the ADC3 reset conversion signal.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **ADCRSTENDCT3[15:0]: ADC RESET 3 end count**

The contents of this register can be used to position the end of the ADC reset conversion signal (default value after reset is 0000h). Refer to the [Using the Timer Module](#) section for details.

**Figure 88. PRPCOUNT: Pulse Repetition Period Count Register (Address = 1Dh, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	PRPCOUNT[15:0]			
11	10	9	8	7	6	5	4	3	2	1	0
PRPCOUNT[15:0]											

This register sets the device pulse repetition period count.

**Bits 23:16**      **Must be 0**

**Bits 15:0**      **PRPCOUNT[15:0]: Pulse repetition period count**

The contents of this register can be used to set the pulse repetition period (in number of clock cycles of the 4-MHz clock). The PRPCOUNT value must be set in the range of 800 to 64000. Values below 800 do not allow sufficient sample time for the four samples; see the [Electrical Characteristics](#) table.

**Figure 89. CONTROL1: Control Register 1 (Address = 1Eh, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	0	0	0	0
11	10	9	8	7	6	5	4	3	2	1	0
CLKALMPIN[2:0]			TIMEREN	0	0	0	0	0	0	1	0

This register configures the clock alarm pin and timer.

**Bits 23:12**      **Must be 0**

**Bits 11:9**      **CLKALMPIN[2:0]: Clocks on ALM pins**

Internal clocks can be brought to the PD\_ALM and LED\_ALM pins for monitoring. Note that the ALMPINCLKEN register bit must be set before using this register bit. [Table 7](#) defines the settings for the two alarm pins.

**Bit 8**      **TIMEREN: Timer enable**

0 = Timer module is disabled and all internal clocks are off (default after reset)  
1 = Timer module is enabled

**Bits 7:2**      **Must be 0**

**Bit 1**      **Must be 1**

**Bit 0**      **Must be 0**

**Table 7. PD\_ALM and LED\_ALM Pin Settings**

CLKALMPIN[2:0]	PD_ALM PIN SIGNAL	LED_ALM PIN SIGNAL
000	Sample LED2 pulse	Sample LED1 pulse
001	LED2 LED pulse	LED1 LED pulse
010	Sample LED2 ambient pulse	Sample LED1 ambient pulse
011	LED2 convert	LED1 convert
100	LED2 ambient convert	LED1 ambient convert
101	No output	No output
110	No output	No output
111	No output	No output

**Figure 90. SPARE1: SPARE1 Register For Future Use (Address = 1Fh, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	0	0	0	0
11	10	9	8	7	6	5	4	3	2	1	0
0	0	0	0	0	0	0	0	0	0	0	0

This register is a spare register and is reserved for future use.

**Bits 23:0            Must be 0**

**Figure 91. TIAGAIN: Transimpedance Amplifier Gain Setting Register  
(Address = 20h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	0	0	0	0
11	10	9	8	7	6	5	4	3	2	1	0
0	0	0	0	0	0	0	0	0	0	0	0

This register is reserved for factory use.

**Bits 23:0            Must be 0**

**Figure 92. TIA\_AMB\_GAIN: Transimpedance Amplifier and Ambient Cancellation Stage Gain Register  
(Address = 21h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	AMBDAC[3:0]				0	STAGE2 EN	0	0
11	10	9	8	7	6	5	4	3	2	1	0
0	STG2GAIN[2:0]			CF_LED2[4:0]				RF_LED2[2:0]			

This register configures the ambient light cancellation amplifier gain, cancellation current, and filter corner frequency.

**Bits 23:20**      **Must be 0**

**Bits 19:16**      **AMBDAC[3:0]: Ambient DAC value**

These bits set the value of the cancellation current.

0000 = 0 $\mu$ A (default after reset)	1000 = 8 $\mu$ A
0001 = 1 $\mu$ A	1001 = 9 $\mu$ A
0010 = 2 $\mu$ A	1010 = 10 $\mu$ A
0011 = 3 $\mu$ A	1011 = Do not use
0100 = 4 $\mu$ A	1100 = Do not use
0101 = 5 $\mu$ A	1101 = Do not use
0110 = 6 $\mu$ A	1110 = Do not use
0111 = 7 $\mu$ A	1111 = Do not use

**Bit 15**      **Must be 0**

**Bit 14**      **STAGE2EN: Stage 2 enable for LED 2**

0 = Stage 2 is bypassed (default after reset)

1 = Stage 2 is enabled with the gain value specified by the STG2GAIN[2:0] bits

**Bits 13:11**      **Must be 0**

**Bits 10:8**      **STG2GAIN[2:0]: Stage 2 gain setting**

000 = 0 dB, or linear gain of 1 (default after reset)
001 = 3.5 dB, or linear gain of 1.5
010 = 6 dB, or linear gain of 2
011 = 9.5 dB, or linear gain of 3
100 = 12 dB, or linear gain of 4
101 = Do not use
110 = Do not use
111 = Do not use

**Bits 7:3**      **CF\_LED[4:0]: Program  $C_F$  for LEDs**

00000 = 5 pF (default after reset)	00100 = 25 pF + 5 pF
00001 = 5 pF + 5 pF	01000 = 50 pF + 5 pF
00010 = 15 pF + 5 pF	10000 = 150 pF + 5 pF

Note that any combination of these  $C_F$  settings is also supported by setting multiple bits to 1. For example, to obtain  $C_F = 100$  pF, set D[7:3] = 01111.

**Bits 2:0**      **RF\_LED[2:0]: Program  $R_F$  for LEDs**

000 = 500 k $\Omega$	100 = 25 k $\Omega$
001 = 250 k $\Omega$	101 = 10 k $\Omega$
010 = 100 k $\Omega$	110 = 1 M $\Omega$
011 = 50 k $\Omega$	111 = None

**Figure 93. LEDCNTRL: LED Control Register (Address = 22h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	LEDCUR OFF	1	LED1[7:0]			
11	10	9	8	7	6	5	4	3	2	1	0
LED1[7:0]						LED2[7:0]					

This register sets the LED current range and the LED1 and LED2 drive current.

**Bits 23:18**            **Must be 0**

**Bit 17**                **LEDCUROFF: Turns the LED current source on or off**

0 = On (50 mA)

1 = Off

**Bit 16**                **Must be 1**

**Bits 15:8**            **LED1[7:0]: Program LED current for LED1 signal**

Use these register bits to specify the LED current setting for LED1 (default after reset is 00h).

The nominal value of the LED current is given by [Equation 3](#), where the full-scale LED current is 50 mA.

**Bits 7:0**             **LED2[7:0]: Program LED current for LED2 signal**

Use these register bits to specify the LED current setting for LED2 (default after reset is 00h).

The nominal value of LED current is given by [Equation 4](#), where the full-scale LED current is 50 mA.

$$\frac{\text{LED1[7:0]}}{256} \times \text{Full-Scale Current} \tag{3}$$

$$\frac{\text{LED2[7:0]}}{256} \times \text{Full-Scale Current} \tag{4}$$

**Figure 94. CONTROL2: Control Register 2 (Address = 23h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	1	0	0	0	0	0
11	10	9	8	7	6	5	4	3	2	1	0
TXBRG MOD	DIGOUT_ TRI STATE	XTAL DIS	1	0	0	0	0	0	PDNTX	PDNRX	PDNAFE

This register controls the LED transmitter, crystal, and the AFE, transmitter, and receiver power modes.

**Bits 23:18**            **Must be 0**

**Bit 17**                **Must be 1**

**Bits 16:12**           **Must be 0**

**Bit 11**                **TXBRGMOD: Tx bridge mode**

0 = LED driver is configured as an H-bridge (default after reset)  
1 = LED driver is configured as a push-pull

**Bit 10**                **DIGOUT\_TRISTATE: Digital output 3-state mode**

This bit determines the state of the device digital output pins, including the clock output pin and SPI output pins. In order to avoid loading the SPI bus when multiple devices are connected, this bit must be set to 1 (3-state mode) whenever the device SPI is inactive.

0 = Normal operation (default)  
1 = 3-state mode

**Bit 9**                 **XTALDIS: Crystal disable mode**

0 = The crystal module is enabled; the 8-MHz crystal must be connected to the XIN and XOUT pins  
1 = The crystal module is disabled; an external 8-MHz clock must be applied to the XIN pin

**Bit 8**                 **Must be 1**

**Bits 7:3**              **Must be 0**

**Bit 2**                 **PDN\_TX: Tx power-down**

0 = The Tx is powered up (default after reset)  
1 = Only the Tx module is powered down

**Bit 1**                 **PDN\_RX: Rx power-down**

0 = The Rx is powered up (default after reset)  
1 = Only the Rx module is powered down

**Bit 0**                 **PDN\_AFE: AFE power-down**

0 = The AFE is powered up (default after reset)  
1 = The entire AFE is powered down (including the Tx, Rx, and diagnostics blocks)

**Figure 95. SPARE2: SPARE2 Register For Future Use (Address = 24h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	0	0	0	0
11	10	9	8	7	6	5	4	3	2	1	0
0	0	0	0	0	0	0	0	0	0	0	0

This register is a spare register and is reserved for future use.

**Bits 23:0**            **Must be 0**

**Figure 96. SPARE3: SPARE3 Register For Future Use (Address = 25h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	0	0	0	0
11	10	9	8	7	6	5	4	3	2	1	0
0	0	0	0	0	0	0	0	0	0	0	0

This register is a spare register and is reserved for future use.

**Bits 23:0**            **Must be 0**

**Figure 97. SPARE4: SPARE4 Register For Future Use (Address = 26h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	0	0	0	0
11	10	9	8	7	6	5	4	3	2	1	0
0	0	0	0	0	0	0	0	0	0	0	0

This register is a spare register and is reserved for future use.

**Bits 23:0**            **Must be 0**

**Figure 98. RESERVED1: RESERVED1 Register For Factory Use Only  
(Address = 27h, Reset Value = XXXXh)**

23	22	21	20	19	18	17	16	15	14	13	12
X <sup>(1)</sup>	X	X	X	X	X	X	X	X	X	X	X
11	10	9	8	7	6	5	4	3	2	1	0
X	X	X	X	X	X	X	X	X	X	X	X

(1) X = don't care.

This register is reserved for factory use. Readback values vary between devices.

**Figure 99. RESERVED2: RESERVED2 Register For Factory Use Only  
(Address = 28h, Reset Value = XXXXh)**

23	22	21	20	19	18	17	16	15	14	13	12
X <sup>(1)</sup>	X	X	X	X	X	X	X	X	X	X	X
11	10	9	8	7	6	5	4	3	2	1	0
X	X	X	X	X	X	X	X	X	X	X	X

(1) X = don't care.

This register is reserved for factory use. Readback values vary between devices.

**Figure 100. ALARM: Alarm Register (Address = 29h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	0	0	0	0
11	10	9	8	7	6	5	4	3	2	1	0
0	0	0	0	ALMPIN CLKEN	0	0	0	0	0	0	0

This register controls the alarm pin functionality.

**Bits 23:8**            **Must be 0**

**Bit 7**                **ALMPINCLKEN: Alarm pin clock enable**

0 = Disables the monitoring of internal clocks; the PD\_ALM and LED\_ALM pins function as diagnostic fault alarm output pins (default after reset)

1 = Enables the monitoring of internal clocks; these clocks can be brought out on PD\_ALM and LED\_ALM selectively (depending on the value of the CLKALMPIN[2:0] register bits).

**Bits 6:0**            **Must be 0**

**Figure 101. LED2VAL: LED2 Digital Sample Value Register (Address = 2Ah, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
LED2VAL[23:0]											
11	10	9	8	7	6	5	4	3	2	1	0
LED2VAL[23:0]											

**Bits 23:0**            **LED2VAL[23:0]: LED2 digital value**

This register contains the digital value of the latest LED2 sample converted by the ADC. The ADC\_RDY signal goes high each time that the contents of this register are updated. The host processor must readout this register before the next sample is converted by the AFE.

**Figure 102. ALED2VAL: Ambient LED2 Digital Sample Value Register (Address = 2Bh, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
ALED2VAL[23:0]											
11	10	9	8	7	6	5	4	3	2	1	0
ALED2VAL[23:0]											

**Bits 23:0**            **ALED2VAL[23:0]: LED2 ambient digital value**

This register contains the digital value of the latest LED2 ambient sample converted by the ADC. The ADC\_RDY signal goes high each time that the contents of this register are updated. The host processor must readout this register before the next sample is converted by the AFE.

**Figure 103. LED1VAL: LED1 Digital Sample Value Register (Address = 2Ch, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
LED1VAL[23:0]											
11	10	9	8	7	6	5	4	3	2	1	0
LED1VAL[23:0]											

**Bits 23:0 LED1VAL[23:0]: LED1 digital value**

This register contains the digital value of the latest LED1 sample converted by the ADC. The ADC\_RDY signal goes high each time that the contents of this register are updated. The host processor must readout this register before the next sample is converted by the AFE.

**Figure 104. ALED1VAL: Ambient LED1 Digital Sample Value Register (Address = 2Dh, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
ALED1VAL[23:0]											
11	10	9	8	7	6	5	4	3	2	1	0
ALED1VAL[23:0]											

**Bits 23:0 ALED1VAL[23:0]: LED1 ambient digital value**

This register contains the digital value of the latest LED1 ambient sample converted by the ADC. The ADC\_RDY signal goes high each time that the contents of this register are updated. The host processor must readout this register before the next sample is converted by the AFE.

**Figure 105. LED2-ALED2VAL: LED2-Ambient LED2 Digital Sample Value Register (Address = 2Eh, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
LED2-ALED2VAL[23:0]											
11	10	9	8	7	6	5	4	3	2	1	0
LED2-ALED2VAL[23:0]											

**Bits 23:0 LED2-ALED2VAL[23:0]: (LED2 – LED2 ambient) digital value**

This register contains the digital value of the LED2 sample after the LED2 ambient is subtracted. The host processor must readout this register before the next sample is converted by the AFE.

Note that this value is inverted when compared to waveforms shown in many publications.

**Figure 106. LED1-ALED1VAL: LED1-Ambient LED1 Digital Sample Value Register (Address = 2Fh, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
LED1-ALED1VAL[23:0]											
11	10	9	8	7	6	5	4	3	2	1	0
LED1-ALED1VAL[23:0]											

**Bits 23:0 LED1-ALED1VAL[23:0]: (LED1 – LED1 ambient) digital value**

This register contains the digital value of the LED1 sample after the LED1 ambient is subtracted from it. The host processor must readout this register before the next sample is converted by the AFE.

Note that this value is inverted when compared to waveforms shown in many publications.

**Figure 107. DIAG: Diagnostics Flag Register (Address = 30h, Reset Value = 0000h)**

23	22	21	20	19	18	17	16	15	14	13	12
0	0	0	0	0	0	0	0	0	0	0	PD_ALM
11	10	9	8	7	6	5	4	3	2	1	0
LED_ALM	LED1_OPEN	LED2_OPEN	LEDSC	OUTPSH_GND	OUTNSH_GND	PDOC	PDSC	INNSC_GND	INPSC_GND	INNSC_LED	INPSC_LED

This register is read only. This register contains the status of all diagnostic flags at the end of the diagnostics sequence. The end of the diagnostics sequence is indicated by the signal going high on DIAG\_END pin.

**Bits 23:13**
**Read only**
**Bit 12**
**PD\_ALM: Power-down alarm status diagnostic flag**

This bit indicates the status of PD\_ALM (and the PD\_ALM pin).

0 = No fault (default after reset)

1 = Fault present

**Bit 11**
**LED\_ALM: LED alarm status diagnostic flag**

This bit indicates the status of LED\_ALM (and the LED\_ALM pin).

0 = No fault (default after reset)

1 = Fault present

**Bit 10**
**LED1OPEN: LED1 open diagnostic flag**

This bit indicates that LED1 is open.

0 = No fault (default after reset)

1 = Fault present

**Bit 9**
**LED2OPEN: LED2 open diagnostic flag**

This bit indicates that LED2 is open.

0 = No fault (default after reset)

1 = Fault present

This bit indicates that LED2 is open.

0 = No fault (default after reset)

1 = Fault present

**Bit 8**
**LEDSC: LED short diagnostic flag**

This bit indicates an LED short.

0 = No fault (default after reset)

1 = Fault present

**Bit 7**
**OUTPSHGND: OUTP to GND diagnostic flag**

This bit indicates that OUTP is shorted to the GND cable.

0 = No fault (default after reset)

1 = Fault present

**Bit 6**
**OUTNSHGND: OUTN to GND diagnostic flag**

This bit indicates that OUTN is shorted to the GND cable.

0 = No fault (default after reset)

1 = Fault present

**Bit 5**
**PDOC: PD open diagnostic flag**

This bit indicates that PD is open.

0 = No fault (default after reset)

1 = Fault present

- Bit 4 PDSC: PD short diagnostic flag**  
This bit indicates a PD short.  
0 = No fault (default after reset)  
1 = Fault present
- Bit 3 INNSCGND: INN to GND diagnostic flag**  
This bit indicates a short from the INN pin to the GND cable.  
0 = No fault (default after reset)  
1 = Fault present
- Bit 2 INPSCGND: INP to GND diagnostic flag**  
This bit indicates a short from the INP pin to the GND cable.  
0 = No fault (default after reset)  
1 = Fault present
- Bit 1 INNSCLED: INN to LED diagnostic flag**  
This bit indicates a short from the INN pin to the LED cable.  
0 = No fault (default after reset)  
1 = Fault present
- Bit 0 INPSCLED: INP to LED diagnostic flag**  
This bit indicates a short from the INP pin to the LED cable.  
0 = No fault (default after reset)  
1 = Fault present

## 9 Applications and Implementation

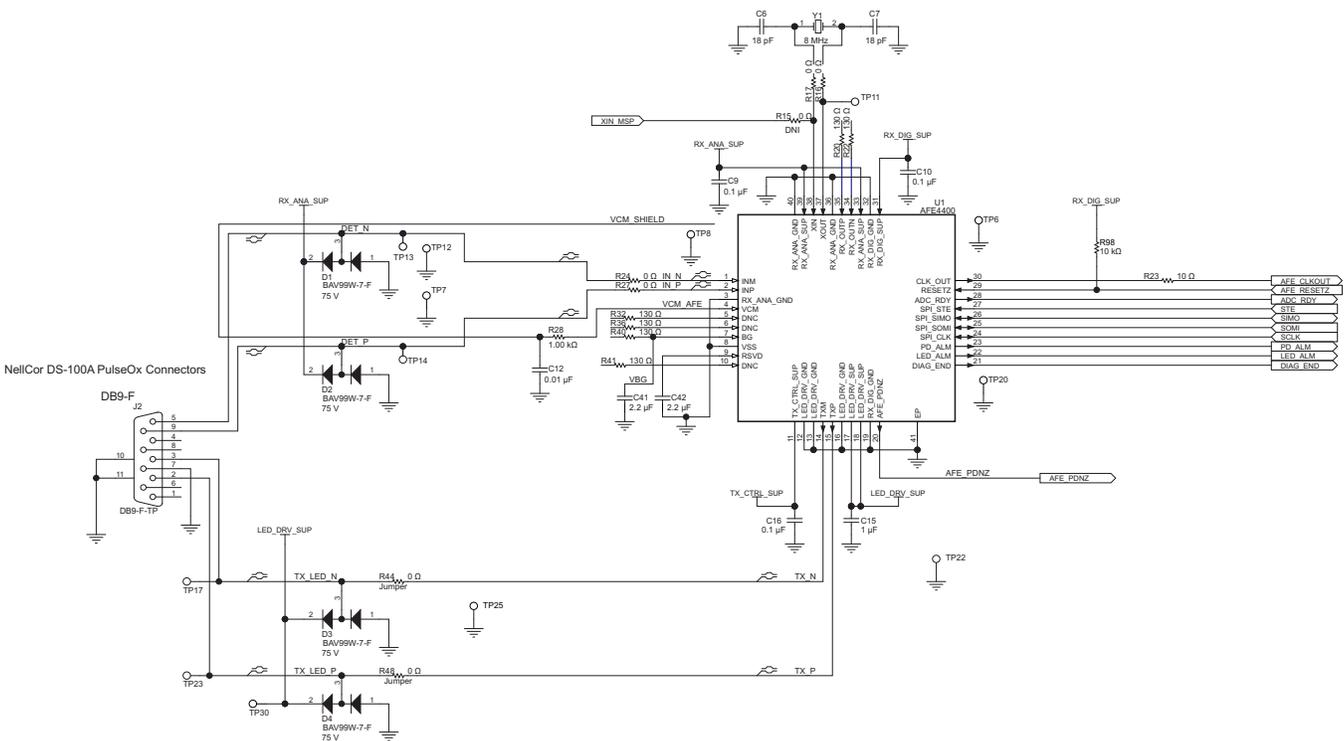
### 9.1 Application Information

The AFE4400 can be used for measuring SPO2 and for monitoring heart rate. The high dynamic range of the device enables measuring SPO2 with a high degree of accuracy even under low-perfusion (ac-to-dc ratio) conditions. An SPO2 measurement system involves two different wavelength LEDs—usually Red and IR. By computing the ratio of the ac to dc at the two different wavelengths, the SPO2 can be calculated. Heart rate monitoring systems can also benefit from the high dynamic range of the device, which enables capturing a high-fidelity pulsating signal even in cases where the signal strength is low.

For more information on application guidelines, refer to the [AFE44x0SPO2EVM User's Guide \(SLAU480\)](#).

### 9.2 Typical Application

Device connections in a typical application are shown in [Figure 108](#). Refer to the [AFE44x0SPO2EVM User's Guide \(SLAU480\)](#) for more details. The schematic in [Figure 108](#) is a part of the AFE44x0SPO2EVM and shows a cabled application in which the LEDs and photodiode are connected to the AFE4400 through a cable. However, in an application without cables, the LEDs and photodiode can be directly connected to the TXP, TXN and INP, INN pins directly, as shown in the [Design Requirements](#) section.



NOTE: The following signals must be considered as two sets of differential pairs and routed as adjacent signals within each pair: TXM, TXP and INM, INP.

INM and INP must be guarded with VCM\_SHIELD the signal. Run the VCM\_SHIELD signal to the DB9 connector and back to the device.

**Figure 108. AFE44x0SPO2EVM: Connections to the AFE4490**

## Typical Application (continued)

### 9.2.1 Design Requirements

An SPO2 application usually involves a Red LED and an IR LED. These LEDs can be connected either in the common anode configuration or H-bridge configuration to the TXP, TXN pins. [Figure 109](#) shows common anode configuration and [Figure 110](#) shows H-bridge configuration.

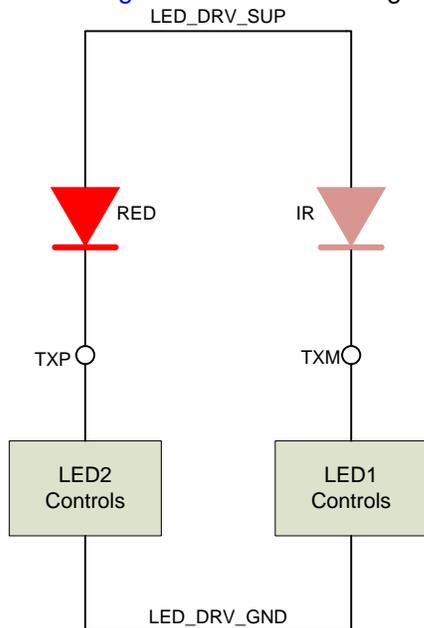


Figure 109. LEDs in Common Anode Configuration

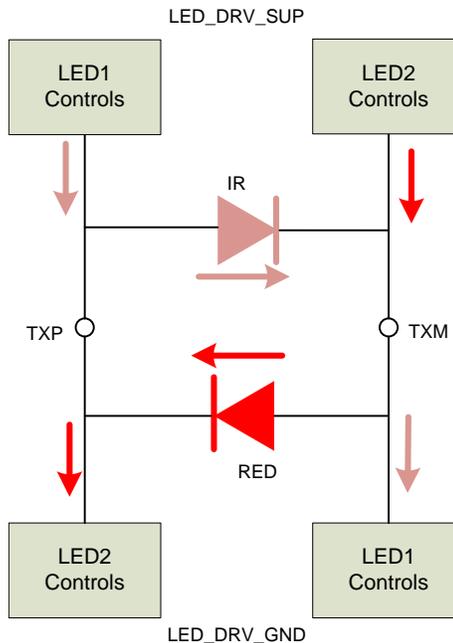


Figure 110. LEDs in H-Bridge Configuration

### 9.2.2 Detailed Design Procedure

The photodiode receives the light from both the Red and IR phases and usually has good sensitivities at both these wavelengths.

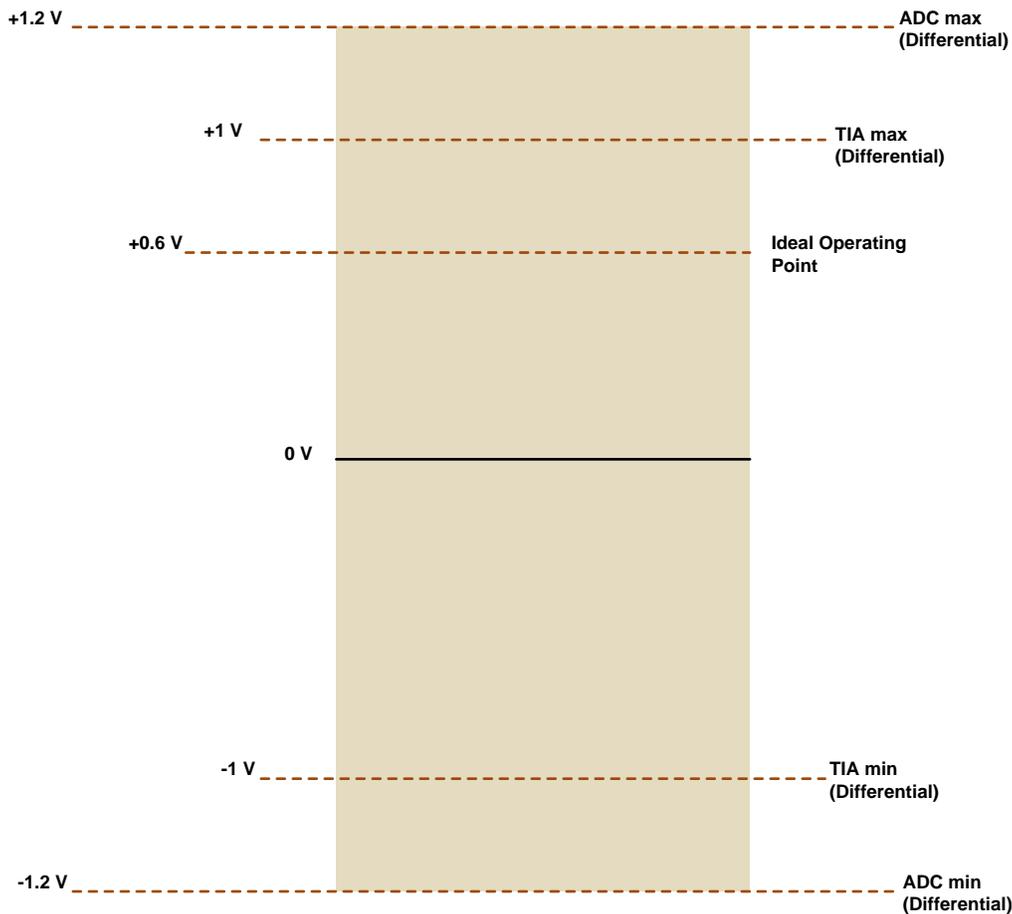
The photodiode connected in this manner operates in zero bias because of the negative feedback from the transimpedance amplifier. The connections of the photodiode to the AFE inputs are shown in [Figure 111](#).



Figure 111. Photodiode Connection

### Typical Application (continued)

The signal current generated by the photodiode is converted into a voltage by the transimpedance amplifier, which has a programmable transimpedance gain. The rest of the signal chain then presents a voltage to the ADC. The full-scale output of the transimpedance amplifier is  $\pm 1$  V and the full-scale input to the ADC is  $\pm 1.2$  V. An automatic gain control loop can be used to set the target dc voltage at the ADC input to approximately 50% of full scale. This type of AGC loop can control a combination of LED current and TIA gain to achieve this target value; see [Figure 112](#).



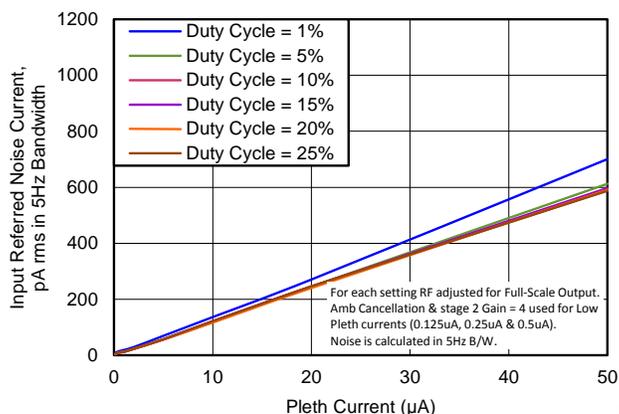
**Figure 112. AGC Loop**

The ADC output is a 22-bit code that is obtained by discarding the two MSBs of the 24-bit registers. The data format is binary two's complement format, MSB first. TI recommends that the input to the ADC does not exceed  $\pm 1$  V (which is approximately 80% full-scale) because the TIA has a full-scale range of  $\pm 1$  V.

## Typical Application (continued)

### 9.2.3 Application Curve

The dc component of the current from the PPG signal is referred to as *Pleth* (short for photoplethysmography) current. The input-referred noise current (referred differentially to the INP, INN inputs) as a function of the Pleth current is shown in Figure 113 at a PRF of 100 Hz and for various duty cycles of LED pulsing. For example, a duty cycle of 25% refers to a case where the LED is pulsed for 25% of the pulse repetition period and the receiver samples the photodiode current for the same period of time. The noise shown in Figure 113 is the integrated noise over a 5-Hz bandwidth from dc.



**Figure 113. Input-Referred Noise Current vs Pleth Current (PRF = 100 Hz)**

## 10 Power Supply Recommendations

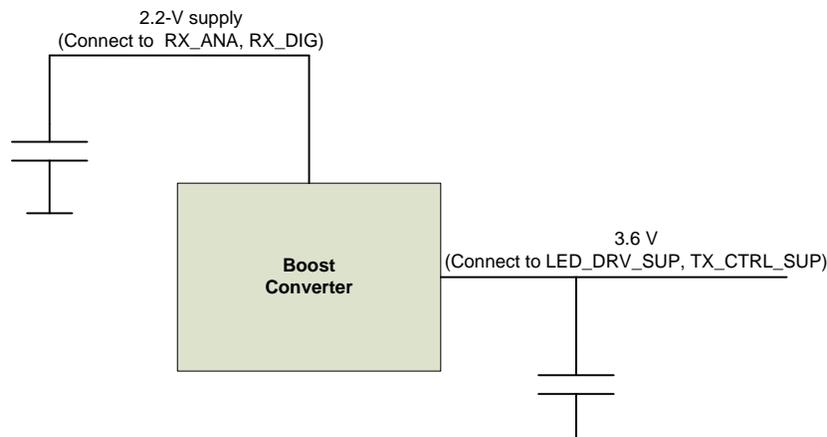
The AFE4400 has two sets of supplies: the receiver supplies (RX\_ANA\_SUP, RX\_DIG\_SUP) and the transmitter supplies (TX\_CTRL\_SUP, LED\_DRV\_SUP). The receiver supplies can be between 2.0 V to 3.6 V, whereas the transmitter supplies can be between 3.0 V to 5.25 V. Another consideration that determines the minimum allowed value of the transmitter supplies is the forward voltage of the LEDs being driven. The current source and switches inside the AFE require voltage headroom that mandates the transmitter supply to be a few hundred millivolts higher than the LED forward voltage. TX\_REF is the voltage that governs the generation of the LED current from the internal reference voltage. Choosing the lowest allowed TX\_REF setting reduces the additional headroom required but results in higher transmitter noise. Other than for the highest-end clinical SPO2 applications, this extra noise resulting from a lower TX\_REF setting can be acceptable.

LED\_DRV\_SUP and TX\_CTRL\_SUP are recommended to be tied together to the same supply (between 3.0 V to 5.25 V). The external supply (connected to the common anode of the two LEDs) must be high enough to account for the forward drop of the LEDs as well as the voltage headroom required by the current source and switches inside the AFE. In most cases, this voltage is expected to fall below 5.25 V; thus the external supply can be the same as LED\_DRV\_SUP. However, there may be cases (for instance when two LEDs are connected in series) where the voltage required on the external supply is higher than 5.25 V. Such a case must be handled with care to ensure that the voltage on the TXP and TXN pins remains less than 5.25 V and never exceeds the supply voltage of LED\_DRV\_SUP, TX\_CTRL\_SUP by more than 0.3 V.

Many scenarios of power management are possible.

**Case 1:** The LED forward voltage is such that a voltage of 3.3 V is acceptable on LED\_DRV\_SUP. In this case, a single 3.3-V supply can be used to drive all four pins (RX\_ANA\_SUP, RX\_DIG\_SUP, TX\_CTRL\_SUP, LED\_DRV\_SUP). Care should be taken to provide some isolation between the transmit and receive supplies because LED\_DRV\_SUP carries the high-switching current from the LEDs.

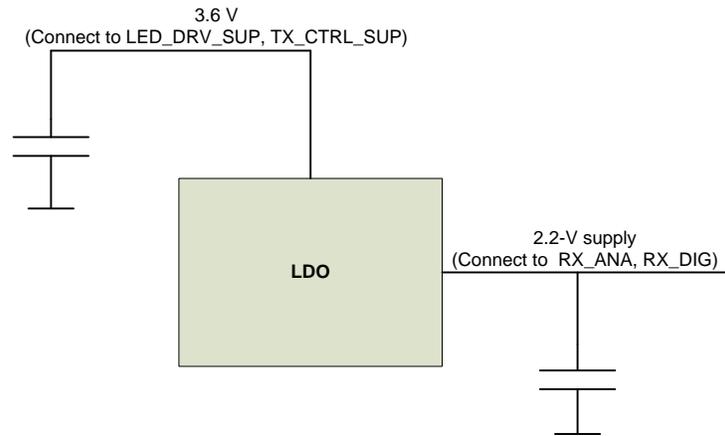
**Case 2:** A low-voltage supply of 2.2 V is available in the system. In this case, a boost converter can be used to derive the voltage for LED\_DRV\_SUP, as shown in [Figure 114](#).



**Figure 114. Boost Converter**

The boost converter requires a clock (usually in the megahertz range) and there is usually a ripple at the boost converter output at this switching frequency. While this frequency is much higher than the signal frequency of interest (which is at maximum a few tens of hertz around dc), a small fraction of this switching noise can possibly alias to the low-frequency band. Therefore, TI strongly recommends that the switching frequency of the boost converter be offset from every multiple of the PRF by at least 20 Hz. This offset can be ensured by choosing the appropriate PRF.

**Case 3:** In cases where a high-voltage supply is available in the system, a buck converter or an LDO can be used to derive the voltage levels required to drive RX\_ANA and RX\_DIG, as shown in [Figure 115](#).



**Figure 115. Buck Converter or an LDO**

For more information on power-supply recommendations, see the [AFE44x0SPO2EVM User's Guide \(SLAU480\)](#).

## 11 Layout

### 11.1 Layout Guidelines

Some key layout guidelines are mentioned below:

1. TXP, TXN are fast-switching lines and should be routed away from sensitive reference lines as well as from the INP, INN inputs.
2. If the INP, INN lines are required to be routed over a long trace, TI recommends that VCM be used as a shield for the INP, INN lines.
3. The device can draw high-switching currents from the LED\_DRV\_SUP pin. Therefore, TI recommends having a decoupling capacitor electrically close to the pin.

### 11.2 Layout Example

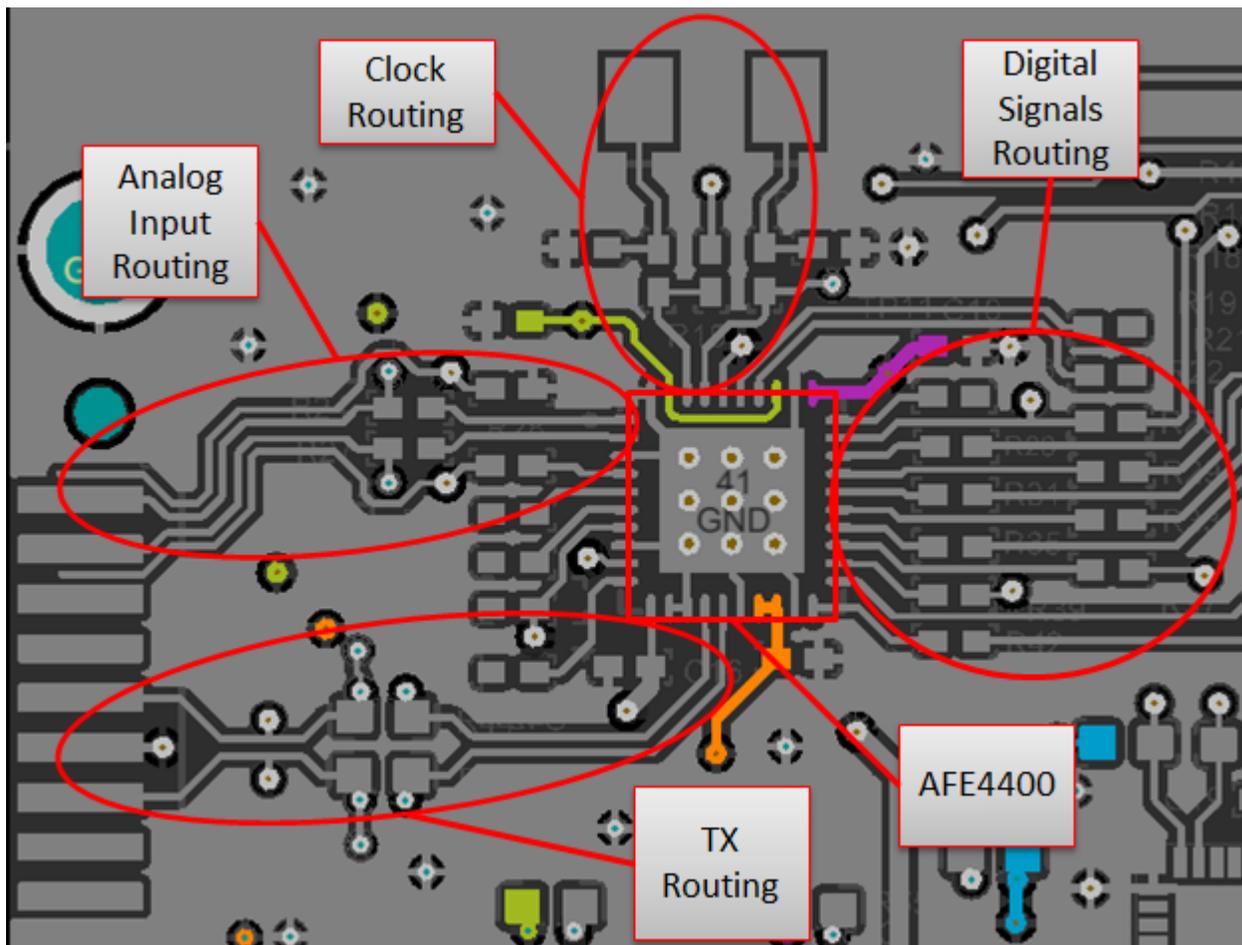


Figure 116. Typical Layout of the AFE4400 Board

## 12 器件和文档支持

### 12.1 Trademarks

SPI is a trademark of Motorola.

All other trademarks are the property of their respective owners.

### 12.2 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

### 12.3 术语表

[SLYZ022](#) — *TI* 术语表。

这份术语表列出并解释术语、首字母缩略词和定义。

## 13 机械封装和可订购信息

以下页中包括机械封装和可订购信息。 这些信息是针对指定器件可提供的最新数据。 这些数据会在无通知且不对本文档进行修订的情况下发生改变。 欲获得该数据表的浏览器版本，请查阅左侧的导航栏。

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
AFE4400RHAR	ACTIVE	VQFN	RHA	40	2500	RoHS & Green	NIPDAU	Level-3-260C-168 HR	0 to 70	AFE4400	<a href="#">Samples</a>
AFE4400RHAT	ACTIVE	VQFN	RHA	40	250	RoHS & Green	NIPDAU	Level-3-260C-168 HR	0 to 70	AFE4400	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

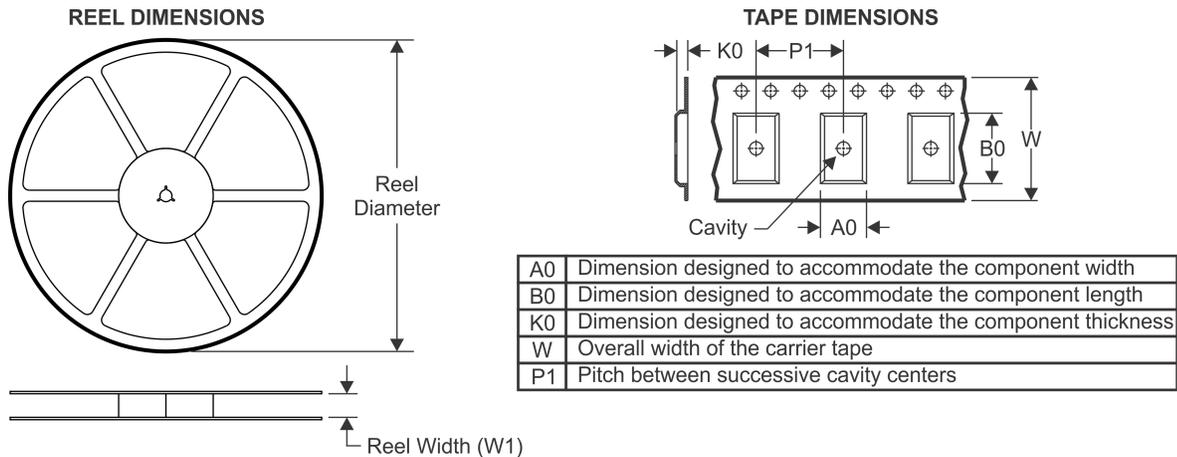
(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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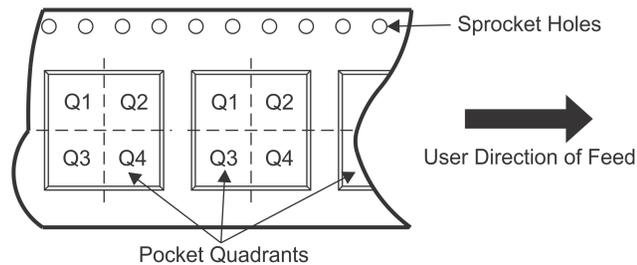
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## TAPE AND REEL INFORMATION

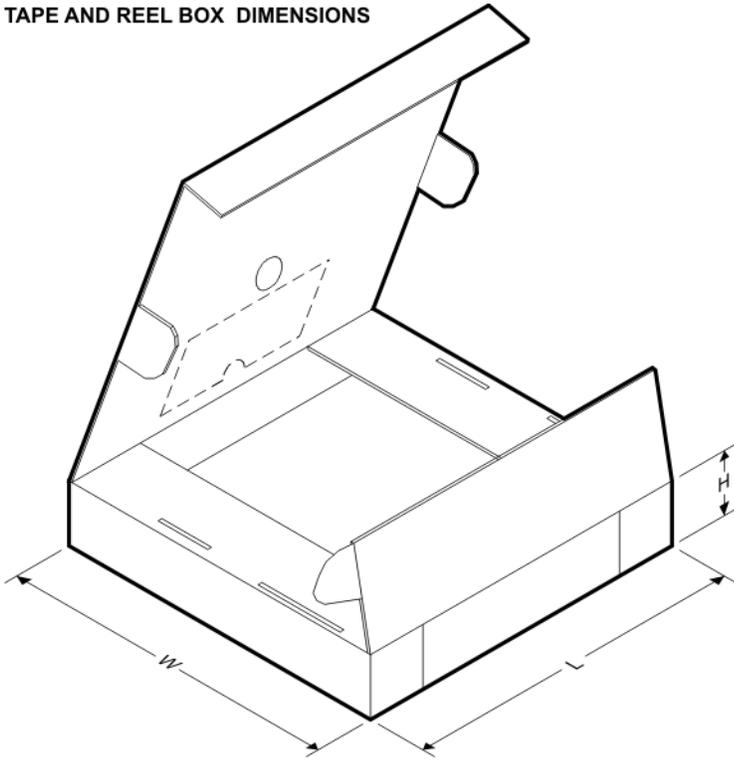


### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
AFE4400RHAR	VQFN	RHA	40	2500	330.0	16.4	6.3	6.3	1.1	12.0	16.0	Q2
AFE4400RHAT	VQFN	RHA	40	250	180.0	16.4	6.3	6.3	1.1	12.0	16.0	Q2

**TAPE AND REEL BOX DIMENSIONS**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
AFE4400RHAR	VQFN	RHA	40	2500	367.0	367.0	38.0
AFE4400RHAT	VQFN	RHA	40	250	210.0	185.0	35.0

## GENERIC PACKAGE VIEW

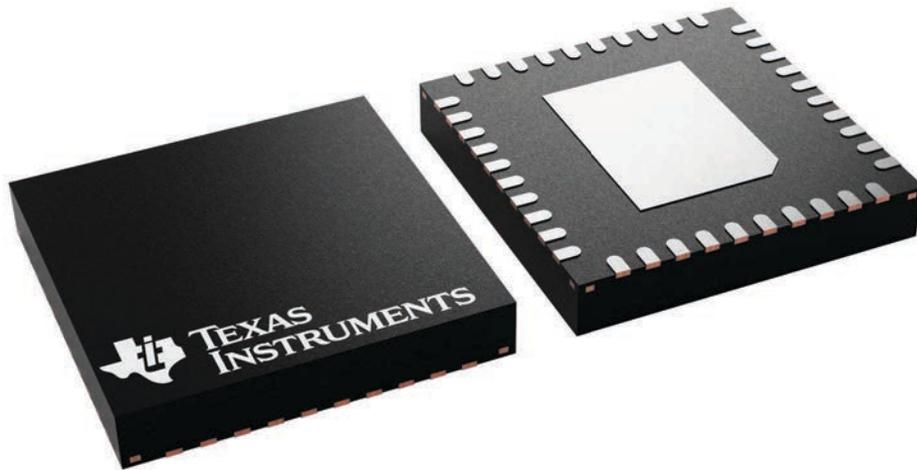
**RHA 40**

**VQFN - 1 mm max height**

6 x 6, 0.5 mm pitch

PLASTIC QUAD FLATPACK - NO LEAD

This image is a representation of the package family, actual package may vary.  
Refer to the product data sheet for package details.





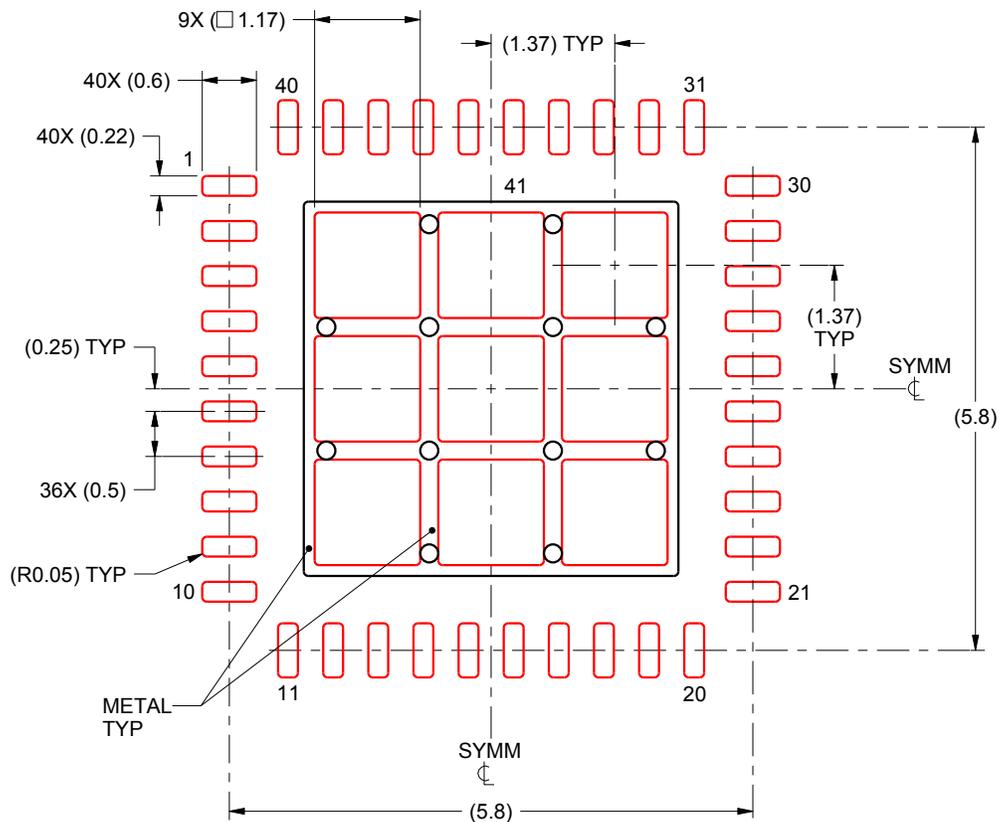


# EXAMPLE STENCIL DESIGN

RHA0040B

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



**SOLDER PASTE EXAMPLE**  
BASED ON 0.125 mm THICK STENCIL

EXPOSED PAD 41:  
72% PRINTED SOLDER COVERAGE BY AREA UNDER PACKAGE  
SCALE:12X

4219052/A 06/2016

NOTES: (continued)

5. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

## 重要声明和免责声明

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